

SNAS439B-AUGUST 2008-REVISED MARCH 2009

Boomer® Audio Power Amplifier Series Mono Class D Audio Subsystem with Earpiece LM49150 **Driver and Stereo Ground Referenced Headphone Amplifiers**

Check for Samples: LM49150, LM49150TLEVAL

FEATURES

- E²S class D amplifier
- Ground referenced headphone outputs eliminates output coupling capacitors
- I²C volume and mode control
- Mono earpiece amplifier
- Flexible output for speaker and headphone output
- 20-bump micro SMD package
- Soft enable function
- "Click and Pop" suppression circuitry

DESCRIPTION

- Thermal shutdown protection •
- Low supply current
- **Micro-power shutdown**

APPLICATIONS

- **Mobile Phones**
- **PDAs**
- **Portable Electronics**

The LM49150 is a fully integrated audio subsystem designed for portable handheld applications such as cellular phones. Part of National's Power Wise® product family, the LM49150 consumes very low power in the various modes of operation and still providing great audio performance. The LM49150 combines a 1.25W mono E²S (Enhanced Emission Suppression) class D amplifier, 135mW Class AB earpiece amplifier, 42mW/channel stereo ground reference headphone amplifiers, volume control, and mixing circuitry into a single device.

The filterless class D amplifier delivers 1.25W into an 8 Ω load with <1% THD+N with a 5V supply. The E²S class D amplifier features a patented, ultra low EMI PWM architecture that significantly reduces RF emissions while preserving audio quality. The 42mW/channel headphone drivers feature National's ground referenced architecture that creates a ground-referenced output from a single supply, eliminating the need for bulky and expensive DC-blocking capacitors, saving space and minimizing cost.

The LM49150 features a fully differential mono input, and two single-ended stereo inputs. The three inputs can be mixed/multiplexed to either the speaker or headphone amplifiers. Each input channel has an independent, 32step digital volume control. The headphone output stage features an additional, 8-step gain control, while the speaker output stage has a selectable 6dB or 12dB gain. The mixer, volume control and device mode select are controlled through an I²C compatible serial interface.

The LM49150's superior click and pop suppression eliminates audible transients on power-up/down and during shutdown. The LM49150 is available in a ultra-small 20-bump micro SMD package (2.225mm X 2.644mm).

	VALUE	UNIT		
Output power at V _{DD} = 5V:				
Speaker: R _L = 8Ω BTL, THD+N ≤ 1%	1.25	W (typ)		
Headphone: R _L = 32Ω SE, THD+N $\leq 1\%$	42	mW (typ)		
Earpiece: $R_L = 8\Omega$ SE, THD+N ≤ 1%	135	mW (typ)		
Output power at V _{DD} = 3.3V:				
Speaker: R _L = 8 Ω BTL, THD+N ≤ 1%	520	mW (typ)		

Table 1. Key Specifications



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Table 1. Key Specifications (continued)

Headphone: $R_L = 32\Omega BTL, THD+N \le 1\%$	42	mW (typ)
Earpiece: R _L = 8Ω SE, THD+N ≤ 1%	35	mW (typ)
Output Offset		
LS Mode	9	mV (typ)
HP Mode	1	mV (typ)
Earpiece	1	mV (typ)
Single Supply Operation (V _{DD})	2.7 to 5.5	V
I ² C Single Supply Operation	1.7 to 5.5	V

Typical Application



Figure 1. Typical Audio Amplifier Application Circuit-Output Capacitor-less

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Connection Diagram

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XY — Date Code TT — Die Traceability G — Boomer K7 — LM49150TL





	Table 2. Bump Descriptions				
Bump	Name	Description			
A1	I ² CV _{DD}	I ² C Power Supply			
A2	GND	Ground			
A3	LSOUT-	Inverting Loudspeaker Output			
A4	LSOUT+	Non-Inverting Loudspeaker Output			
B1	V _{DD}	Analog Power Supply			
B2	SDA	I ² C Data Input			
B3	SCL	I ² C Clock Input			
B4	LSV _{DD}	Loudspeaker Power Supply			
C1	INL	Left Channel Input			
C2	INR	Right Channel Input			
C3	BYPASS	Mid-Rail Supply Bypass			
C4	CPV _{SS}	Charge Pump Output			
D1	INM-	Mono Channel Inverting Input			
D2	V _{O(LDO)}	Internal LDO Output			
D3	CPN	Charge Pump Flying Capacitor - Negative Terminal			
D4	CPP	Charge Pump Flying Capacitor - Positive Terminal			
E1	INM+	Mono Channel Non-Inverting Input			
E2	HPR	Right Channel Headphone Amplifier Output			
E3	HPL	Left Channel Headphone Amplifier Output			
E4	CPGND	Charge Pump Ground			



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings (1) (2)

Supply Voltage ⁽¹⁾	6.0V
Storage Temperature	−65°C to +150°C
Input Voltage	-0.3 to V _{DD} +0.3
Power Dissipation ⁽³⁾	Internally Limited
ESD Rating ⁽⁴⁾	2.0kV
ESD Rating ⁽⁵⁾	200V
Junction Temperature	150°C
Soldering Information	See AN-1112 "Micro SMD Wafer Level Chip Scale Package"
Thermal Resistance	
θ _{JA} (typ) - TLA20KGA	46.1°C/W

- (1) "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional and the device should not be operated beyond such conditions. All voltages are measured with respect to the ground pin, unless otherwise specified
- (2) The Electrical Characteristics tables list guaranteed specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not guaranteed.
- (3) The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX}, θ_{JA}, and the ambient temperature, T_A. The maximum allowable power dissipation is P_{DMAX} = (T_{JMAX} T_A) / θ_{JA} or the number given in *Absolute Maximum Ratings*, whichever
- (4) Human body model, applicable std. JESD22-A114C.
- (5) Machine model, applicable std. JESD22-A115-A.

