

# LMS485

## 5V Low Power RS-485 / RS-422 Differential Bus Transceiver

### General Description

The LMS485 is a low power differential bus/line transceiver designed for high speed bidirectional data communication on multipoint bus transmission lines. It is designed for balanced transmission lines. It meets ANSI Standards TIA/EIA RS422-B, TIA/EIA RS485-A and ITU recommendation and V.11 and X.27. The LMS485 combines a TRI-STATE™ differential line driver and differential input receiver, both of which operate from a single 5.0V power supply. The driver and receiver have an active high and active low, respectively, that can be externally connected to function as a direction control. The driver and receiver differential inputs are internally connected to form differential input/output (I/O) bus ports that are designed to offer minimum loading to bus whenever the driver is disabled or when  $V_{CC} = 0V$ . These ports feature wide positive and negative common mode voltage ranges, making the device suitable for multipoint applications in noisy environments. The LMS485 is available in a 8-Pin SOIC and 8-Pin DIP packages. It is a drop-in socket replacement to Maxim's MAX485

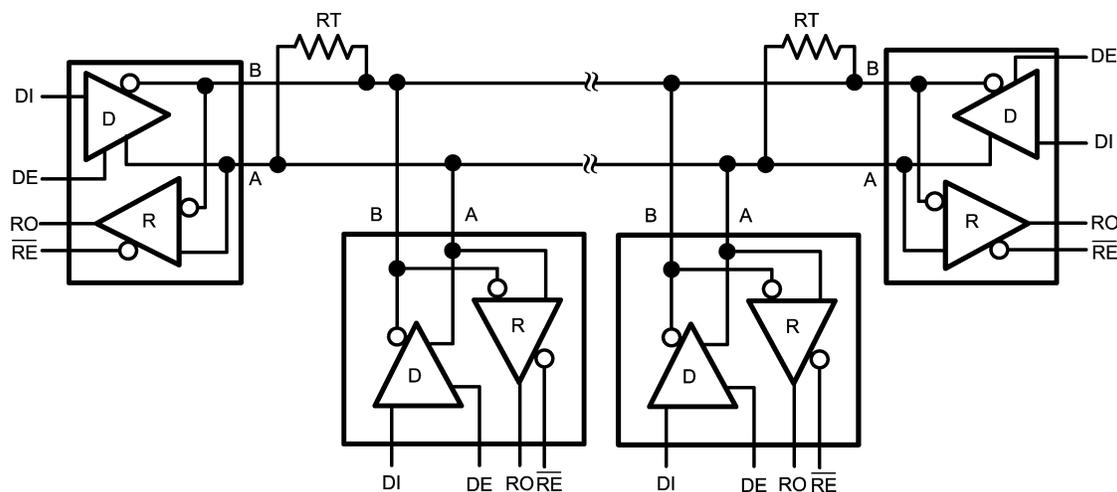
### Features

- Meet ANSI standard RS-485-A and RS-422-B
- Data rate 2.5 Mbps
- Single supply voltage operation, 5V
- Thermal shutdown protection
- Short circuit protection
- Low power BiCMOS
- Allows up to 32 transceivers on the bus
- Open circuit fail-safe for receiver
- Extended operating temperature range  $-40^{\circ}C$  to  $85^{\circ}C$
- Drop-in replacement to MAX485
- Available in 8-pin SOIC and 8-Pin DIP package

### Applications

- Low power RS-485 systems
- Network hubs, bridges, and routers
- Point of sales equipment (ATM, barcode scanners,...)
- Local area networks (LAN)
- Integrated service digital network (ISDN)
- Industrial programmable logic controllers
- High speed parallel and serial applications
- Multipoint applications with noisy environment

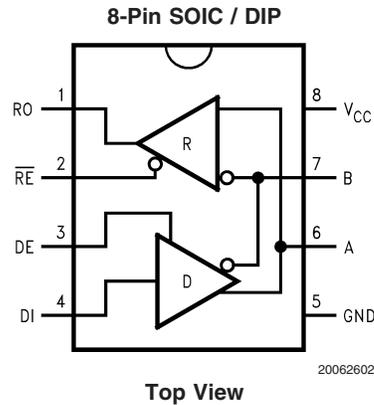
### Typical Application



A Typical multipoint application is shown in the above figure. Terminating resistors, RT, are typically required but only located at the two ends of the cable. Pull up and pull down resistors maybe required at the end of the bus to provide failsafe biasing. The biasing resistors provide a bias to the cable when all drivers are in TRI-STATE, See National Application Note, AN-847 for further information.

