

## LM2984C Microprocessor Power Supply System

### General Description

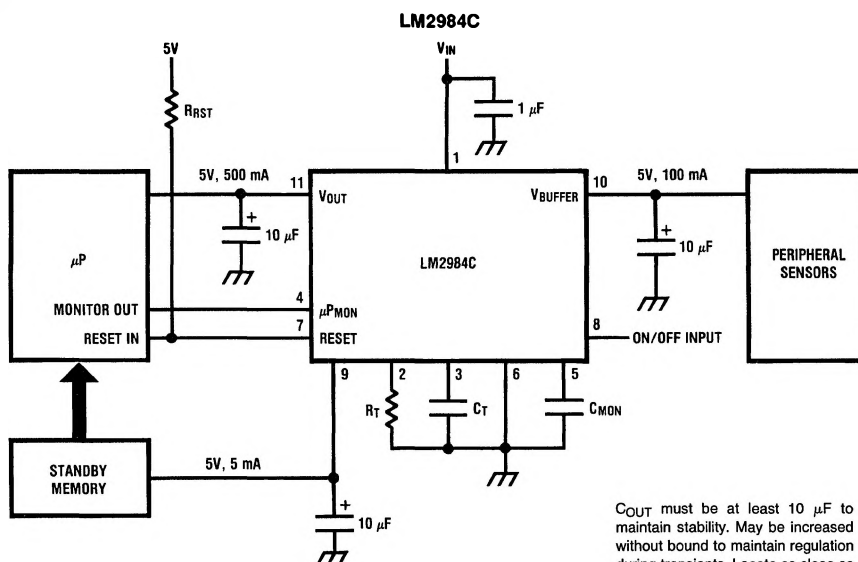
The LM2984C positive voltage regulator features three independent and tracking outputs capable of delivering the power for logic circuits, peripheral sensors and standby memory in a typical microprocessor system. The LM2984C includes circuitry which monitors both its own high-current output and also an external  $\mu\text{P}$ . If any error conditions are sensed in either, a reset error flag is set and maintained until the malfunction terminates. Since these functions are included in the same package with the three regulators, a great saving in board space can be realized in the typical microprocessor system. The LM2984C also features very low dropout voltages on each of its three regulator outputs (0.6V at the rated output current). Furthermore, the quiescent current can be reduced to 1 mA in the standby mode. Designed also for vehicular applications, the LM2984C and all regulated circuitry are protected from reverse battery installations or 2-battery jumps. Familiar regulator features such as short circuit and thermal overload protection are

also provided. Fixed outputs of 5V are available in the plastic TO-220 power package.

### Features

- Three low dropout tracking regulators
- Output current in excess of 500 mA
- Low quiescent current standby regulator
- Microprocessor malfunction RESET flag
- Delayed RESET on power-up
- Accurate pretrimmed 5V outputs
- Reverse battery protection
- Overvoltage protection
- Reverse transient protection
- Short circuit protection
- Internal thermal overload protection
- ON/OFF switch for high current outputs
- 100% electrical burn-in in thermal limit

### Typical Application Circuit



**Order Number LM2984CT**  
**See NS Package Number TA11A**

$C_{OUT}$  must be at least  $10 \mu\text{F}$  to maintain stability. May be increased without bound to maintain regulation during transients. Locate as close as possible to the regulator. This capacitor must be rated over the same operating temperature range as the regulator. The equivalent series resistance (ESR) of this capacitor should be less than  $1\Omega$  over the expected operating temperature range.

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## Absolute Maximum Ratings

If Military/Aerospace specified devices are required, contact the National Semiconductor Sales Office/Distributors for availability and specifications.

### Input Voltage

Survival Voltage (< 100 ms)	35V
Operational Voltage	26V

### Internal Power Dissipation

Operating Temperature Range ( $T_A$ )	Internally Limited 0°C to +125°C
Maximum Junction Temperature (Note 1)	150°C
Storage Temperature Range	-65°C to +150°C
Lead Temperature (Soldering, 10 sec.)	230°C
ESD rating is to be determined.	

## Electrical Characteristics

$V_{IN} = 14V$ ,  $I_{OUT} = 5\text{ mA}$ ,  $C_{OUT} = 10\text{ }\mu\text{F}$ ,  $T_J = 25^\circ\text{C}$  (Note 6) unless otherwise indicated

Parameter	Conditions	Typical	Tested Limit (Note 2)	Design Limit (Note 3)	Units
<b><math>V_{OUT}</math> (Pin 11)</b>					
Output Voltage	$5\text{ mA} \leq I_O \leq 500\text{ mA}$ $6V \leq V_{IN} \leq 26V$	5.00	4.85 5.15	4.75 5.25	$V_{min}$ $V_{max}$
Line Regulation	$9V \leq V_{IN} \leq 16V$	2	25		$mV_{max}$
	$7V \leq V_{IN} \leq 26V$	5	50		$mV_{max}$
Load Regulation	$5\text{ mA} \leq I_{OUT} \leq 500\text{ mA}$	12	50		$mV_{max}$
Output Impedance	$250\text{ mA}_{dc}$ and $10\text{ mA}_{rms}$ , $f_o = 120\text{ Hz}$	24			$m\Omega$
Quiescent Current	$I_{OUT} = 500\text{ mA}$	38	100		$mA_{max}$
	$I_{OUT} = 250\text{ mA}$	14	50		$mA_{max}$
Output Noise Voltage	$10\text{ Hz} - 100\text{ kHz}$ , $I_{OUT} = 100\text{ mA}$	100			$\mu V$
Long Term Stability		20			$mV/1000\text{ hr}$
Ripple Rejection	$f_o = 120\text{ Hz}$	70	60		$dB_{min}$
Dropout Voltage	$I_{OUT} = 500\text{ mA}$	0.53	0.80	1.00	$V_{max}$
	$I_{OUT} = 250\text{ mA}$	0.28	0.50	0.60	$V_{max}$
Current Limit		0.92	0.75		$A_{min}$
Maximum Operational Input Voltage	Continuous DC	32	26	26	$V_{min}$
Maximum Line Transient	$V_{OUT} \leq 6V$ , $R_{OUT} = 100\Omega$	45	35	35	$V_{min}$
Reverse Polarity Input Voltage DC	$V_{OUT} \geq -0.6V$ , $R_{OUT} = 100\Omega$	-30	-15	-15	$V_{min}$
Reverse Polarity Input Voltage Transient	$T \leq 100\text{ ms}$ , $R_{OUT} = 100\Omega$	-55	-35	-35	$V_{min}$