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Operating Ratings	
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NSTRUMENTS

Temperature Range	−40°C to 85°C
Supply Voltage	$2.7V \le V_{DD} \le 5.5V$
Supply Voltage (I ² C)	$1.7 \text{V} \le \text{I}^2 \text{CV}_{\text{DD}} \le 5.5 \text{V}$

TEXAS INSTRUMENTS

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Electrical Characteristics 3.3V⁽¹⁾

The following specifications apply for $V_{DD} = LSV_{DD} = 3.3V$, $A_V = 0dB$, Loudspeaker $R_L = 15\mu H + 8\Omega + 15\mu H$, Earpiece $R_L = 8\Omega$, f = 1kHz, unless otherwise specified. Limits apply for $T_A = 25C$. LS = Loudspeaker, HP = Headphone, EP = Earpiece.

Symbol	Parameter	Conditions	LM4	9150	Units (Limits)
Symbol	Parameter	Conditions	Typical ⁽²⁾	Limits ⁽³⁾	
		V _{IN} = 0, No Load			
		LS mode 1	3.7	5	mA (max
		HP mode 8	4.7	6.7	mA (max
I _{DD}	Supply Current	EP Bypass mode	0.8	1.2	mA (max
		LS + HP mode 5 and mode 10	7	9.5	mA (max
		LS mode 1, GAMP_SD = 1	3	4	mA (max
		HP mode 8, GAMP_SD = 1	4.3	6.1	mA (max
I _{SD}	Shutdown Current		0.04	1	μA (max)
		$V_{IN} = 0V, LS, R_L = 8\Omega$ LS Gain = 6dB, Stereo mode 10	9	40	mV (max)
V _{OS}	Output Offset Voltage	$V_{IN} = 0V$, HP, $R_L = 32\Omega$ Ground Referenced, Stereo mode 10	1	5	mV (max)
		V_{IN} = 0V, EP Bypass only, R_L = 8 Ω	0.8	5	mV (max
	Output Power	LS mode 1, THD+N = 1%, f = 1kHz LS Gain = 6dB, R_L = 4 Ω	845		mW
		LS mode 1, THD+N = 1%, f = 1kHz LS Gain = 6dB, R_L = 8 Ω	520	450	mW (min
Po		HP mode 8, THD+N = 1%, f = 1kHz HP Attenuation = 0dB, R_L = 16 Ω	42		mW
		HP mode 8, THD+N = 1%, f = 1kHz HP Attenuation = 0dB, R_L = 32 Ω	43	39	mW (min
		EP Bypass only, THD+N = 1%, f = 1kHz R_L = 8 Ω	35	28	mW (min
		LS mode 1, f = 1kHz P _{OUT} = 250mW; R _L = 8 Ω	0.02		%
THD+N	Total Harmonic Distortion + Noise	HP mode 8, f = 1kHz P _{OUT} = 20mW; R _L = 32Ω	0.009		%
		EP Bypass only, f = 1kHz P _{OUT} = 20mW; R _L = 8 Ω	0.15		%
η	Efficiency	LS output	88		%
		A-weighted, inputs terminated to AC GND, Output referred	ed		
		EP Bypass	11		μV
		LS; Mode 1	41		μV
∈out	Output Noise	LS; Mode 2	41		μV
001		LS; Mode 3	43		μV
		HP; Mode 4	9		μV
		HP; Mode 8	10		μV
		HP; Mode 12	12		μV

- (2) Typical values represent most likely parametric norms at $T_A = +25^{\circ}C$, and at the *Recommended Operation Conditions* at the time of product characterization and are not guaranteed.
- (3) Datasheet min/max specification limits are guaranteed by test or statistical analysis.
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⁽¹⁾ The Electrical Characteristics tables list guaranteed specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not guaranteed.



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Electrical Characteristics 3.3V ⁽¹⁾ (continued)

The following specifications apply for $V_{DD} = LSV_{DD} = 3.3V$, $A_V = 0dB$, Loudspeaker $R_L = 15\mu H + 8\Omega + 15\mu H$, Earpiece $R_L = 8\Omega$, f = 1kHz, unless otherwise specified. Limits apply for $T_A = 25C$. LS = Loudspeaker, HP = Headphone, EP = Earpiece.