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## Connection Diagram



## Truth Table

DRIVER SECTION				
$\overline{RE}$	DE	DI	A	B
X	H	H	H	L
X	H	L	L	H
X	L	X	Z	Z
RECEIVER SECTION				
$\overline{RE}$	DE	A-B		RO
L	L	$\geq +0.2V$		H
L	L	$\leq -0.2V$		L
H	X	X		Z
L	L	OPEN *		H

**Note:** \* = Non Terminated, Open Input only

X = Irrelevant

Z = TRI-STATE

H = High level

L = Low level

## Pin Descriptions

Pin #	I/O	Name	Function
1	O	RO	Receiver Output: If $A > B$ by 200 mV, RO will be high; If $A < B$ by 200mV, RO will be low. RO will be high also if the inputs (A and B) are open (non-terminated)
2	I	$\overline{RE}$	Receiver Output Enable: RO is enabled when $\overline{RE}$ is low; RO is in TRI-STATE when $\overline{RE}$ is high
3	I	DE	Driver Output Enable: The driver outputs (A and B) are enabled when DE is high; they are in TRI-STATE when DE is low. Pins A and B also function as the receiver input pins (see below)
4	I	DI	Driver Input: A low on DI forces A low and B high while a high on DI forces A high and B low when the driver is enabled
5	N/A	GND	Ground
6	I/O	A	Non-inverting Driver Output and Receiver Input pin. Driver Output levels conform to RS-485 signaling levels
7	I/O	B	Inverting Driver Output and Receiver Input pin. Driver Output levels conform to RS-485 signaling levels
8	N/A	$V_{CC}$	Power Supply: $4.75V \leq V_{CC} \leq 5.25V$

## Ordering Information

Package	Part Number	Package Marking	Transport Media	NSC Drawing
8-Pin SOIC	LMS485CM	LMS485CM	95 Units/Rail	M08A
	LMS485CMX		2.5k Units Tape and Reel	
	LMS485IM	LMS485IM	95 Units/Rail	
	LMS485IMX		2.5k Units Tape and Reel	
8-Pin DIP	LMS485CNA	LMS485CNA	40 Units/Rail	N08E
	LMS485INA	LMS485INA	40 Units/Rail	

**Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage, $V_{CC}$ (Note 2)	7V
Input Voltage, $V_{IN}$ (DI, DE, or $\overline{RE}$ )	-0.3V to $V_{CC} + 0.3V$
Voltage Range at Any Bus Terminal (AB)	-7V to 12V
Receiver Outputs	-0.3V to $V_{CC} + 0.3V$
Package Thermal Impedance, $\theta_{JA}$	
SOIC	125°C/W
DIP	88°C/W
Junction Temperature (Note 3)	150°C
Operating Free-Air Temperature Range, $T_A$	
Commercial	0°C to 70°C
Industrial	-40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Soldering Information	
Infrared or Convection (20 sec.)	235°C
Lead Temperature (4 sec.)	260°C
ESD Rating (Note 4)	7kV

**Operating Ratings**

	Min	Nom	Max	
Supply Voltage, $V_{CC}$	4.75	5.0	5.25	V
Voltage at any Bus Terminal (Separately or Common Mode)	-7		12	V
$V_{IN}$ or $V_{IC}$				
High-Level Input Voltage, $V_{IH}$ (Note 5)	2			V
Low-Level Input Voltage, $V_{IL}$ (Note 5)			0.8	V
Differential Input Voltage, $V_{ID}$ (Note 6)			±12	V
High-Level Output				
Driver, $I_{OH}$			-150	mA
Receiver, $I_{OH}$			-42	mA
Low-Level Output				
Driver, $I_{OL}$			80	mA
Receiver, $I_{OL}$			26	mA