**Electrical Characteristics** (Continued) $V_{IN} = 14V$ ,  $I_{buf} = 5\text{ mA}$ ,  $C_{buf} = 10\text{ }\mu\text{F}$ ,  $T_j = 25^\circ\text{C}$  (Note 6) unless otherwise indicated

Parameter	Conditions	Typical	Tested Limit (Note 2)	Design Limit (Note 3)	Units
<b><math>V_{buffer}</math> (Pin 10)</b>					
Output Voltage	$5\text{ mA} \leq I_o \leq 100\text{ mA}$ $6V \leq V_{IN} \leq 26V$	5.00	4.85 5.15	4.75 5.25	$V_{min}$ $V_{max}$
Line Regulation	$9V \leq V_{IN} \leq 16V$	2	25		$mV_{max}$
	$7V \leq V_{IN} \leq 26V$	5	50		$mV_{max}$
Load Regulation	$5\text{ mA} \leq I_{buf} \leq 100\text{ mA}$	15	50		$mV_{max}$
Output Impedance	$50\text{ mA}_{dc}$ and $10\text{ mA}_{rms}$	200			$m\Omega$
Quiescent Current	$I_{buf} = 100\text{ mA}$	8.0	15.0		$mA_{max}$
Output Noise Voltage	10 Hz–100 kHz, $I_{OUT} = 100\text{ mA}$	100			$\mu\text{V}$
Long Term Stability		20			$mV/1000\text{ hr}$
Ripple Rejection	$f_o = 120\text{ Hz}$	70	60		$dB_{min}$
Dropout Voltage	$I_{buf} = 100\text{ mA}$	0.35	0.50	0.60	$V_{max}$
Current Limit		0.23	0.15		$A_{min}$
Maximum Operational Input Voltage	Continuous DC	32	26	26	$V_{min}$
Maximum Line Transient	$V_{buf} \leq 6V$ , $R_{buf} = 100\Omega$	45	35	35	$V_{min}$
Reverse Polarity Input Voltage DC	$V_{buf} \geq -0.6V$ , $R_{buf} = 100\Omega$	-30	-15	-15	$V_{min}$
Reverse Polarity Input Voltage Transient	$T \leq 100\text{ ms}$ , $R_{buf} = 100\Omega$	-55	-35	-35	$V_{min}$

**Electrical Characteristics** $V_{IN} = 14V$ ,  $I_{stby} = 1\text{ mA}$ ,  $C_{stby} = 10\text{ }\mu\text{F}$ ,  $T_j = 25^\circ\text{C}$  (Note 6) unless otherwise indicated

Parameter	Conditions	Typical	Tested Limit (Note 2)	Design Limit (Note 3)	Units
<b><math>V_{standby}</math> (Pin 9)</b>					
Output Voltage	$1\text{ mA} \leq I_o \leq 7.5\text{ mA}$ $6V \leq V_{IN} \leq 26V$	5.00	4.85 5.15	4.75 5.25	$V_{min}$ $V_{max}$
Line Regulation	$9V \leq V_{IN} \leq 16V$	2	25		$mV_{max}$
	$7V \leq V_{IN} \leq 26V$	5	50		$mV_{max}$
Load Regulation	$0.5\text{ mA} \leq I_{stby} \leq 7.5\text{ mA}$	6	50		$mV_{max}$
Output Impedance	$5\text{ mA}_{dc}$ and $1\text{ mA}_{rms}$ , $f_o = 120\text{ Hz}$	0.9			$\Omega$
Quiescent Current	$I_{stby} = 7.5\text{ mA}$	1.2	2.0		$mA_{max}$
	$I_{stby} = 2\text{ mA}$	0.9	1.5		$mA_{max}$

**Electrical Characteristics** (Continued) $V_{IN} = 14V$ ,  $I_{stby} = 1\text{ mA}$ ,  $C_{stby} = 10\text{ }\mu\text{F}$ ,  $T_j = 25^\circ\text{C}$  (Note 6) unless otherwise indicated

Parameter	Conditions	Typical	Tested Limit (Note 2)	Design Limit (Note 3)	Units
<b><math>V_{standby}</math> (Continued)</b>					
Output Noise Voltage	10 Hz–100 kHz, $I_{stby} = 1\text{ mA}$	100			$\mu\text{V}$
Long Term Stability		20			mV/1000 hr
Ripple Rejection	$f_o = 120\text{ Hz}$	70	60		dB <sub>min</sub>
Dropout Voltage	$I_{stby} = 1\text{ mA}$	0.26	0.50	0.50	$V_{max}$
Dropout Voltage	$I_{stby} = 7.5\text{ mA}$	0.38	0.60	0.70	$V_{max}$
Current Limit		15	12		mA <sub>min</sub>
Maximum Operational Input Voltage	$4.5V \leq V_{stby} \leq 6V$ $R_{stby} = 1000\Omega$	45	35	35	$V_{min}$
Maximum Line Transient	$V_{stby} \leq 6V$ , $R_{stby} = 1000\Omega$	45	35	35	$V_{min}$
Reverse Polarity Input Voltage DC	$V_{stby} \geq -0.6V$ , $R_{stby} = 1000\Omega$	-30	-15	-15	$V_{min}$
Reverse Polarity Input Voltage Transient	$T \leq 100\text{ ms}$ , $R_{stby} = 1000\Omega$	-55	-35	-35	$V_{min}$

**Electrical Characteristics** $V_{IN} = 14V$ ,  $T_j = 25^\circ\text{C}$  (Note 6)  $C_{OUT} = 10\text{ }\mu\text{F}$ ,  $C_{buf} = 10\text{ }\mu\text{F}$ ,  $C_{stby} = 10\text{ }\mu\text{F}$  unless otherwise specified