0	Baramatar	Conditions	LM4	9150	Units
Symbol	Parameter	Conditions	Typical ⁽²⁾	Limits (3)	(Limits)
		$V_{RIPPLE} = 200 \text{mV}_{PP}$; f = 217Hz, R _L = 8 Ω , 0 All audio inputs terminated to AC GND, ou		I	
		EP Bypass	95		dB
		Loudspeaker Output; LS Gain = 6dB		U	
		LS; Mode 1	72		dB
PSRR	Power Supply Rejection Ratio	LS; Mode 2	67		dB
-		LS; Mode 3	71		dB
		Headphone Output, HP Attenuation = 0dB	6		
		HP; Mode 4	91		dB
		HP; Mode 8	83		dB
		HP; Mode 12	81		dB
	Volume Control Step Size Error		±0.2		dB
		Maximum Attenuation	-92		dB
	Digital Volume Control Range	Volume Step 2	-46.5	-49 -44	dB (min) dB (max)
		Maximum Gain	18	17 19	dB (min) dB (max)
		HP	98		dB
A _M	Mute Attenuation	LS	98		dB
	Mono Channel Input Impedance	Maximum gain setting	12.9	10 15	kΩ (min) kΩ (max)
Z _{IN}	$L_{\rm IN}$ and $R_{\rm IN}$ Input Impedance	Maximum attenuation setting	111	90 130	kΩ (min) kΩ (max)
	EP Bypass Resistance		62	50 80	kΩ (min) kΩ (max)
		f = 217Hz, V_{CM} = 1 V_{P-P} , R_L = 8 Ω EP Bypass	55		dB
CMRR	Common-Mode Rejection Ratio	f = 217Hz, V_{CM} = 1 V_{P-P} , R_L = 8 Ω LS, Mode 1	55		dB
		f = 217Hz, V _{CM} = 1V _{P-P} , R _L = 32 Ω HP, Mode 4	61		dB
X _{TALK}	Crosstalk	HP mode 8; $P_0 = 12mW$ $R_L = 32\Omega$, f = 1kHz	78		dB
т	Turn-On Time	C _B = 2.2µF, HP, Normal Turn-On Mode	27		ms
T _{ON}		$C_B = 2.2\mu$ F, HP, Fast Turn-On Mode	15		ms



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Electrical Characteristics 5.0V (1) (2)

The following specifications apply for $V_{DD} = LSV_{DD} = 5.0V$, $A_V = 0dB$, Loudspeaker $R_L = 15\mu H + 8\Omega + 15\mu H$, Earpiece $R_L = 8\Omega$, f = 1kHz, unless otherwise specified. Limits apply for $T_A = 25C$. LS = Loudspeaker, HP = Headphone, EP = Earpiece.

Symbol	Parameter	Conditions	LM4		Units
Symbol	Parameter	Conditions	Typical ⁽³⁾	Limits ⁽²⁾	(Limits)
		V _{IN} = 0, No Load		P	
		LS mode 1	4.5		mA
		HP mode 8	4.9		mA
I _{DD}	Supply Current	EP Bypass Mode	0.9		mA
		LS + HP Mode 5 and Mode 10	7.7		mA
		LS Mode 1, GAMP_SD = 1	3.7		mA
		HP Mode 8, GAMP_SD = 1	4.4		mA
I _{SD}	Shutdown Current		0.02	1	μA (max)
		$V_{IN} = 0V, LS, R_L = 8\Omega$ LS Gain = 6dB, Stereo Mode 10	9	40	mV (max
V _{OS}	Output Offset Voltage	$V_{IN} = 0V$, HP, $R_L = 32\Omega$ Ground Reference, Stereo Mode 10	1	5	mV (max
		V_{IN} = 0V, EP Bypass only, R_{L} = 8 Ω	1	5	mV (max
	Output Power	LS Mode 1, THD+N = 1%, f = 1kHz LS Gain = 6dB, $R_L = 4\Omega$	2.1		W
		LS Mode 1, THD+N = 1%, f = 1kHz LS Gain = 6dB, $R_L = 8\Omega$	1.25		W
Po		HP Mode 8, THD+N = 1%, f = 1kHz HP Attenuation = 0dB, R_L = 16 Ω	42		mW
		HP Mode 8, THD+N = 1%, f = 1kHz HP Attenuation = 0dB, R_L = 32 Ω	42		mW
		EP Bypass Only, THD+N = 1% f = 1kHz, $R_L = 8\Omega$	135		mW
		LS Mode 1, f = 1kHz P _{OUT} = 600mW; R _L = 8 Ω	0.015		%
THD+N	Total Harmonic Distortion + Noise	HP Mode 8, f = 1kHz P _{OUT} = 20mW; R _L = 32Ω	0.01		%
		EP Bypass only, f = 1kHz, P _{OUT} = 60mW; R _L = 8 Ω	0.08		%
η	Efficiency	LS Output	88		%
		A-weighted, inputs terminated to AC GND, Output refe	erred		
		EP Bypass	10		μV
		LS; Mode 1	40		μV
⊂out	Output Noise	LS; Mode 2	47		μV
20.		LS; Mode 3	48		μV
		HP; Mode 4	9		μV
		HP; Mode 8	10		μV
		HP; Mode 12	11		μV

- (2) Datasheet min/max specification limits are guaranteed by test or statistical analysis.
- (3) Typical values represent most likely parametric norms at $T_A = +25^{\circ}C$, and at the *Recommended Operation Conditions* at the time of product characterization and are not guaranteed.
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⁽¹⁾ The Electrical Characteristics tables list guaranteed specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not guaranteed.



