

# LMP91051 Configurable AFE for Nondispersive Infrared (NDIR) Sensing Applications

Check for Samples: LMP91051

### **FEATURES**

- **Dual Channel Input**
- Programmable gain amplifier
- "Dark Signal" offset cancellation
- Supports external filtering
- Common mode generator and 8 bit DAC
- Package 14 pin TSSOP

## **APPLICATIONS**

- NDIR sensing •
- **Demand control ventilation**
- **Building monitoring** .
- CO2 cabin control Automotive •
- Alcohol detection Automotive .
- Industrial safety and security
- **GHG & Freons detection platforms**

### DESCRIPTION

The LMP91051 is a dual channel programmable integrated Sensor Analog Front End (AFE) optimized for thermopile sensors, as typically used in NDIR applications. It provides a complete signal path solution between a sensor and microcontroller that generates an output voltage proportional to the thermopile voltage. The LMP91051's programmability enables it to support multiple thermopile sensors with a single design as opposed to the multiple discrete solutions.

The LMP91051 features a programmable gain amplifier (PGA), "dark phase" offset cancellation, and an adjustable common mode generator (1.15V or 2.59V) which increases output dynamic range. The PGA offers a low gain range of 167V/V to 1335V/V plus a high gain range of 1002V/V to 7986V/V which enables the user to utilize thermopiles with different sensitivities. The PGA is highlighted by low gain drift (20 ppm/°C), output offset drift (230  $\mu$ V/°C at G = 1002 V/V), phase delay drift (300 ns) and noise specifications (0.1  $\mu$ V<sub>RMS</sub> 0.1 to 10Hz). The offset cancellation circuitry compensates for the "dark signal" by adding an equal and opposite offset to the input of the second stage, thus removing the original offset from the output signal. This offset cancellation circuitry allows optimized usage of the ADC full scale and relaxes ADC resolution requirements.

The LMP91051 allows extra signal filtering (high pass, low pass or band pass) through dedicated pins A0 and A1, in order to remove out of band noise. The user can program through the on board SPI interface. Available in a small form factor 14 pin TSSOP package, the LMP91051 operates from -40 to +105°C.



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# LMP91051



#### SNAS581A - MARCH 2012 - REVISED NOVEMBER 2012

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.



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Table 1. Key Specifications	

	VALUE	UNIT
Programmable gain	167 to 7986	V/V
■ Low noise (0.1 to 10 Hz)	0.1	μV <sub>RMS</sub>
■ Gain Drift	20	ppm/°C (typ)
Phase Delay Drift	300	ns (typ)
Power supply voltage range	2.7 to 5.5	V

### **Block Diagram**



Figure 1. Configurable AFE for NDIR

# **Typical Application**



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Figure 2. Typical NDIR Sensing Application Circuit

## **Connection Diagram**



### **Pin Functions**

		Pin Description	
Pin	Symbol	Туре	Description
1	IN1	Analog Input	Signal Input
2	IN2	Analog Input	Signal Input
3	CMOUT	Analog Output	Common Mode Voltage Output
4	A0	Analog Output	First Stage Output
5	A1	Analog Input	Second Stage Input
6	GND	Power	Ground
7	NC		No Connect
8	NC	_	No Connect
9	OUT	Analog Output	Signal Output, reference to the same potential as CMOUT
10	CSB	Digital Input	Chip Select, active low
11	SCLK	Digital Input	Interface Clock
12	SDIO	Digital Input / Output	Serial Data Input / Output
13	VIO	Power	Digital Input/Output Supply
14	VDD	Power	Positive Supply

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#### Absolute Maximum Ratings <sup>(1)</sup>

ESD Tolerance <sup>(2)</sup>	
Human Body Model	1000V
Charged Device Model	250V
Supply Voltage (VDD)	-0.3V to 6.0V
Digital I/O supply (VIO)	-0.3V to 6.0V
Voltage at Any Pin	- 0.3V to VDD + 0.3V
Input Current at Any Pin	5mA
Storage Temperature Range	-65°C to 150°C
Junction Temperature <sup>(3)</sup>	150°C
For soldering specifications:	
see product folder at www.national.com and	
www.national.com/ms/MS/MS-SOLDERING.pdf	

(1) "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Operating Ratings is not implied. Operating Ratings indicate conditions at which the device is functional and the device should not be operated beyond such conditions.