Parameter	Conditions	Typical	Tested Limit (Note 2)	Design Limit (Note 3)	Units
<b>Tracking and Isolation</b>					
Tracking $V_{OUT} - V_{stby}$	$I_{OUT} \leq 500\text{ mA}$ , $I_{buf} = 5\text{ mA}$ , $I_{stby} \leq 7.5\text{ mA}$	$\pm 30$	$\pm 100$		mV <sub>max</sub>
Tracking $V_{buf} - V_{stby}$	$I_{OUT} = 5\text{ mA}$ , $I_{buf} \leq 100\text{ mA}$ , $I_{stby} \leq 7.5\text{ mA}$	$\pm 30$	$\pm 100$		mV <sub>max</sub>
Tracking $V_{OUT} - V_{buf}$	$I_{OUT} \leq 500\text{ mA}$ , $I_{buf} \leq 100\text{ mA}$ , $I_{stby} = 1\text{ mA}$	$\pm 30$	$\pm 100$		mV <sub>max</sub>
Isolation* $V_{buf}$ from $V_{OUT}$	$R_{OUT} = 1\Omega$ , $I_{buf} \leq 100\text{ mA}$	5.00	4.50 5.50		$V_{min}$ $V_{max}$
Isolation* $V_{stby}$ from $V_{OUT}$	$R_{OUT} = 1\Omega$ , $I_{stby} \leq 7.5\text{ mA}$	5.00	4.50 5.50		$V_{min}$ $V_{max}$
Isolation* $V_{OUT}$ from $V_{buf}$	$R_{buf} = 1\Omega$ , $I_{OUT} \leq 500\text{ mA}$	5.00	4.50 5.50		$V_{min}$ $V_{max}$
Isolation* $V_{stby}$ from $V_{buf}$	$R_{buf} = 1\Omega$ , $I_{stby} \leq 7.5\text{ mA}$	5.00	4.50 5.50		$V_{min}$ $V_{max}$

\*Isolation refers to the ability of the specified output to remain within the tested limits when the other output is shorted to ground.

**Electrical Characteristics** (Continued)

$V_{IN} = 14V$ ,  $I_{OUT} = 5\text{ mA}$ ,  $I_{buf} = 5\text{ mA}$ ,  $I_{stby} = 5\text{ mA}$ ,  $R_t = 130k$ ,  $C_t = 0.33\text{ }\mu\text{F}$ ,  $C_{mon} = 0.47\text{ }\mu\text{F}$ ,  $T_j = 25^\circ\text{C}$  (Note 6) unless otherwise specified

Parameter	Conditions	Typical	Tested Limit (Note 2)	Design Limit (Note 3)	Units
<b>Computer Monitor/Reset Functions</b>					
$I_{reset\ Low}$	$V_{IN} = 4V$ , $V_{rst} = 0.4V$	5	2	1	$\text{mA}_{min}$
$V_{reset\ Low}$	$V_{IN} = 4V$ , $I_{rst} = 1\text{ mA}$	0.10	0.40		$V_{max}$
$R_t$ voltage	(Pin 2)	1.22	1.15		$V_{min}$
		1.22	1.30		$V_{max}$
Power On Reset Delay	$V_{\mu P_{mon}} = 5V$ ( $T_{dly} = 1.2 R_t C_t$ )	50	45		$\text{ms}_{min}$
		50	55		$\text{ms}_{max}$
$V_{OUT\ Low}$ Reset Threshold	(Note 4)	4.00	3.60		$V_{min}$
		4.00	4.40		$V_{max}$
$V_{OUT\ High}$ Reset Threshold	(Note 4)	5.50	5.25		$V_{min}$
		5.50	6.00		$V_{max}$
Reset Output Leakage	$V_{\mu P_{mon}} = 5V$ , $V_{rst} = 12V$	0.01	1		$\mu\text{A}_{max}$
$\mu\text{P}_{mon}$ Input Current (Pin 4)	$V_{\mu P_{mon}} = 2.4V$	7.5	25		$\mu\text{A}_{max}$
	$V_{\mu P_{mon}} = 0.4V$	0.01	10		$\mu\text{A}_{max}$
$\mu\text{P}_{mon}$ Input Threshold Voltage		1.22	0.80	0.80	$V_{min}$
		1.22	2.00	2.00	$V_{max}$
$\mu\text{P}$ Monitor Reset Oscillator Period	$V_{\mu P_{mon}} = 0V$ ( $T_{window} = 0.82 R_t C_{mon}$ )	50	45		$\text{ms}_{min}$
		50	55		$\text{ms}_{max}$
$\mu\text{P}$ Monitor Reset Oscillator Pulse Width	$V_{\mu P_{mon}} = 0V$ ( $RESET_{pw} = 2000 C_{mon}$ )	1.0	0.7	0.5	$\text{ms}_{min}$
		1.0	1.3	2.0	$\text{ms}_{max}$
Minimum $\mu\text{P}$ Monitor Input Pulse Width	(Note 5)	2			$\mu\text{s}_{max}$
Reset Fall Time	$R_{rst} = 10k$ , $V_{rst} = 5V$ , $C_{rst} \leq 10\text{ pF}$	0.20	1.00		$\mu\text{s}_{max}$
Reset Rise Time	$R_{rst} = 10k$ , $V_{rst} = 5V$ , $C_{rst} \leq 10\text{ pF}$	0.60	1.00		$\mu\text{s}_{max}$
On/Off Switch Input Current (Pin 8)	$V_{ON} = 2.4V$	7.5	25		$\mu\text{A}_{max}$
	$V_{ON} = 0.4V$	0.01	10		$\mu\text{A}_{max}$
On/Off Switch Input Threshold Voltage		1.22	0.80	0.80	$V_{min}$
		1.22	2.00	2.00	$V_{max}$

**Note 1:** Thermal resistance without a heatsink for junction-to-case temperature is  $3^\circ\text{C/W}$ . Thermal resistance case-to-ambient is  $40^\circ\text{C/W}$ .

**Note 2:** Tested Limits are guaranteed and 100% production tested.

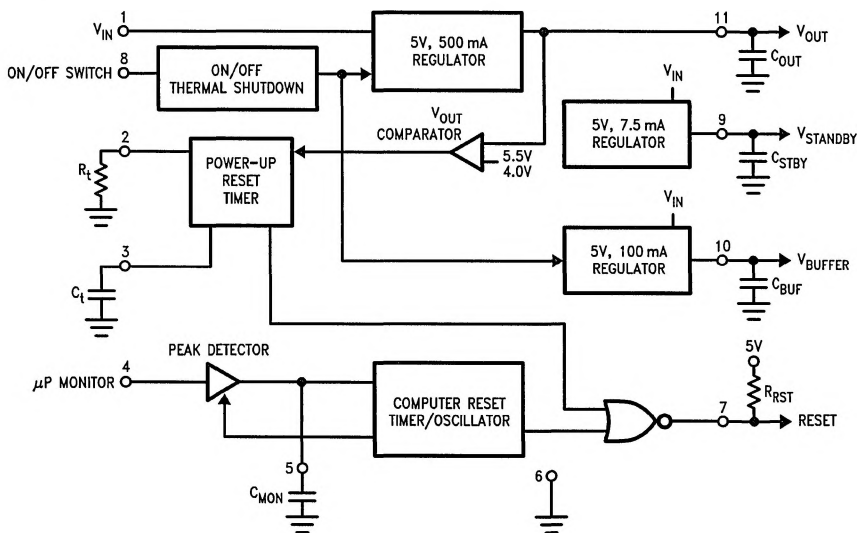
**Note 3:** Design Limits are guaranteed (but not 100% production tested) over the indicated temperature and supply voltage range. These limits are not used to calculate outgoing quality levels.