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Electrical Characteristics 5.0V ^{(1) (2)} (continued)

The following specifications apply for $V_{DD} = LSV_{DD} = 5.0V$, $A_V = 0dB$, Loudspeaker $R_L = 15\mu H + 8\Omega + 15\mu H$, Earpiece $R_L = 8\Omega$, f = 1kHz, unless otherwise specified. Limits apply for $T_A = 25C$. LS = Loudspeaker, HP = Headphone, EP = Earpiece.

Symbol	Desemptor	Conditions	LM4	9150	Units
Symbol	Parameter	Conditions	Typical ⁽³⁾	Limits (2)	(Limits)
		$V_{RIPPLE} = 200 \text{mV}_{PP}$; f = 217Hz, R _L = 8 Ω , C All audio inputs terminated to AC GND; ou	$C_{\rm B} = 2.2 \mu F$, htput referred		
		EP Bypass	97		dB
		Loudspeaker Output; LS Gain = 6dB			
		LS; Mode 1	75		dB
PSRR	Power Supply Rejection Ratio	LS; Mode 2	71		dB
		LS; Mode 3	71		dB
		Headphone Output, HP Attenuation = 0dB			
		HP; Mode 4	91		dB
		HP; Mode 8	80		dB
		HP; Mode 12	79		dB
	Volume Control Step Size Error		±0.2		dB
		Maximum Attenuation	-92		dB
	Digital Volume Control Range	Volume Step 2	-46.5	-49 -44	dB (min) dB (max)
		Maximum Gain	18	17 19	dB (min) dB (max)
٨		HP	98		dB
۹ _M	Mute Attenuation	LS	98		dB (max) dB (min) dB (max) dB dB dB kΩ
	Mono Channel Input Impedance	Maximum gain setting	12		kΩ
Z _{IN}	L _{IN} and R _{IN} Input Impedance	Maximum attenuation setting	111		kΩ
-IN	EP Bypass Resistance		62	50 80	kΩ (min) kΩ (max)
		f = 217Hz, V_{CM} = $1V_{P\text{-}P}$, R_{L} = 8Ω EP Bypass	55		dB
CMRR	Common-Mode Rejection Ratio	f = 217Hz, V_{CM} = 1 V_{P-P} , R_L = 8 Ω LS, Mode 1	55		dB
		f = 217Hz, V_{CM} = $1V_{P-P}$, R_L = 32Ω HP, Mode 4	61		dB
X _{TALK}	Crosstalk	HP mode 8; $P_0 = 12mW$ $R_L = 32\Omega$, f = 1kHz	78		dB
г	Turn-On Time	C _B = 2.2µF, HP, Normal Turn-On Mode	27		ms
T _{ON}		C _B = 2.2µF, HP, Fast Turn-On Mode	15		ms

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I²C micro ⁽¹⁾

The following specifications apply for V_{DD} = 5.0V and 3.3V, T_A = 25°C, 2.2V ≤ $I^2C_V_{DD}$ ≤ 5.5V, unless otherwise specified.

Symbol			1	.M49150	Units (Limits) µs (min) ns (min) ns (min)
	Parameter	Conditions	Typical	Limits (3) (4)	
t ₁	I ² C Clock Period			2.5	µs (min)
t ₂	I ² C Data Setup Time			100	ns (min)
t ₃	I ² C Data Stable Time			0	ns (min)
t ₄	Start Condition Time			100	ns (min)
t ₅	Stop Condition Time			100	ns (min)
t ₆	I ² C Data Hold Time			100	ns (min)
V _{IH}	I ² C Input Voltage High			0.7xl ² CV _{DD}	V (min)
VIL	I ² C Input Voltage Low			0.3xl ² CV _{DD}	V (max)

(1) The Electrical Characteristics tables list guaranteed specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not guaranteed.

(2) Human body model, applicable std. JESD22-A114C.

(3) Datasheet min/max specification limits are guaranteed by test or statistical analysis.

(4) Machine model, applicable std. JESD22-A115-A.

I²C micro⁽¹⁾

The following specifications apply for V_{DD} = 5.0V and 3.3V, T_A = 25°C, 1.7V ≤ $I^2C_V_{DD}$ ≤ 2.2V, unless otherwise specified.

			LM49150		Unito
Symbol	Parameter	Conditions	Typical	Limits (3)	Units (Limits) μs (min) ns (min) ns (min)
t ₁	I ² C Clock Period			2.5	µs (min)
t ₂	I ² C Data Setup Time			250	ns (min)
t ₃	I ² C Data Stable Time			0	ns (min)
t ₄	Start Condition Time			250	ns (min)
t ₅	Stop Condition Time			250	ns (min)
t ₆	I ² C Data Hold Time			250	ns (min)
VIH	I ² C Input Voltage High			0.7xl ² CV _{DD}	V (min)
V _{IL}	I ² C Input Voltage Low			0.3xl ² CV _{DD}	V (max)

(1) The Electrical Characteristics tables list guaranteed specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not guaranteed.

(2) Typical values represent most likely parametric norms at $T_A = +25^{\circ}C$, and at the *Recommended Operation Conditions* at the time of product characterization and are not guaranteed.