(2) Human Body Model, applicable std. MIL-STD-883, Method 3015.7. Machine Model, applicable std. JESD22-A115-A (ESD MM std. of JEDEC) Field- Induced Charge-Device Model, applicable std. JESD22-C101-C (ESD FICDM std. of JEDEC).

(3) The maximum power dissipation is a function of T<sub>J(MAX)</sub>, θ<sub>JA</sub>, and the ambient temperature, TA. The maximum allowable power dissipation at any ambient temperature is P<sub>DMAX</sub> = (T<sub>J(MAX)</sub> - T<sub>A</sub>)/ θ<sub>JA</sub> All numbers apply for packages soldered directly onto a PC board.

### Operating Ratings (1)

Supply Voltage	2.7V to 5.5V
Junction Temperature Range	-40°C to 105°C
Package Thermal Resitance, θ <sub>JA</sub> Package 14 pin TSSOP	140°C/W

(1) "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Operating Ratings is not implied. Operating Ratings indicate conditions at which the device is functional and the device should not be operated beyond such conditions.

(2) The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$ , and the ambient temperature, TA. The maximum allowable power dissipation at any ambient temperature is  $P_{DMAX} = (T_{J(MAX)}, -T_A)/\theta_{JA}$  All numbers apply for packages soldered directly onto a PC board.



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### Electrical Characteristics<sup>(1)</sup>

The following specifications apply for VDD = 3.3V, VIO = 3.3V, VCM = 1.15V, **Bold** values for  $T_A = -40^{\circ}$ C to +85°C unless otherwise specified. All other limits apply to  $T_A = T_J = +25^{\circ}$ C.

Symbol	Parameter	Conditions	Min (2)	Тур (3)	Max (2)	Units
Power Sup	ply			1		
VDD	Supply Voltage		2.7	3.3	5.5	V
VIO	Digital I/O supply		2.7	3.3	5.5	V
IDD	Supply Current	All analog block ON	3.1	3.6	4.2	mA
	Power Down Supply Current	All analog block OFF	45	75	121	μA
	Digital Supply Current			8		μA
Offset Can	cellation (Offset DAC)	· · · · · · · · · · · · · · · · · · ·		1		
	Resolution			256		steps
	LSB	All gains		33.8		mV
	DNL		-1		+2	LSB
	Error	Output referred offset error, all gains		±100		mV
	Offset adjust range	Output referred, all gains	0.2		VDD - 0.2	V
	DAC settling time			480		μs
Programma	ble Gain Amplifier (PGA) 1st S	tage, $R_L = 10k\Omega$ , $C_L = 15pF$		1		
IBIAS	Bias Current			50	200	pА
VINMAX _HGM	Max input signal High gain mode	Referenced to CMOUT voltage, it refers to the maximum voltage at the IN pin before clipping; It		±2		mV
VINMAX _LGM	Max input signal Low gain mode	includes dark voltage of the thermopile and signal voltage.		±12		mV
VOS	Input Offset Voltage			-165		μV
G_HGM	Gain High gain mode			250		V/V
G_LGM	Gain Low gain mode			42		V/V
GE	Gain Error	Both HGM and LGM		2.5		%
VOUT	Output Voltage Range		0.5		VDD - 0.5	V
PhDly	Phase Delay	1mV input step signal, HGM, Vout measured at Vdd/2		6		μs
TCPhDly	Phase Delay variation with Temperature	1mV input step signal, HGM, Vout measured at Vdd/2,		416		ns
SSBW	Small Signal Bandwidth	Vin = 1mVpp, Gain = 250 V/V		18		kHz
Cin	Input Capacitance			100		pF
Programma	ble Gain Amplifier (PGA) 2nd S	Stage, R <sub>S</sub> = 1kΩ, C <sub>L</sub> = 1μF				
VINMAX	Max input signal	GAIN = 4 V/V		1.65		V
VINMIN	Min input signal			0.82		V
G	Gain	Programmable in 4 steps	4		32	V/V
GE	Gain Error	Any gain		2.5		%
VOUT	Output Voltage Range		0.2		VDD - 0.2	V
PhDly	Phase Delay	100mV input sine 35kHz signal, Gain = 8, VOUT measured at 1.65V, $R_L$ = 10k $\Omega$		1		μs

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T<sub>J</sub> = T<sub>A</sub>. No guarantee of parametric performance is indicated in the electrical tables under conditions of internal self-heating where T<sub>J</sub> > T<sub>A</sub>. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.

