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If Military/Aerospace specified d please contact the National S Office/Distributors for availability	emiconductor Sales	Soldering Information Dual-In-Line Package Soldering (10 sec)	+ 260°C	
Supply Voltage (LM386N-1, -3, LM386M-1) 15V		Small Outline Package		
Supply Voltage (LM386N-4) 22V		Vapor Phase (60 sec)	+ 215°0 + 220°	
Package Dissipation (Note 1) (LM386N) 1.25W		Infrared (15 sec)		
(LM386M) 0.73W		See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering sur		
Input Voltage	\pm 0.4V	face mount devices.	nous of soldering sur	
Storage Temperature	ge Temperature -65°C to +150°C			
Operating Temperature	0°C to +70°C	$\theta_{\rm JC}$ (DIP)	37°C/W	
Junction Temperature	+150°C	$\theta_{\rm JA}$ (DIP)	107°C/W	
		$\theta_{\rm JC}$ (SO Package)	35°C/V	
		θ_{JA} (SO Package)	172°C/V	

Electrical Characteristics $T_A = 25^{\circ}C$

Parameter	Conditions	Min	Тур	Max	Units
Operating Supply Voltage (V _S) LM386N-1, -3, LM386M-1 LM386N-4		4 5		12 18	V V
Quiescent Current (I _Q)	$V_{S} = 6V, V_{IN} = 0$		4	8	mA
Output Power (P _{OUT}) LM386N-1, LM386M-1 LM386N-3 LM386N-4	$\begin{split} V_S &= 6V, R_L = 8\Omega, \text{THD} = 10\% \\ V_S &= 9V, R_L = 8\Omega, \text{THD} = 10\% \\ V_S &= 16V, R_L = 32\Omega, \text{THD} = 10\% \end{split}$	250 500 700	325 700 1000		mW mW mW
Voltage Gain (A _V)	$V_{S} = 6V$, f = 1 kHz 10 μ F from Pin 1 to 8		26 46		dB dB
Bandwidth (BW)	$V_{S} = 6V$, Pins 1 and 8 Open		300		kHz
Total Harmonic Distortion (THD)	$V_S = 6V$, $R_L = 8\Omega$, $P_{OUT} = 125$ mW f = 1 kHz, Pins 1 and 8 Open		0.2		%
Power Supply Rejection Ratio (PSRR)	$V_{S} = 6V$, f = 1 kHz, $C_{BYPASS} = 10 \ \mu F$ Pins 1 and 8 Open, Referred to Output		50		dB
Input Resistance (R _{IN}) Input Bias Current (I _{BIAS})	$V_S = 6V$, Pins 2 and 3 Open		50 250		kΩ nA

Note 1: For operation in ambient temperatures above 25°C, the device must be derated based on a 150°C maximum junction temperature and 1) a thermal resistance of 80°C/W junction to ambient for the dual-in-line package and 2) a thermal resistance of 170°C/W for the small outline package.

Application Hints

GAIN CONTROL

To make the LM386 a more versatile amplifier, two pins (1 and 8) are provided for gain control. With pins 1 and 8 open the 1.35 k Ω resistor sets the gain at 20 (26 dB). If a capacitor is put from pin 1 to 8, bypassing the 1.35 k Ω resistor, the gain will go up to 200 (46 dB). If a resistor is placed in series with the capacitor, the gain can be set to any value from 20 to 200. Gain control can also be done by capacitively coupling a resistor (or FET) from pin 1 to ground.

Additional external components can be placed in parallel with the internal feedback resistors to tailor the gain and frequency response for individual applications. For example, we can compensate poor speaker bass response by frequency shaping the feedback path. This is done with a series RC from pin 1 to 5 (paralleling the internal 15 k Ω resistor). For 6 dB effective bass boost: R \cong 15 k Ω , the lowest value for good stable operation is R = 10 k Ω if pin 8 is open. If pins 1 and 8 are bypassed then R as low as 2 k Ω can be used. This restriction is because the amplifier is only compensated for closed-loop gains greater than 9.

INPUT BIASING

The schematic shows that both inputs are biased to ground with a 50 k Ω resistor. The base current of the input transistors is about 250 nA, so the inputs are at about 12.5 mV when left open. If the dc source resistance driving the LM386 is higher than 250 k Ω it will contribute very little additional offset (about 2.5 mV at the input, 50 mV at the output). If the dc source resistance is less than 10 k Ω , then shorting the unused input to ground will keep the offset low (about 2.5 mV at the input, 50 mV at the output). For dc source resistances between these values we can eliminate excess offset by putting a resistor from the unused input to ground, equal in value to the dc source resistance. Of course all offset problems are eliminate if the input is capacitively coupled.

When using the LM386 with higher gains (bypassing the 1.35 k Ω resistor between pins 1 and 8) it is necessary to bypass the unused input, preventing degradation of gain and possible instabilities. This is done with a 0.1 μF capacitor or a short to ground depending on the dc source resistance on the driven input.









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