(3) Datasheet min/max specification limits are guaranteed by test or statistical analysis.







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THD+N

vs

Output Power

 $R_1 = 8\Omega$, f = 1kHz

V_{DD} = 3.6V

√_{DD} = 3.3√

100

100

10m

OUTPUT POWER (W)

CMRR

vs

Frequency

Headphone Mode 4

1k

FREQUENCY (Hz)

Crosstalk vs

Frequency

Headphone Mode 8

10k

100k

100m



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10k

1k

FREQUENCY (Hz)

100k

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Application Information

I²C COMPATIBLE INTERFACE

The LM49150 is controlled through an I²C compatible serial interface that consists of a serial data line (SDA) and a serial clock (SCL). The clock line is uni-directional. The data line is bi-directional (open drain). The LM49150 and the master can communicate at clock rates up to 400kHz. Figure 4 shows the I²C interface timing diagram. Data on the SDA line must be stable during the HIGH period of SCL. The LM49150 is a transmit/receive slave-only device, reliant upon the master to generate the SCL signal. Each transmission sequence is framed by a START condition and a STOP condition (Figure 5). Each data word, device address and data, transmitted over the bus is 8 bits long and is always followed by an acknowledge pulse (Figure 6). The LM49150 device address is 1111000.

I²C INTERFACE POWER SUPPLY PIN (I²CV_{DD})

The LM49150's I²C interface is powered up through the I²CV_{DD} pin. The LM49150's I²C interface operates at a voltage level set by the I²CV_{DD} pin which can be set independent to that of the main power supply pin V_{DD}. This is ideal whenever logic levels for the I²C interface are dictated by a microcontroller or microprocessor that is operating at a lower supply voltage than the main battery of a portable system.

I²C BUS FORMAT

The I²C bus format is shown in Figure 6. The START signal, the transition of SDA from HIGH to LOW while SCL is HIGH, is generated, alerting all devices on the bus that a device address is being written to the bus.

The 7-bit device address is written to the bus, most significant bit (MSB) first, followed by the R/W bit. R/W = 0 indicates the master is writing to the slave device, R/W = 1 indicates the master wants to read data from the slave device. Set R/W = 0; the LM49150 is a WRITE-ONLY device and will not respond to the R/W = 1. The data is latched in on the rising edge of the clock. Each address bit must be stable while SCL is HIGH. After the last address bit is transmitted, the master device releases SDA, during which time, an acknowledge clock pulse is generated by the slave device. If the LM49150 receives the correct address, the device pulls the SDA line low, generating an acknowledge bit (ACK).

Once the master device registers the ACK bit, the 8-bit register data word is sent. Each data bit should be stable while SCL is HIGH. After the 8-bit register data word is sent, the LM49150 sends another ACK bit. Following the acknowledgement of the register data word, the master issues a STOP bit, allowing SDA to go high while SCL is high.



Figure 4. I²C Timing Diagram



Figure 5. Start and Stop Diagram

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Figure 6. Start and Stop Diagram

Table 3. Chip Address

	B7	B6	B5	B4	B3	B2	B1	B0 (R/ W)
Chip Address	1	1	1	1	1	0	0	0

Table 4. Control Registers

	B7	B 6	B5	B4	B3	B2	B1	B0
Shutdown Control	0	0	Spread Spectrum	GAMP_SD	0	I ² CV _{DD} SD	Turn_On _Time	PWR_On
Output Mode Control	0	1	EP Bypass	HPR_SD	MC3 (HP L&R)	MC2 (HP Mono)	MC1 (LS L&R)	MC0 (LS Mono)
Output Gain Control	1	0	0	INPUT_MUTE	LS_GAIN	HP_GAIN2	HP_GAIN1	HP_GAIN0
Mono Input Volume Control	1	0	1	MG4	MG3	MG2	MG1	MG0
Left Input Volume Control	1	1	0	LG4	LG3	LG2	LG1	LG0
Right Input Volume Control	1	1	1	RG4	RG3	RG2	RG1	RG0

Table 5. Shutdown Control Register

Bit	Name	Value	Description
Dr		0	Spread Spectrum Disabled
B5	Spread Spectrum	1	Spread Spectrum Enabled
		0	Normal Operation
B4	GAMP_SD	1	Disables the gain amplifiers that are not in use, to minimize I_{DD} . Recommended for Output Modes 1, 2, 4, 5, 8, 10
B3		0	
B2	I ² CV _{DD} SD	0	$\rm I^2CV_{DD}$ acts as an active low RESET input. If $\rm I^2CV_{DD}$ drops below 1.1V, the device resets and the $\rm I^2C$ registers are restored to their default state.
		1	Normal Operation. $I^2 CV_{DD}$ voltage does not reset the device.
B1	Turn On Time	0	Normal Turn-On Time (27ms)
ы	Turn_On_Time	1	Fast Turn-On Time (15ms)
B0	PWR On	0	Device Disabled
ВО	FVVK_OD	1	Device Enabled

Table 6. Output Mode Control Register

Bit	Name	Value	Description
			Normal Output Mode Operation
B5	EP Bypass	1	Speaker and Headphone amplifier goes into shutdown mode and enables Receiver Bypass path
D4		0	Normal Operation
B4	HPR_SD	1	Disables Right Headphone Output