**Note 4:** An internal comparator detects when the main regulator output ( $V_{OUT}$ ) drops below 4.0V or rises above 5.5V. If either condition exists at the output, the Reset Error Flag is held low until the error condition has terminated. The Reset Error Flag is then allowed to go high again after a delay set by  $R_t$  and  $C_t$ . (See Applications Section.)

**Note 5:** This parameter is a measure of how short a pulse can be detected at the  $\mu\text{P}$  Monitor Input. This parameter is primarily influenced by the value of  $C_{mon}$ . (See Typical Performance Characteristics and Applications Section.)

**Note 6:** To ensure constant junction temperature, low duty cycle pulse testing is used.

## Block Diagram



TL/H/8821-2

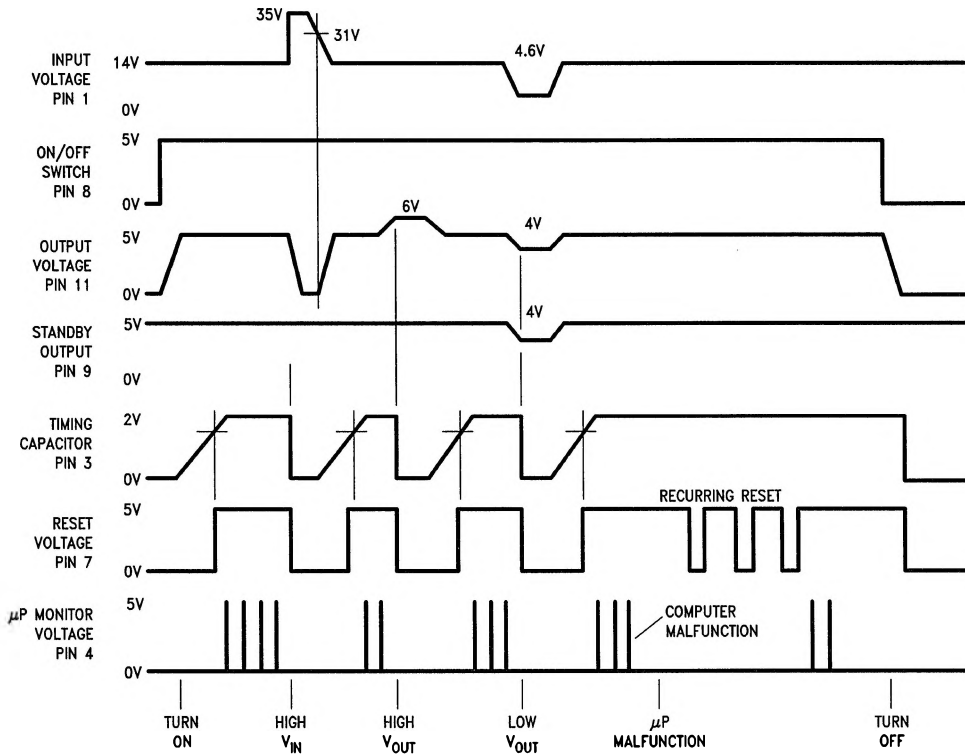
## Pin Description

Pin No.	Pin Name	Comments
1	$V_{IN}$	Positive supply input voltage
2	$R_t$	Sets internal timing currents
3	$C_t$	Sets power-up reset delay timing
4	$\mu P_{mon}$	Microcomputer monitor input
5	$C_{mon}$	Sets $\mu C$ monitor timing
6	Ground	Regulator ground
7	Reset	Reset error flag output
8	ON/OFF	Enables/disables high current regulators
9	$V_{standby}$	Standby regulator output (7.5 mA)
10	$V_{buffer}$	Buffer regulator output (100 mA)
11	$V_{OUT}$	Main regulator output (500 mA)

## External Components

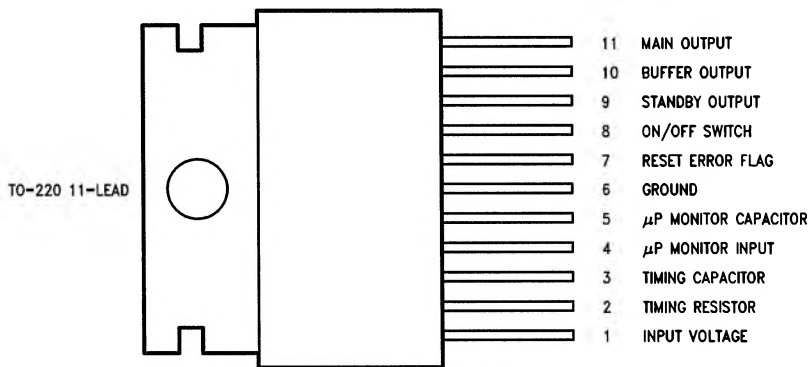
Component	Typical Value	Component Range	Comments
$C_{IN}$	1 $\mu F$	0.47 $\mu F$ –10 $\mu F$	Required if device is located far from power supply filter.
$R_t$	130k	24k–1.2M	Sets internal timing currents.
$C_t$	0.33 $\mu F$	0.033 $\mu F$ –3.3 $\mu F$	Sets power-up reset delay.
$C_{tc}$	0.01 $\mu F$	0.001 $\mu F$ –0.1 $\mu F$	Establishes time constant of AC coupled computer monitor.
$R_{tc}$	10k	1k–100k	Establishes time constant of AC coupled computer monitor. (See applications section.)
$C_{mon}$	0.47 $\mu F$	0.047 $\mu F$ –4.7 $\mu F$	Sets time window for computer monitor. Also determines period and pulse width of computer malfunction reset. (See applications section.)
$R_{rst}$	10k	5k–100k	Load for open collector reset output. Determined by computer reset input requirements.
$C_{stby}$	10 $\mu F$	10 $\mu F$ –no bound	A 10 $\mu F$ is required for stability but larger values can be used to maintain regulation during transient conditions.
$C_{buf}$	10 $\mu F$	10 $\mu F$ –no bound	A 10 $\mu F$ is required for stability but larger values can be used to maintain regulation during transient conditions.
$C_{OUT}$	10 $\mu F$	10 $\mu F$ –no bound	A 10 $\mu F$ is required for stability but larger values can be used to maintain regulation during transient conditions.