(2) Limits are 100% production tested at 25°C. Limits over the operating temperature range are guaranteed through correlations using statistical quality control (SQC) method.

(3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not guaranteed on shipped production material.



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# Electrical Characteristics<sup>(1)</sup> (continued)

The following specifications apply for VDD = 3.3V, VIO = 3.3V, VCM = 1.15V, **Bold** values for  $T_A = -40^{\circ}C$  to +85°C unless otherwise specified. All other limits apply to  $T_A = T_1 = +25^{\circ}C$ .

Symbol	Parameter	Conditions	Min (2)	Тур (3)	Max (2)	Units
TCPhDly	Phase Delay variation with Temperature	250mV input step signal, Gain = 8, Vout measured at Vdd/2		84		ns
SSBW	Small Signal Bandwidth	Gain = 32 V/V		360		kHz
Cin	Input Capacitance			5		pF
CLOAD, OUT	OUT Pin Load Capacitance	Series RC		1		μF
RLOAD, OUT	OUT Pin Load Resistance	Series RC		1		kΩ
Combined A	Amplifier Chain Specification					
en	Input-Referred Noise Density	Combination of both current and voltage noise, with a $86k\Omega$ source impedance at 5Hz, Gain = 7986		30		nV√Hz
	Input-Referred Integrated Noise	Combination of both current and voltage noise, with a $86k\Omega$ source impedance 0.1Hz to 10Hz, Gain = 7986		0.1	0.12 (4)	µVrms
		PGA1 GAIN = 42, PGA2 GAIN = 4		167		
		PGA1 GAIN = 42, PGA2 GAIN = 8		335		
		PGA1 GAIN = 42, PGA2 GAIN = 16		669		
		PGA1 GAIN = 42, PGA2 GAIN = 32		1335		
G	Gain	PGA1 GAIN = 250, PGA2 GAIN = 4		1002		- V/V - -
		PGA1 GAIN = 250, PGA2 GAIN = 8		2004		
		PGA1 GAIN = 250, PGA2 GAIN = 16		4003		
		PGA1 GAIN = 250, PGA2 GAIN = 32		7986		
GE	Gain Error	Any gain		5		%
	<b>o</b> i <b>o</b> (5)	Gain = 167 V/V, 335 V/V, 669 V/V, 1335 V/V		6		ppm/°C
TCCGE	Gain Temp Coefficient <sup>(5)</sup>	Gain = 1002 V/V, 2004 V/V, 4003 V/V, 7986V/V		20		ppm/°C
PSRR	Power Supply Rejection Ratio	DC, 3.0V to 3.6V supply, gain = 1002V/V	90	110		dB
PhDly	Phase Delay	1mV input step signal, Gain = 1002, Vout measured at Vdd/2		9		μs
TCPhDly	Phase Delay variation with Temperature <sup>(6)</sup>	1mV input step signal, Gain=1002, Vout measured at Vdd/2		300		ns
		Gain = 167 V/V		70		
		Gain = 335 V/V		100		
		Gain = 669 V/V		160		
TCVOS	Output Offset Voltage Temperature Drift <sup>(5)</sup>	Gain = 1335 V/V		290		
10005	Temperature Drift <sup>(5)</sup>	Gain = 1002 V/V		230		µV/°C
		Gain = 2004 V/V		420		-
		Gain = 4003 V/V		800		
		Gain = 7986V/V		1550		
Common M	ode Generator					
	Common Made Maltana	VDD = 3.3V		1.15		
VCM	Common Mode Voltage	VDD = 5V		2.59		- V
	VCM accuracy			2		%
CLOAD	CMOut Load Capacitance			10		nF

(4)

Guaranteed by design and characterization. Not tested on shipped production material. TCCGE and TCVOS are calculated by taking the largest slope between -40°C and 25°C linear interpolation and 25°C and 85°C linear (5) interpolation.