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	Table 7. Output Mode Selection (See ⁽¹⁾)						
Output Mode Number	MC3	MC2	MC1	MC0	LS Output	HP R Output	HP L Output
0	0	0	0	0	SD	SD	SD
1	0	0	0	1	G _P x M	SD	SD
2	0	0	1	0	2 x (G _L x L + G _R x R)	SD	SD
3	0	0	1	1	2 x (G _L x L + G _R x R) + G _P x M	SD	SD
4	0	1	0	0	SD	G _Р х М/2	G _P x M/2
5	0	1	0	1	G _P x M	G _P x M/2	G _P x M/2
6	0	1	1	0	2 x (G _L x L + G _R x R)	G _P x M/2	G _P x M/2
7	0	1	1	1	2 x (G _L x L + G _R x R) + G _P x M	G _P x M/2	G _P x M/2
8	1	0	0	0	SD	G _R x R	G _L x L
9	1	0	0	1	G _P x M	G _R x R	G _L x L
10	1	0	1	0	2 x (G _L x L + G _R x R)	G _R x R	G _L x L
11	1	0	1	1	2 x (G _L x L + G _R x R) + G _P x M	G _R x R	G _L x L
12	1	1	0	0	SD	G _R x R + G _P x M/2	G _L x L + G _P x M/2
13	1	1	0	1	G _P x M	G _R x R + G _P x M/2	G _L x L + G _P x M/2
14	1	1	1	0	2 x (G _L x L + G _R x R)	G _R x R + G _P x M/2	G _L x L + G _P x M/2
15	1	1	1	1	2 x (G _L x L + G _R x R) + G _P x M	G _R x R + G _P x M/2	G _L x L + G _P x M/2

(1) MC3: HP Select L and R In MC2: HP Select Mono In MC1: Loud Speaker Select L and R In MC0: Loud Speaker Select Mono In M : Phone In (Mono) R: Right In L: Left In SD: Shutdown G_P: Phone In (Mono) Volume Control Gain G_R : Right Stereo Volume Control Gain G_L : Left Stereo Volume Control Gain

MC1	MCO	LSOUT
0	0	SD
0	1	М
1	0	L+R
1	1	M+L+R

MC3	MC2	HPR Output	HPL Output
0	0	SD	SD
0	1	М	Μ
1	0	L	R
1	1	M+L	M+R

Table 8. Output Gain Control (Loudspeaker)

Bit	Value	Gain (dB)
	0	+6
LS_GAIN	1	+12

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	Table 9. Headphone Output Gain Setting					
HP_Gain2	HP_Gain1	HP_Gain0	Gain (dB)			
0	0	0	0			
0	0	1	-1.2			
0	1	0	-2.5			
0	1	1	-4.0			
1	0	0	-6.0			
1	0	1	-8.5			
1	1	0	-12			
1	1	1	-18			

Table 10. Volume Control Table

Volume Step	xG4 ⁽¹⁾	xG3	xG2	xG1	xG0	Gain (dB) ⁽²⁾
1	0	0	0	0	0	-80.00
2	0	0	0	0	1	-46.50
3	0	0	0	1	0	-40.50
4	0	0	0	1	1	-34.50
5	0	0	1	0	0	-30.00
6	0	0	1	0	1	-27.00
7	0	0	1	1	0	-24.00
8	0	0	1	1	1	-21.00
9	0	1	0	0	0	-18.00
10	0	1	0	0	1	-15.00
11	0	1	0	1	0	-13.50
12	0	1	0	1	1	-12.00
13	0	1	1	0	0	-10.50
14	0	1	1	0	1	-9.00
15	0	1	1	1	0	-7.50
16	0	1	1	1	1	-6.00
17	1	0	0	0	0	-4.50
18	1	0	0	0	1	-3.00
19	1	0	0	1	0	-1.50
20	1	0	0	1	1	0.00
21	1	0	1	0	0	1.50
22	1	0	1	0	1	3.00
23	1	0	1	1	0	4.50
24	1	0	1	1	1	6.00
25	1	1	0	0	0	7.50
26	1	1	0	0	1	9.00
27	1	1	0	1	0	10.50
28	1	1	0	1	1	12.00
29	1	1	1	0	0	13.50
30	1	1	1	0	1	15.00
31	1	1	1	1	0	16.50
32	1	1	1	1	1	18.00

(1) x = M, L and R

(2) Gain / Attenuation is from input to output

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SHUTDOWN FUNCTION

The LM49150 features the following shutdown controls.

Bit B4 (GAMP_SD) of the SHUTDOWN CONTROL register controls the gain amplifiers. When GAMP_SD = 1, it disables the gain amplifiers that are not in use. For example, in Modes 1, 4 and 5, the Mono inputs are in use, so the Left and Right input gain amplifiers are disabled, causing the I_{DD} to be minimized.

Bit B0 (PWR_On) of the SHUTDOWN CONTROL register is the global shutdown control for the entire device. Set PWR_On = 0 for normal operation. PWR_On = 1 overrides any other shutdown control bit.