## Typical Circuit Waveforms



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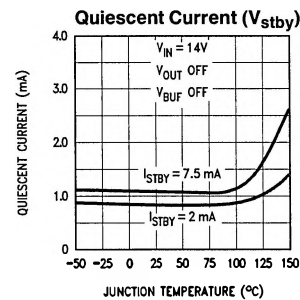
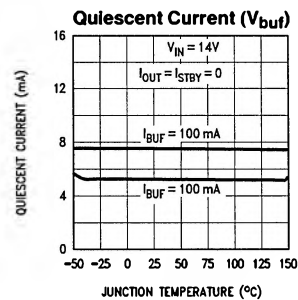
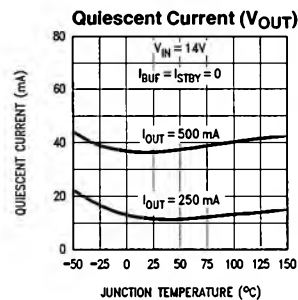
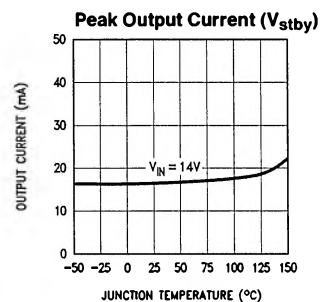
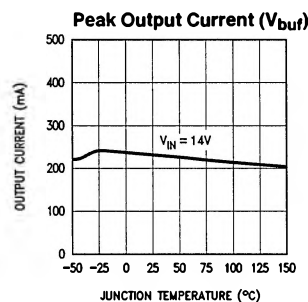
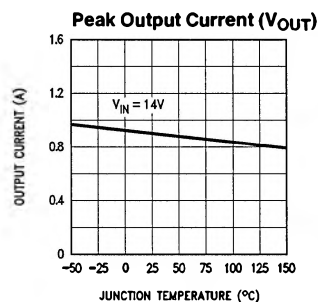
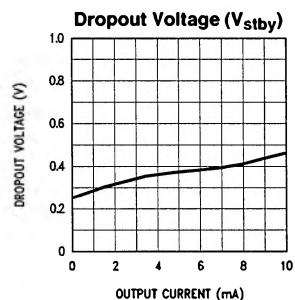
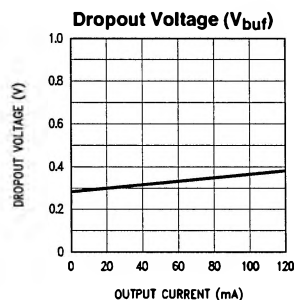
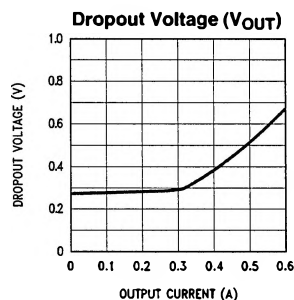
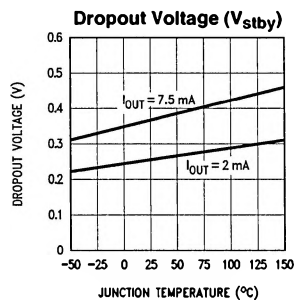
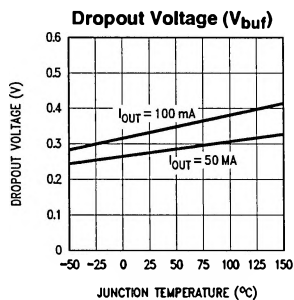
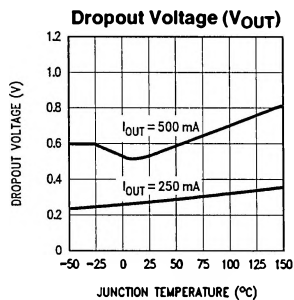
## Connection Diagram



Order Number LM2984CT  
See NS Package Number T11A

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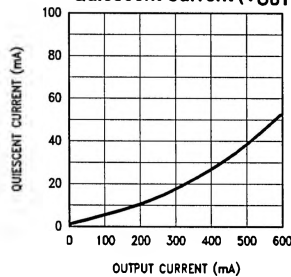
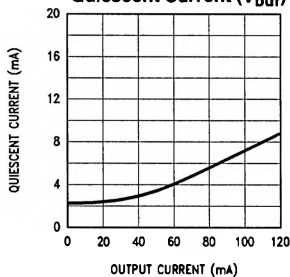
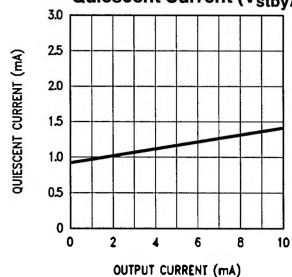
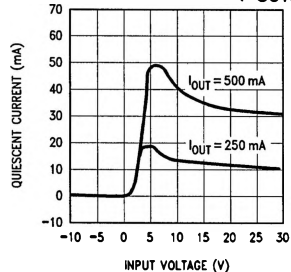
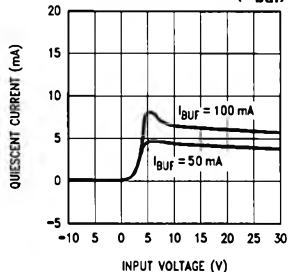
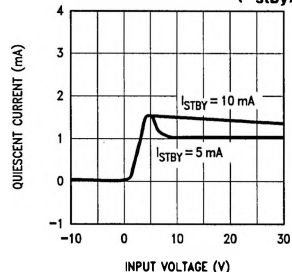
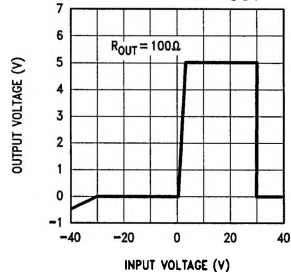
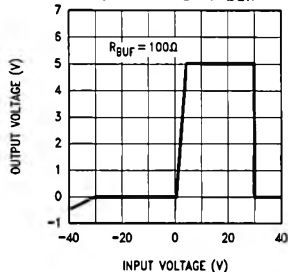
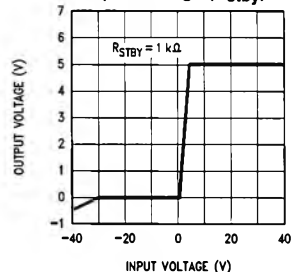
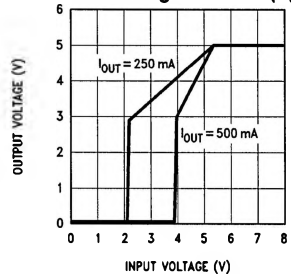
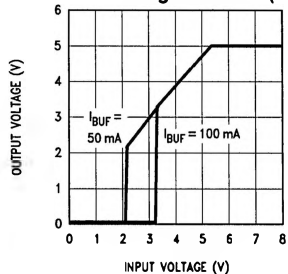
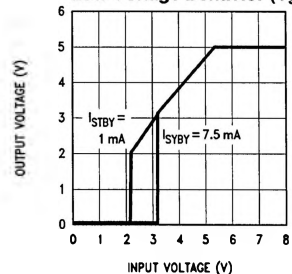
# Typical Performance Characteristics