**Electrical Characteristics**

Over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
Driver Section						
$ V_{OD1} $	Differential Output Voltage	$R = \infty$ (Figure 1)			5.25	V
$ V_{OD2} $	Differential Output Voltage	$R = 50\Omega$ (Figure 1), RS-422	2.0			V
		$R = 27\Omega$ (Figure 1), RS-485	1.5		5.0	
$\Delta V_{OD}$	Change in Magnitude of Driver Differential Output Voltage for Complementary Output States	$R = 27\Omega$ or $50\Omega$ (Figure 1), (Note 7)			0.2	V
$V_{OC}$	Common-Mode Output Voltage	$R = 27\Omega$ or $50\Omega$ (Figure 1)			3.0	V
$\Delta V_{OC}$	Change in Magnitude of Driver Common-Mode Output Voltage for Complementary Output States	$R = 27\Omega$ or $50\Omega$ (Figure 1), (Note 7)			0.2	V
$V_{IH}$	CMOS Inout Logic Threshold High	DE, DI, $\overline{RE}$	2.0			V
$V_{IL}$	CMOS Input Logic Threshold Low	DE, DI, $\overline{RE}$			0.8	V
$I_{IN1}$	Logic Input Current	DE, DI, $\overline{RE}$			±2	μA
Receiver Section						
$I_{IN2}$	Input Current (A, B)	DE = 0V, $V_{CC} = 0V$ or 5.25V $V_{IN} = 12V$			1.0	mA
		$V_{IN} = -7V$			-0.8	
$V_{TH}$	Differential Input Threshold Voltage	$-7V \leq V_{CM} \leq +12V$	-0.2		+0.2	V
$\Delta V_{TH}$	Input Hysteresis Voltage ( $V_{TH+} - V_{TH-}$ )	$V_{CM} = 0$		95		mV

## Electrical Characteristics (Continued)

Over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
$V_{OH}$	CMOS High-level Output Voltage	$I_{OH} = -4\text{mA}$ , $V_{ID} = 200\text{mV}$	3.5			V
$V_{OL}$	CMOS Low-level	$I_{OL} = 4\text{mA}$ , $V_{ID} = -200\text{mV}$			0.40	V
$I_{OZR}$	Tristate Output Leakage Current	$0.4\text{V} \leq V_O \leq +2.4\text{V}$			$\pm 1$	$\mu\text{A}$
$R_{IN}$	Input Resistance	$-7\text{V} \leq V_{CM} \leq +12\text{V}$	12			$\text{k}\Omega$
Power Supply Current						
$I_{CC}$	Supply Current	$DE = V_{CC}$ , $\overline{RE} = \text{GND}$ or $V_{CC}$		320	500	$\mu\text{A}$
		$DE = 0\text{V}$ , $\overline{RE} = \text{GND}$ or $V_{CC}$		315	400	
$I_{OSD1}$	Driver Short-circuit Output Current	$V_O = \text{high}$ , $-7\text{V} \leq V_{CM} \leq +12\text{V}$ (Note 8)	35		250	$\text{mA}$
$I_{OSD2}$	Driver Short-circuit Output Current	$V_O = \text{low}$ , $-7\text{V} \leq V_{CM} \leq +12\text{V}$ (Note 8)	35		250	$\text{mA}$
$I_{OSR}$	Receiver Short-circuit Output Current	$0\text{V} \leq V_O \leq V_{CC}$	7		95	$\text{mA}$
Switching Characteristics						
Driver						
$T_{PLH}$ , $T_{PHL}$	Propagation Delay Input to Output	$R_L = 54\Omega$ , $C_L = 100\text{pF}$ (Figure 3, Figure 7)	10	35	60	$\text{nS}$
$T_{SKEW}$	Driver Output Skew	$R_L = 54\Omega$ , $C_L = 100\text{pF}$ (Figure 3, Figure 7)		5	10	$\text{nS}$
$T_R$ , $T_F$	Driver Rise and Fall Time	$R_L = 54\Omega$ , $C_L = 100\text{pF}$ (Figure 3, Figure 7)	3	8	40	$\text{nS}$
$T_{ZH}$ , $T_{ZL}$	Driver Enable to Output Valid Time	$C_L = 100\text{pF}$ , $R_L = 500\Omega$ (Figure 4, Figure 8)		25	70	$\text{nS}$
$T_{HZ}$ , $T_{LZ}$	Driver Output Disable Time	$C_L = 15\text{pF}$ , $R_L = 500\Omega$ (Figure 4, Figure 8)		30	70	$\text{nS}$
Receiver						
$T_{PLH}$ , $T_{PHL}$	Propagation Delay Input to Output	$R_L = 54\Omega$ , $C_L = 100\text{pF}$ (Figure 5, Figure 7)	20	50	200	$\text{nS}$
$T_{SKEW}$	Receiver Output Skew	$R_L = 54\Omega$ , $C_L = 100\text{pF}$ (Figure 5, Figure 7)		5		$\text{nS}$
$T_{ZH}$ , $T_{ZL}$	Receiver Enable Time	$C_L = 15\text{pF}$ , $R_L = 1\text{k}\Omega$ (Figure 6, Figure 10)		20	50	$\text{nS}$
	Receiver Disable Time			20	50	$\text{nS}$
$F_{MAX}$	Maximum Data Rate		2.5			$\text{Mbps}$