OUTPUT MODE CONTROL

In the LM49150 OUTPUT MODE CONTROL register (Table 6), Bit B5 (EP Bypass) controls the operation of the Earpiece Bypass path. If EP Bypass = 0, it would act under normal output mode operation set by bits B3, B2, B1, and B0. If EP Bypass = 1, it overrides the B3, B2, B1, and B0 Bits and enables the Receiver Bypass path, a class AB amplifier, to the speaker output.

Bit B4 (HPR_SD) of the OUPUT MODE CONTROL register controls the right headphone shutdown. If HPR_SD = 1, the right headphone output is disabled.

DIFFERENTIAL AMPLIFIER EXPLANATION

The LM49150 features a differential input stage, which offers improved noise rejection compared to a singleended input amplifier. Because a differential input amplifier amplifies the difference between the two input signals, any component common to both signals is cancelled. An additional benefit of the differential input structure is the possible elimination of the DC input blocking capacitors. Since the DC component is common to both inputs, and thus cancelled by the amplifier, the LM49150 can be used without input coupling capacitors when configured with a differential input signal.

SINGLE-ENDED INPUT CONFIGURATION

The left and right stereo inputs of the LM49150 are configured for single-ended sources (see Figure 1).

INPUT CAPACITOR SELECTION

Input capacitors may be required for some applications, or when the audio source is single-ended. Input capacitors block the DC component of the audio signal, eliminating any conflict between the DC component of the audio source and the bias voltage of the LM49150. The input capacitors create a high-pass filter with the input resistors R_{IN} . The -3dB point of the high-pass filter is found using Equation 1 below.

$$f = 1 / 2\pi R_{IN}C_{IN} \quad (Hz)$$

(1)

Where the value of R_{IN} is given in the Electrical Characteristics Table.

High-pass filtering the audio signal helps protect the speakers. When the LM49150 is using a single-ended source, power supply noise on the ground is seen as an input signal. Setting the high-pass filter point above the power supply noise frequencies, 217Hz in a GSM phone, for example, filters out the noise such that it is not amplified and heard on the output. Capacitors with a tolerance of 10% or better are recommended for impedance matching and improved CMRR and PSRR.

INPUT MIXER/MULTIPLEXER

The LM49150 includes a comprehensive mixer multiplexer controlled through the I²C interface. The mixer/multiplexer allows any input combination to appear on any output of LM49150. Multiple input paths can be selected simultaneously. Under these conditions, the selected inputs are mixed together and output on the selected channel. Table 7 shows how the input signals are mixed together for each possible input selection.

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CLASS D AMPLIFIER

The LM49150 features a high-efficiency, filterless, class D amplifier, which features a filterless modulation scheme. When there is no input signal applied, the output switches between V_{DD} and GND at a 50% duty cycle. Since the outputs of the LM49150 class D amplifier are differential and in phase, the result is zero net voltage across the speaker and no load current during the ideal state, thus conserving power. The switching frequency of each output is 300kHz.

When an input signal is applied, the duty cycle (pulse width) changes. For increasing output voltages, the duty cycle of one output increases while the duty cycle of the output decreases. For decreasing output voltages, the converse occurs. The difference between the two pulse widths yields the differential output voltage across the load.

SPREAD SPECTRUM

The LM49150 features a filterless spread spectrum modulation scheme. The switching frequency varies by +/-30% about a 300kHz center frequency, reducing the wideband spectral content, reducing EMI emissions radiated by the speaker and associated cables and traces. When a fixed frequency class D exhibits large amounts of spectral energy at multiples of switching frequency, the spread spectrum architecture of the LM49150 spreads that energy over a larger bandwidth. The cycle-to-cycle variation of the switching period does not affect the audio reproduction, efficiency, or PSRR. To enable spread spectrum, set the spread spectrum bit, B5 = 1 of the SHUTDOWN CONTROL register (see Table 5).

ENHANCED EMMISIONS SUPPRESSION (E²S)

The LM49150 features National's patented E²S system that reduces EMI, while maintaining high quality audio reproduction and efficiency. The LM49150 features Edge Rate Control (ERC) that greatly reduces the high frequency components of the output square waves by controlling the output rise and fall times, slowing the transitions to reduce RF emissions, while optimizing THD+N and efficiency performance.

LDO GENERAL INFORMATION

The LM49150 has different supplies for each portion of the device, allowing for the optimum combination of headroom, power dissipation and noise immunity. The speaker amplifiers are powered from LSV_{DD}. The ground reference headphone amplifiers are powered from the internal LDO. The separate power supplies allow the loudspeaker amplifier to operate from a higher voltage for maximum headroom, while the headphone amplifiers operate from a lower voltage, improving power dissipation.

GROUND REFERENCED HEADPHONE AMPLIFIER

The LM49150 features a low noise inverting charge pump that generates an internal negative supply voltage. This allows the headphone outputs to be biased about GND instead of a nominal DC voltage, like traditional headphone amplifiers. Because there is no DC component, the large DC blocking capacitors (typically 220µF) are not necessary. The coupling capacitors are replaced by two small ceramic charge pump capacitors, saving board space and cost. Eliminating the output coupling capacitors also improves low frequency response. In traditional headphone amplifiers, the headphone impedance and the output capacitor from a high-pass filter that not only blocks the DC component of the output, but also attenuates low frequencies, impacting the bass response. Because the LM49150 does not require the output coupling capacitors, the low frequency response of the device is not degraded by external components. In addition to eliminating the output coupling capacitors, the ground referenced output nearly doubles the available dynamic range of the LM49250 headphone amplifiers when compared to a traditional headphone amplifier operating from the same supply voltage.