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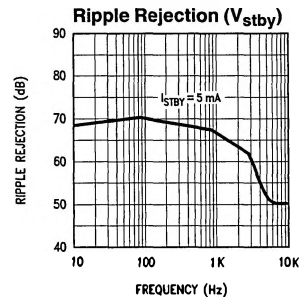
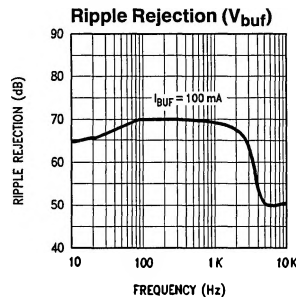
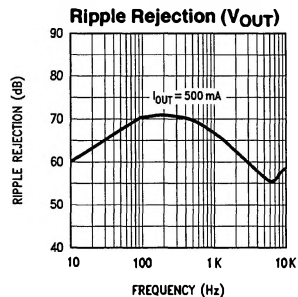
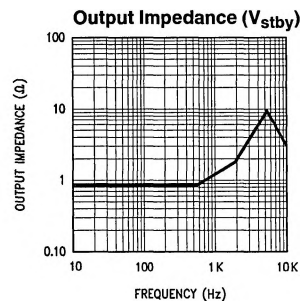
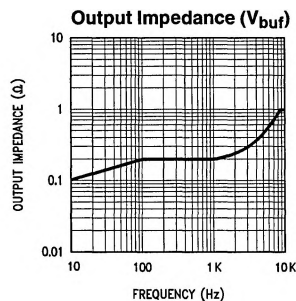
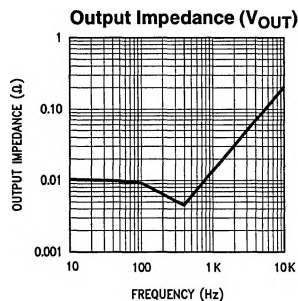
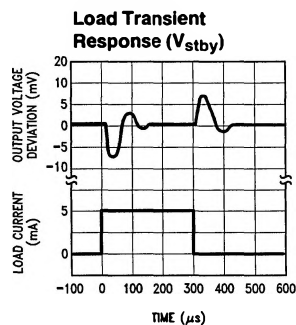
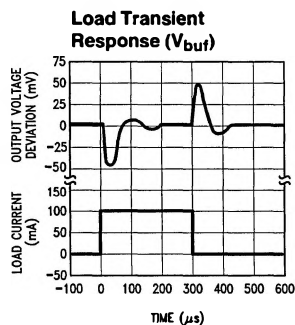
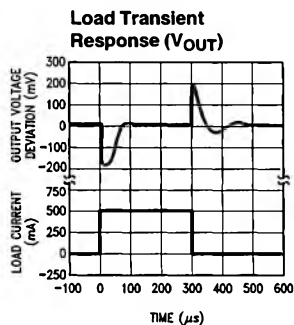
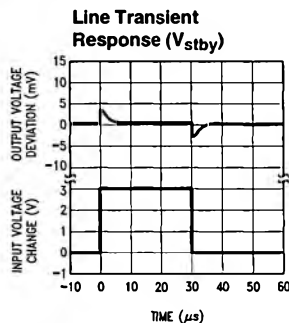
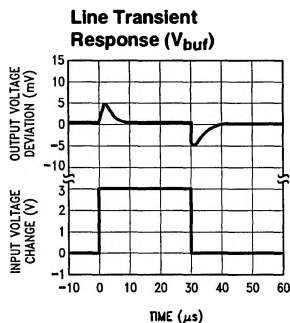
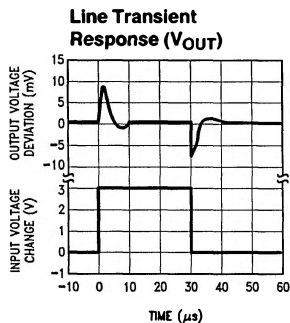


# Typical Performance Characteristics (Continued)

Quiescent Current ( $V_{OUT}$ )Quiescent Current ( $V_{BUF}$ )Quiescent Current ( $V_{STBY}$ )Quiescent Current ( $V_{OUT}$ )Quiescent Current ( $V_{BUF}$ )Quiescent Current ( $V_{STBY}$ )Output Voltage ( $V_{OUT}$ )Output Voltage ( $V_{BUF}$ )Output Voltage ( $V_{STBY}$ )Low Voltage Behavior ( $V_{OUT}$ )Low Voltage Behavior ( $V_{BUF}$ )Low Voltage Behavior ( $V_{STBY}$ )