TCPhDly is largest change in phase delay between -40°C and 25°C measurements and 25°C and 85°C measurements. (6)



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## SPI Interface<sup>(1)</sup>

The following specifications apply for VDD = 3.3V, VIO = 3.3V, VCM = 1.15V,  $C_L = 15pF$ , **Bold** values for  $T_A = -40^{\circ}C$  to +85°C unless otherwise specified. All other limits apply to  $T_A = T_J = +25^{\circ}C$ .

Symbol	Parameter	Conditions	Min (2)	Тур (3)	Max (2)	Units
V <sub>IH</sub>	Logic Input High		0.7 × VDD			V
V <sub>IL</sub>	Logic Input Low				0.8	V
V <sub>OH</sub>	Logic Output High		2.6			V
V <sub>OL</sub>	Logic Output Low				0.4	V
IIH/IIL	Input Digital Leakage Current		–100 <b>–200</b>		100 <b>200</b>	nA

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T<sub>J</sub> = T<sub>A</sub>. No guarantee of parametric performance is indicated in the electrical tables under conditions of internal self-heating where T<sub>J</sub> > T<sub>A</sub>. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.

(2) Limits are 100% production tested at 25°C. Limits over the operating temperature range are guaranteed through correlations using statistical quality control (SQC) method.

(3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not guaranteed on shipped production material.



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### Timing Characteristics <sup>(1)</sup>

The following specifications apply for VDD = 3.3V, VIO = 3.3V, VCM = 1.15V,  $C_L = 15pF$ , **Bold** values for  $T_A = -40^{\circ}C$  to +85°C unless otherwise specified. All other limits apply to  $T_A = T_J = +25^{\circ}C$ .

Symbol	Parameter	Conditions	Min (2)	<b>Тур</b> (3)	Max (2)	Units
t <sub>WU</sub>	Wake up time			1		ms
f <sub>SCLK</sub>	Serial Clock Frequency				10	MHz
t <sub>PH</sub>	SCLK Pulse Width High		0.4/f <sub>SCLK</sub>			ns
t <sub>PL</sub>	SCLK Pulse Width Low		0.4/f <sub>SCLK</sub>			ns
t <sub>CSS</sub>	CSB Setup Time		10			ns
t <sub>CSH</sub>	CSB Hold Time		10			ns
t <sub>SU</sub>	SDI Setup Time prior to rise edge of SCLK		10			ns
t <sub>SH</sub>	SDI Hold Time prior to rise edge of SCLK		10			ns
t <sub>DOD1</sub>	SDO Disable Time after rise edge of CSB				45	ns
t <sub>DOD2</sub>	SDO Disable Time after 16 <sup>th</sup> rise edge of SCLK				45	ns
t <sub>DOE</sub>	SDO Enable Time from the fall edge of 8 <sup>th</sup> SCLK				35	ns
t <sub>DOA</sub>	SDO Access Time after the fall edge of SCLK				35	ns
t <sub>DOH</sub>	SDO hold time after the fall edge of SCLK		5			ns
t <sub>DOR</sub>	SDO Rise time			5		ns
t <sub>DOF</sub>	SDO Fall time			5		ns

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T<sub>J</sub> = T<sub>A</sub>. No guarantee of parametric performance is indicated in the electrical tables under conditions of internal self-heating where T<sub>J</sub> > T<sub>A</sub>. Absolute Maximum Ratings indicate junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.

(2) Limits are 100% production tested at 25°C. Limits over the operating temperature range are guaranteed through correlations using statistical quality control (SQC) method.

(3) Typical values represent the most likely parametric norm as determined at the time of characterization. Actual typical values may vary over time and will also depend on the application and configuration. The typical values are not tested and are not guaranteed on shipped production material.