**Note 1:** Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics

**Note 2:** All voltage values, except differential I/O bus voltage, are with respect to network ground terminal.

**Note 3:** The maximum power dissipation is a function of  $T_{J(\text{MAX})}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(\text{MAX})} - T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

**Note 4:** ESD rating based upon human body model, 100pF discharged through 1.5k $\Omega$ .

**Note 5:** Voltage limits apply to DI, DE,  $\overline{RE}$  pins.

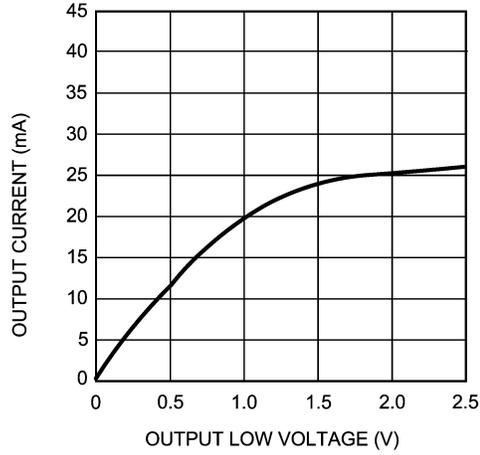
**Note 6:** Differential input/output bus voltage is measured at the non-inverting terminal A with respect to the inverting terminal B.

**Note 7:**  $|\Delta V_{OD}|$  and  $|\Delta V_{OC}|$  are changes in magnitude of  $V_{OD}$  and  $V_{OC}$ , respectively when the input changes from high to low levels.

**Note 8:** Peak current

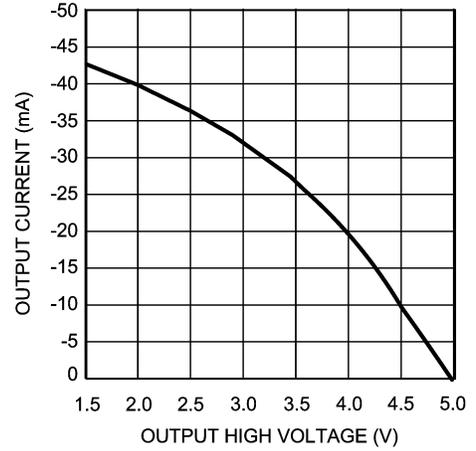
# Typical Performance Characteristics

Output Current vs. Receiver Output Low Voltage



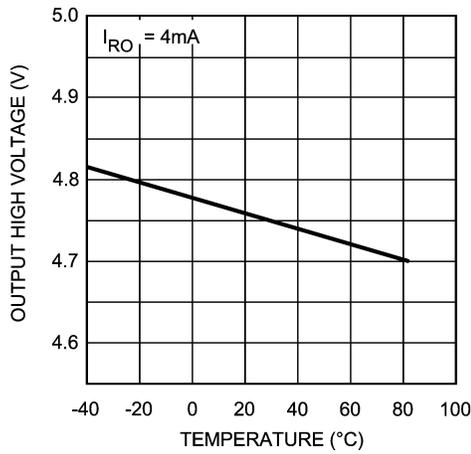
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Output Current vs. Receiver Output High Voltage