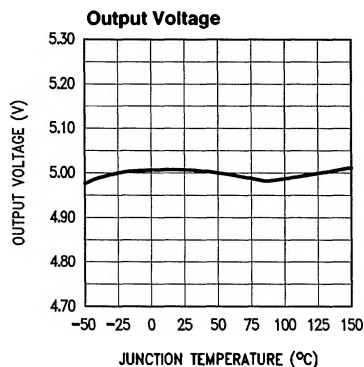
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# Typical Performance Characteristics (Continued)

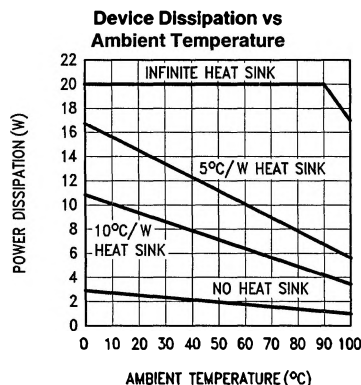


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## Typical Performance Characteristics (Continued)



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## Application Hints

### OUTPUT CAPACITORS

The LM2984C output capacitors are required for stability. Without them, the regulator outputs will oscillate, sometimes by many volts. Though the 10  $\mu$ F shown are the minimum recommended values, actual size and type may vary depending upon the application load and temperature range. Capacitor effective series resistance (ESR) also affects the IC stability. Since ESR varies from one brand to the next, some bench work may be required to determine the minimum capacitor value to use in production. Worst case is usually determined at the minimum ambient temperature and the maximum load expected.

Output capacitors can be increased in size to any desired value above the minimum. One possible purpose of this would be to maintain the output voltages during brief conditions of negative input transients that might be characteristic of a particular system.

Capacitors must also be rated at all ambient temperatures expected in the system. Many aluminum type electrolytics will freeze at temperatures less than  $-30^{\circ}\text{C}$ , reducing their effective capacitance to zero. To maintain regulator stability down to  $-40^{\circ}\text{C}$ , capacitors rated at that temperature (such as tantalums) must be used.

Each output **must** be terminated by a capacitor, even if it is not used.

### STANDBY OUTPUT

The standby output is intended for use in systems requiring standby memory circuits. While the high current regulator outputs are controlled with the ON/OFF pin described later, the standby output remains on under all conditions as long as sufficient input voltage is supplied to the IC. Thus, memory and other circuits powered by this output remain unaffected by positive line transients, thermal shutdown, etc.

The standby regulator circuit is designed so that the quiescent current to the IC is very low ( $<1.5\text{ mA}$ ) when the other regulator outputs are off.

The capacitor on the output of this regulator can be increased without bound. This will help maintain the output voltage during negative input transients and will also help to reduce the noise on all three outputs. Because the other two track the standby output, therefore any noise reduction here will also reduce the other two noise voltages.

### BUFFER OUTPUT

The buffer output is designed to drive peripheral sensor circuitry in a  $\mu\text{P}$  system. It will track the standby and main regulator within a few millivolts in normal operation. Therefore, a peripheral sensor can be powered off this supply and have the same operating voltage as the  $\mu\text{P}$  system. This is important if a ratiometric sensor system is being used.

The buffer output can be short circuited while the other two outputs are in normal operation. This protects the  $\mu\text{P}$  system from disruption of power when a sensor wire, etc. is temporarily shorted to ground, i.e. only the sensor signal would be interrupted, while the  $\mu\text{P}$  and memory circuits would remain operational.

The buffer output is similar to the main output in that it is controlled by the ON/OFF switch in order to save power in the standby mode. It is also fault protected against overvoltage and thermal overload. If the input voltage rises above approximately 30V (e.g. load dump), this output will automatically shut down. This protects the internal circuitry and enables the IC to survive higher voltage transients than would otherwise be expected. Thermal shutdown is necessary since this output is one of the dominant sources of power dissipation in the IC.

### MAIN OUTPUT

The main output is designed to power relatively large loads, i.e. approximately 500 mA. It is therefore also protected against overvoltage and thermal overload.

This output will track the other two within a few millivolts in normal operation. It can therefore be used as a reference voltage for any signal derived from circuitry powered off the standby or buffer outputs. This is important in a ratiometric sensor system or any system requiring accurate matching of power supply voltages.

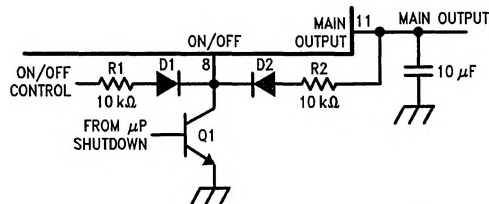
### ON/OFF SWITCH

The ON/OFF switch controls the main output and the buffer output. The threshold voltage is compatible with most logic families and has about 20 mV of hysteresis to insure 'clean' switching from the standby mode to the active mode and vice versa. This pin can be tied to the input voltage through a 10 k $\Omega$  resistor if the regulator is to be powered continuously.

## Application Hints (Continued)

### POWER DOWN OVERRIDE

Another possible approach is to use a diode in series with the ON/OFF signal and another in series with the main output in order to maintain power for some period of time after the ON/OFF signal has been removed (see Figure 1). When the ON/OFF switch is initially pulled high through diode D1, the main output will turn on and supply power through diode D2 to the ON/OFF switch effectively latching the main output. An open collector transistor Q1 is connected to the ON/OFF pin along with the two diodes and forces the regulators off after a period of time determined by the  $\mu P$ . In this way, the  $\mu P$  can override a power down command and store data, do housekeeping, etc. before reverting back to the standby mode.



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FIGURE 1. Power Down Override

### RESET OUTPUT

This output is an open collector NPN transistor which is forced low whenever an error condition is present at the main output or when a  $\mu P$  error is sensed (see  $\mu P$  Monitor section). If the main output voltage drops below 4V or rises above 5.5V, the RESET output is forced low and held low for a period of time set by two external components,  $R_t$  and  $C_t$ . There is a slight amount of hysteresis in these two threshold voltages so that the RESET output has a fast rise and fall time compatible with the requirements of most  $\mu P$  RESET inputs.

### DELAYED RESET

Resistor  $R_t$  and capacitor  $C_t$  set the period of time that the RESET output is held low after a main output error condition has been sensed. The delay is given by the formula:

$$T_{dly} = 1.2 R_t C_t \text{ (seconds)}$$

The delayed RESET will be initiated any time the main output is outside the 4V to 5.5V window, i.e. during power-up, short circuit, overvoltage, low line, thermal shutdown or power-down. The  $\mu P$  is therefore RESET whenever the output voltage is out of regulation. (It is important to note that a RESET is only initiated when the main output is in error. The buffer and standby outputs are not directly monitored for error conditions.)