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### **Timing Diagrams**















Figure 9. SDO rise and fall times

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### **Typical Performance Characteristics**





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### **Typical Performance Characteristics (continued)**





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## **Functional Description**

### PROGRAMMABLE GAIN AMPLIFIER

The LMP91051 offers two programmable gain modes (low/high) with four programmable gain settings each. The purpose of the gain mode is to enable thermopiles with larger dark voltage levels. All gain settings are accessible through bits GAIN1 and GAIN2[1:0]. The low gain mode has a range of 167 V/V to 1335 V/V while the high gain mode has a range of 1002 V/V to 7986 V/V. The PGA is referenced to the internally generated VCM. Input signal, referenced to this VCM voltage, should be within +/-2mV (see VINMAX\_HGM specification) in high gain mode. In the low gain mode the first stage will provide a gain of 42 V/V instead of 250 V/V, thus allowing a larger maximum input signal up to +/-12mV (VINMAX\_LGM).



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Table 2. Gain Modes

Bit Symbol	Gain
GAIN1	0: 250 (default)
GAINT	1: 42
	00: 4 (default)
	01: 8
GAIN2 [1:0]	10: 16
	11: 32

### EXTERNAL FILTER

The LMP91051 offers two different measurement modes selectable through EXT\_FILT bit. EXT\_FILT bit is present in the Device configuration register and is programmable through SPI.

Table 3. Measurement Modes

Bit Symbol Measurement Mode			
EXT_FILT	0: The signal from the thermopile is being processed by the internal PGAs, without additional external decoupling or filtering (default).		
	1: The signal from the thermopile is being processed by the first internal PGA and fed to the A0 pin. An external low pass, high pass or band pass filter can be connected through pins A0, A1.		

An external filter can be applied when EXT\_FILT = 1. A typical band pass filter is shown in the picture below. Resistor and capacitor can be connected to the CMOUT pin of the LMP91051 as shown. Discrete component values have been added for reference.



Figure 10. Typical Bandpass Filter

#### OFFSET ADJUST

Procedure of the offset adjust is to first measure the "dark signal", program the DAC to adjust, and then measure in a second cycle the residual of the dark signal for further signal manipulation within the  $\mu$ C. The signal source is expected to have an offset component (dark signal) larger than the actual signal. During the "dark phase", the time when no light is detected by the sensor, the  $\mu$ C can program LMP91051 internal DAC to compensate for a measured offset. A low output offset voltage temperature drift (TCVOS) ensures system accuracy over temperature.

#### **COMMON MODE GENERATION**

As the sensor's offset is bipolar, there is a need to supply a VCM to the sensor. This can be programmed as 1.15V or 2.59V (approximately mid rail of 3.3V or 5V supply). It is not recommended to use 2.59V VCM with 3.3V supply

#### SPI INTERFACE

An SPI interface is available in order to program the device parameters like PGA gain of two stages, enabling external filter, enabling power for PGAs, offset adjust and common mode (VCM) voltage.

#### Interface Pins

The Serial Interface consists of SDIO (Serial Data Input / Output), SCLK (Serial Interface Clock) and CSB (Chip Select Bar). The serial interface is write-only by default. Read operations are supported after enabling the SDIO mode by programming the SDIO\_MODE\_EN register. This is discussed in detail later in the document.



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### CSB

Chip Select is a active-low signal. CSB needs to be asserted throughout a transaction. That is, CSB should not pulse between the Instruction Byte and the Data Byte of a single transaction.

Note that CSB de-assertion always terminates an on-going transaction, if it is not already complete. Likewise, CSB assertion will always bring the device into a state, ready for next transaction, regardless of the termination status of a previous transaction.

CSB may be permanently tied low for a 2-wire SPI communication protocol.

### SCLK

SCLK can idle High or Low for a write transaction. However, for a READ transaction, SCLK should idle high. SCLK features a Schmitt-triggered input and although it has hysterisis, it is recommened to keep SCLK as clean as possible to prevent glitches from inadvertently spoiling the SPI frame.

#### **Communication Protocol**

Communication on the SPI normally involves Write and Read transactions. Write transaction consists of single Write Command Byte, followed by single Data byte. The following figure shows the SPI Interface Protocol for write transaction.



Figure 11. SPI Interface Protocol

For Read transactions, user first needs to write into a SDIO mode enable register for enabling the SPI read mode. Once the device is enabled for Reading, the data is driven out on the SDIO pin during the Data field of the Read Transaction. SDIO pin is designed as a bidirectional pin for this purpose. Figure 6 shows the Read transaction. The sequence of commands that need to be issued by the SPI Master to enable SPI read mode is illustrated in Figure 13.