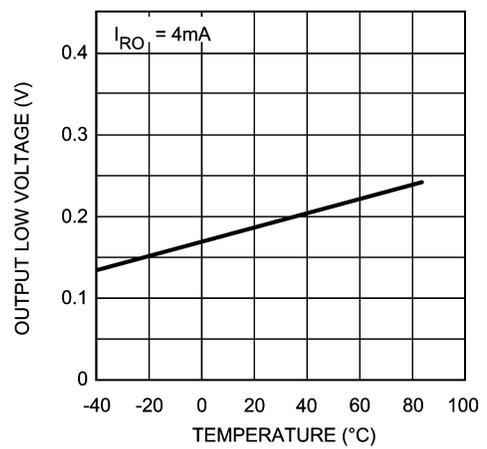
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Receiver Output High Voltage vs. Temperature



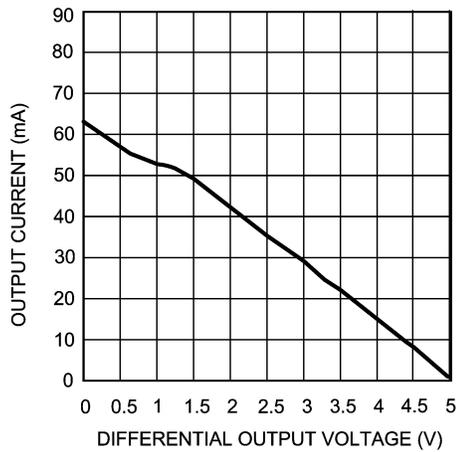
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Receiver Output Low-Voltage vs. Temperature



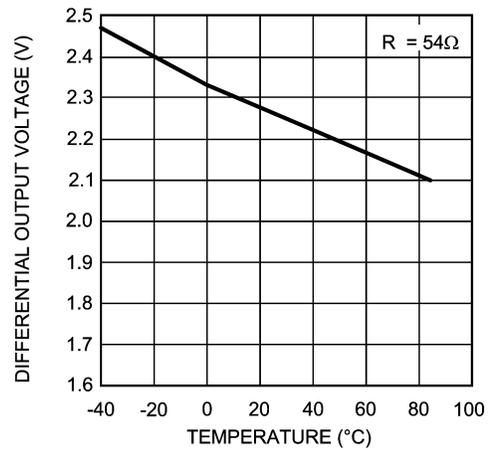
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Driver Output Current vs. Differential Output Voltage



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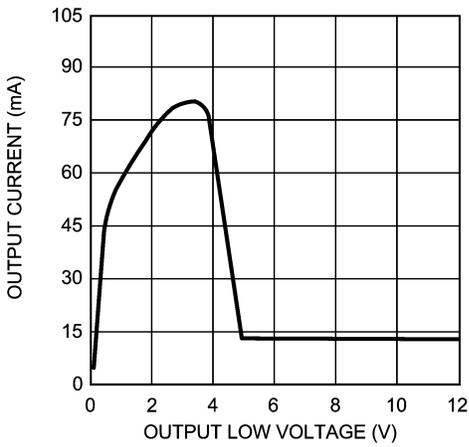
Driver Differential Output Voltage vs. Temperature



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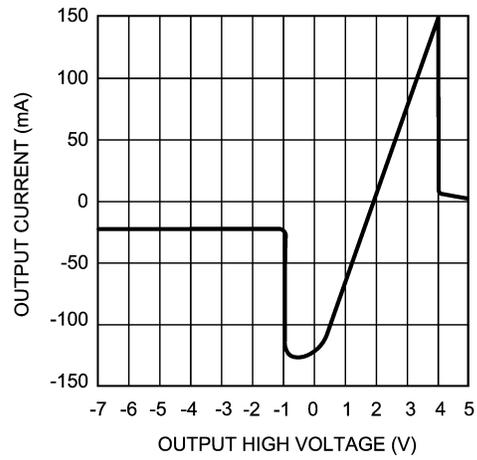
# Typical Performance Characteristics (Continued)