### $\mu P$ MONITOR RESET

There are two distinct and independent error monitoring systems in the LM2984C. The one described above monitors the main regulator output and initiates a delayed RESET whenever this output is in error. The other error monitoring system is the  $\mu P$  watchdog. These two systems are OR'd together internally and both force the RESET output low when either type of error occurs.

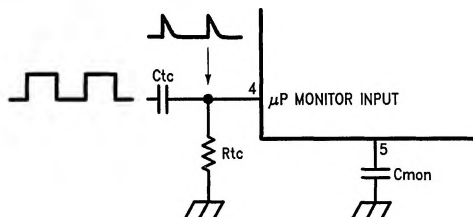
This watchdog circuitry continuously monitors a pin on the  $\mu P$  that generates a positive going pulse during normal operation. The period of this pulse is typically on the order of milliseconds and the pulse width is typically on the order of 10's of microseconds. If this pulse ever disappears, the watchdog circuitry will time out and a RESET low will be sent to the  $\mu P$ . The time out period is determined by two external components,  $R_t$  and  $C_{mon}$ , according to the formula:

$$T_{window} = 0.82 R_t C_{mon} \text{ (seconds)}$$

The width of the RESET pulse is set by  $C_{mon}$  and an internal resistor according to the following:

$$RESET_{pw} = 2000 C_{mon} \text{ (seconds)}$$

A square wave signal can also be monitored for errors by filtering the  $C_{mon}$  input such that only the positive edges of the signal are detected. Figure 2 is a schematic diagram of a typical circuit used to differentiate the input signal. Resistor  $R_{tc}$  and capacitor  $C_{tc}$  pass only the rising edge of the square wave and create a short positive pulse suitable for the  $\mu P$  monitor input. If the incoming signal continues in a high state or in a low state for too long a period of time, a RESET low will be generated.



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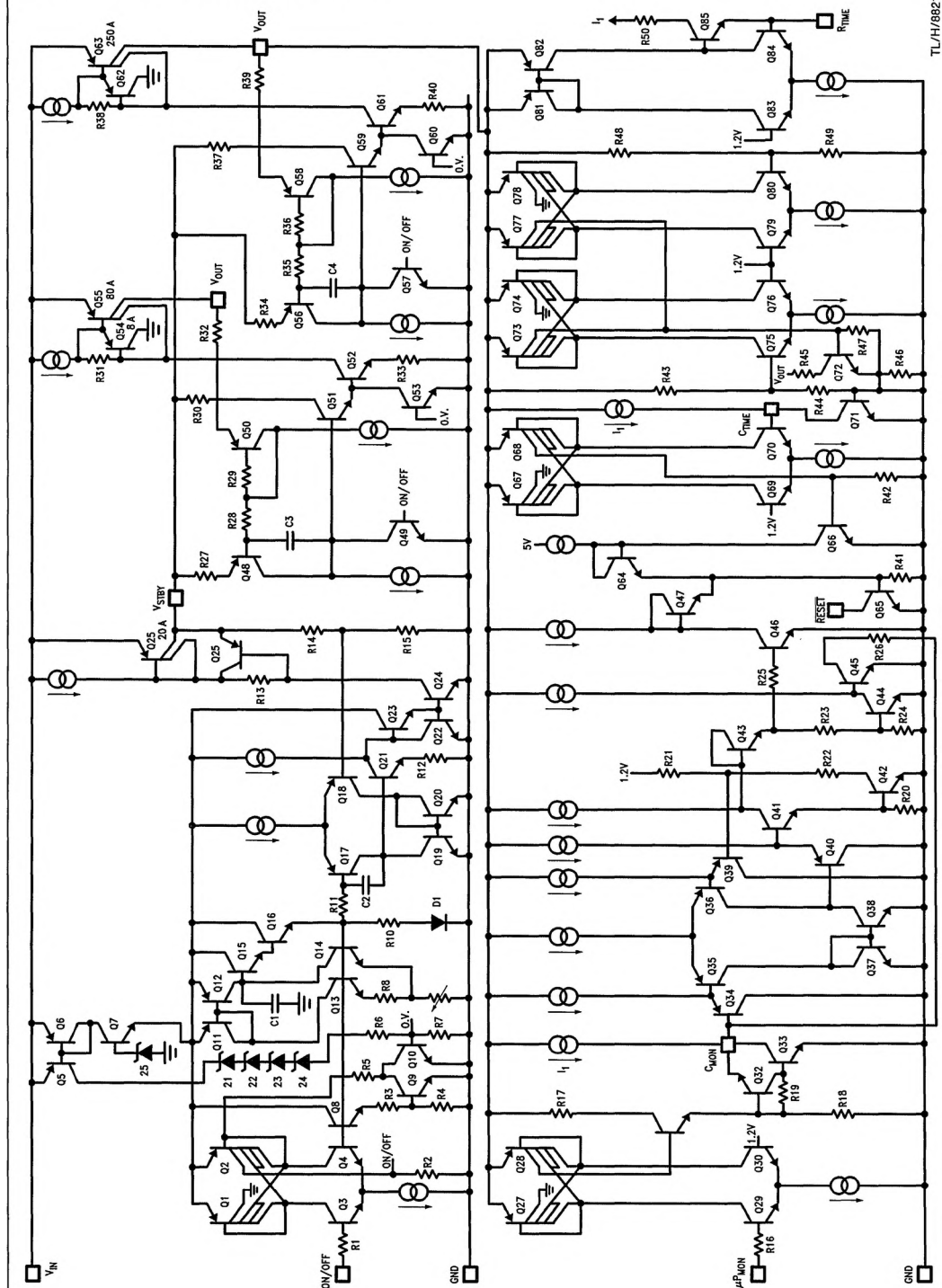
FIGURE 2. Monitoring Square Wave  $\mu P$  Signals

The threshold voltage and input characteristics of this pin are compatible with nearly all logic families.

There is a limit on the width of a pulse that can be reliably detected by the watchdog circuit. This is due to the output resistance of the transistor which discharges  $C_{mon}$  when a high state is detected at the input. The minimum detectable pulse width can be determined by the following formula:

$$PW_{min} = 20 C_{mon} \text{ (seconds)}$$

## Equivalent Schematic Diagram



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