CHARGE PUMP CAPACITOR SELECTION

Use low ESR ceramic capacitors (less than $100m\Omega$) for optimum performance.

CHARGE PUMP FLYING CAPACITOR (C1)

The flying capacitor (C1), see Figure 1, affects the load regulation and output impedance of the charge pump. A C1 value that is too low results in a loss of current drive, leading to a loss of amplifier headroom. A higher valued C1 improves load regulation and lowers charge pump output impedance to an extent. Above 2.2μ F, the R_{DS(ON)} of the charge pump switches and the ESR of C1 and Cs5 dominate the output impedance. A lower value capacitor can be used in systems with low maximum output power requirements.



CHARGE PUMP HOLD CAPACITOR (Cs5)

The value and ESR of the hold capacitor (Cs5) directly affects the ripple on CPV_{SS} . Increasing the value of Cs5 reduces output ripple. Decreasing the ESR of Cs5 reduces both output ripple and charge pump output impedance. A lower value capacitor can be used in systems with low maximum output power requirements.

LM49150 Demoboard Bill Of Materials

Location	Qty	Description	Part Number	Manufacturer
CIN2, CIN1	2	0.22uF, 1206, 10V, X7R Ceramic Capacitor	GRM319R71C224KA01D	Murata
CS4, CS2	2	0.1uF, 0805, 10V, X7R Ceramic Capacitor	GRM219R71C104KA01D	Murata
CS7	1	1.0uF, 0805, 10V, X7R Ceramic Capacitor	GRM21BR71A105KA01L	Murata
CIN3, CIN4	2	1.0uF 1206, 10V, X7R Ceramic Capacitor	GRM319R71C105KAA3D	Murata
CS5, C1	2	2.2uF, 0603, 10V, X7R, Ceramic Capacitor	GRM188R71A225KE15D	Murata
CS1, CS3, CS6	3	2.2uF, Size A, Tantalum Capactior	293D225X9010A2TE3	Vishay
U2	1	LM49150, 16 bump uSMD	LM49510	NSC
R1, R2	2	5K ohm 1/10W 0.05% 0603 SMD	CRCW06035R1KJNEA	Vishay
J11, J12, J13, J14	4	3–Header		
J1, J2, J3, J7, J8, J9, J10	7	2-Header		
J6	1	Header_3M 8516-4500PL		
U1	1	Headphone Jack		

Table 11. Bill Of Materials

Layout Guidelines

Minimize trace impedance of the power, ground and all output traces for optimum performance. Voltage loss due to trace resistance between the LM49150 and the load results in decreased output power and efficiency. Trace resistance between the power supply and the GND of the LM49150 has the same effect as a poorly regulated supply, increased ripple and reduced peak output power. Use wide traces, for power-supply inputs and amplifier outputs to minimize losses due to trace resistance, as well as providing heat dissipation from the device. Proper grounding improves audio performance, minimizes crosstalk between channels and prevents switching noise from interfering with audio signal. Use of power and ground planes is recommended.

The following recommendations should be considered when laying out the different grounds of the LM49150. Refer to the Demo Board Schematic for the corresponding component designators. Bypass capacitors for AV_{DD} (CS7), LSV_{DD} (CS1, CS2), $V_{O(LDO)}$ (CS3, CS4) should be grounded to the GND pin via a ground plane. Bypass capacitor for CPV_{SS}(CS5) should be grounded via a wide trace or a ground plane to the CPGND pin. The headphone grounds should be connected to the GND via a separate trace also. This will help prevent noise from the charge pump from feeding into the power supplies and the output.

Place all digital components and digital signal traces as far as possible from analog components and traces. Do not run digital and analog traces in parallel on the same PCB layer.



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Demo Board Schematic





PC Board Layout



Figure 8. Top Silkscreen Layer

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Figure 9. Top Layer



Figure 10. Layer 2

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Figure 11. Layer 3



Figure 12. Bottom Layer



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Figure 13. Bottom Silkscreen

Revision History

Rev	Date	Description
1.0	08/27/08	Initial release.
1.01	09/09/08	Edited Table 6.
1.02	03/04/09	Added the Layout Guidelines section.



PACKAGING INFORMATION

Orderable Device	Status	Package Type		Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Samples
	(1)		Drawing			(2)		(3)	(Requires Login)
LM49150TL/NOPB	ACTIVE	DSBGA	YZR	20	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	
LM49150TLX/NOPB	ACTIVE	DSBGA	YZR	20	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

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⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM49150TL/NOPB	DSBGA	YZR	20	250	178.0	8.4	2.34	2.85	0.76	4.0	8.0	Q1
LM49150TLX/NOPB	DSBGA	YZR	20	3000	178.0	8.4	2.34	2.85	0.76	4.0	8.0	Q1

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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM49150TL/NOPB	DSBGA	YZR	20	250	203.0	190.0	41.0
LM49150TLX/NOPB	DSBGA	YZR	20	3000	206.0	191.0	90.0

YZR0020



B. This drawing is subject to change without notice.



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