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READ TRANSACTION



Note: Read command is issued by the SPI Master, who after issuing the c0 (LSBit of the command byte) bit should relinquish the data line (high-Z) after meeting the hold timing(10ns) and stop SCK idling high.

Figure 12. Read Transaction

Sequence of transactions for unlocking SDIO\_MODE



Bus turnaround time = half cycle

Note:

 Once the SDIO\_mode is unlocked. The user can read as many registers as long as nothing else is written to sdio\_mode\_en register to disturb the state of SDIO\_mode
The separate signals SDI and SDO are given in the figure for the sake of understanding. However, only one signal SDIO exists in the design



#### **Registers Organization**

Configuring the device is achieved using 'Write' of the designated registers in the device. All the registers are organized into individually addressable byte-long registers that have a unique address. The format of the Write/ Read instruction is as shown below.

Bit[7]	Bit[6:4] <sup>(1)</sup>	Bit[3:0]
0 : Write Instruction	Becoming to 0	Address
1 : Read Instruction	Reserved to 0	Address

(1) Specifying any value other than zero in Bit[6:4] is prohibited.

#### REGISTERS

This section describes the programmable registers and the associated programming sequence, if any, for the device. The following table shows the summary listing of all the registers that are available to the user and their power-up values.



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### Table 5. (1)

Title	Address (Hex)	Туре	Power-up/Reset Value (Hex)
Device Configuration	0x0	Read-Write (Read allowed in SDIO Mode)	0x0
DAC Configuration	0x1	Read-Write (Read allowed in SDIO Mode)	0x80
SDIO Mode Enable	0xF	Write-only	0x0

(1) Recommended values must be programmed where they are indicated in order to avoid unexpected results. Avoid writing to addresses not mentioned in the document; this could cause unexpected results.

#### Device Configuration – Device Configuration Register (Address 0x0)

Bit	Bit Symbol	Description	
7	INP_SEL	0: IN1 (default)	
		1: IN2	
		00: PGA1 OFF PGA2 OFF (default)	
[0.5]	EN	01: PGA1 OFF, PGA2 ON	
[6:5]	EIN	10: PGA1 ON, PGA2 OFF	
		11: PGA1 ON, PGA2 ON	
4		0: PGA1 to PGA2 direct (default)	
4	EXT_FILT	1: PGA1 to PGA2 via external filter	
3	CMN_MODE	0 : 1.15V (default)	
3		1 : 2.59V	
		00: 4 (default)	
[0.4]	GAIN2	01: 8	
[2:1]	GAINZ	10: 16	
		11: 32	
0	GAIN1	0: 250 (default)	
0	GAINT	1: 42	

#### DAC Configuration – DAC Configuration Register (Address 0x1)

The output DC level will shift according to the formula Vout\_shift = -33.8mV \* (NDAC - 128).

Bit	Bit Symbol	Description
[7:0]	NDAC	128 (0x80): Vout_shift = -33.8mV * (128 - 128) = 0mV (default)

#### SDIO Mode – SDIO Mode Enable Register (Address 0xf)

Write-only

Bit	Bit Symbol	Description
[7:0]		To enter SDIO Mode, write the successive sequence 0xFE and 0xED. Write anything other than this sequence to get out of mode.



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### PACKAGING INFORMATION

Orderable Device	Status	Package Type	0	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing			(2)		(3)		(4)	
LMP91051MT/NOPB	ACTIVE	TSSOP	PW	14	94	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 105	LMP910 51MT	Samples
LMP91051MTX/NOPB	ACTIVE	TSSOP	PW	14	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 105	LMP910 51MT	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND**: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> Only one of markings shown within the brackets will appear on the physical device.

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# PACKAGE MATERIALS INFORMATION

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### TAPE AND REEL INFORMATION





# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal	

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMP91051MTX/NOPB	TSSOP	PW	14	2500	330.0	12.4	6.95	8.3	1.6	8.0	12.0	Q1

TEXAS INSTRUMENTS

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# PACKAGE MATERIALS INFORMATION

5-Feb-2013



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMP91051MTX/NOPB	TSSOP	PW	14	2500	349.0	337.0	45.0

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



A. An integration of the information o

Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.

Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.

E. Falls within JEDEC MO-153



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