Output Current vs. Driver Output Low Voltage



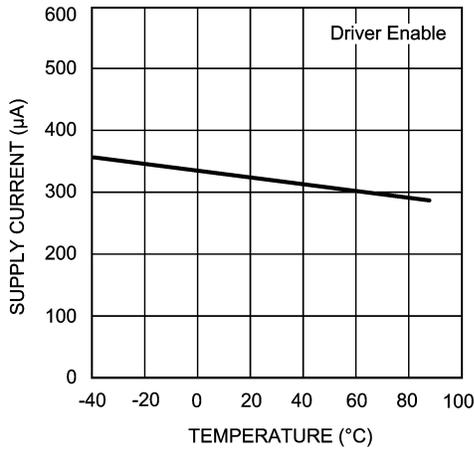
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Output Current vs. Driver Output High Voltage



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Supply Current vs. Temperature



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## Parameter Measuring Information

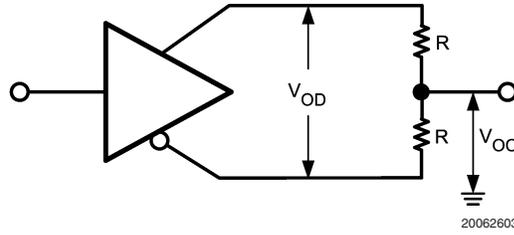


FIGURE 1. Test Circuit for  $V_{OD}$  and  $V_{OC}$

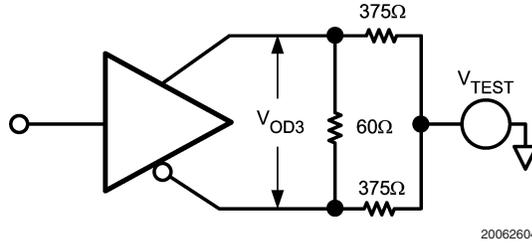


FIGURE 2. Test Circuit for  $V_{OD3}$

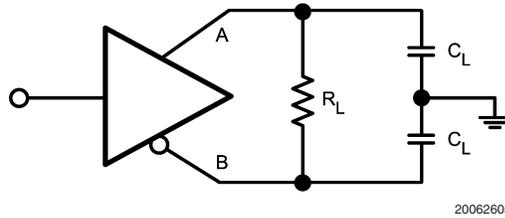


FIGURE 3. Test Circuit for Driver Propagation Delay

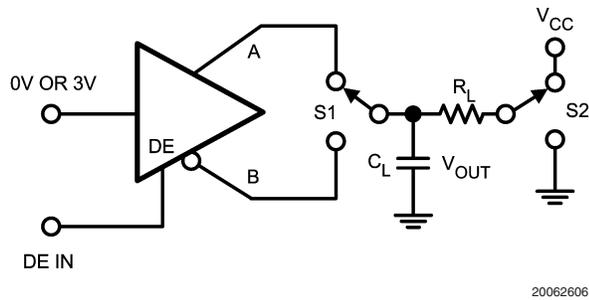
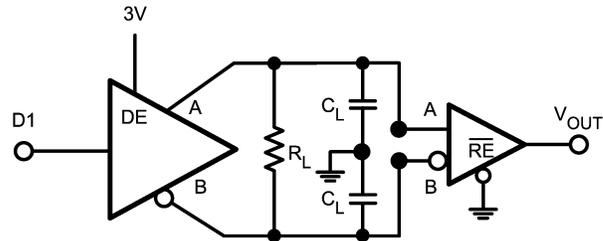


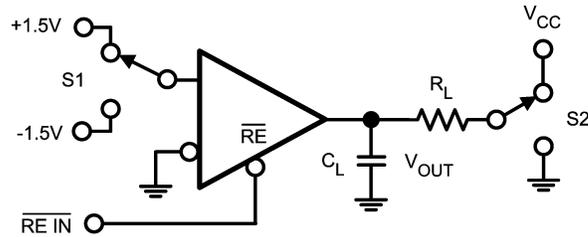
FIGURE 4. Test Circuit for Driver Enable / Disable

Parameter Measuring Information (Continued)



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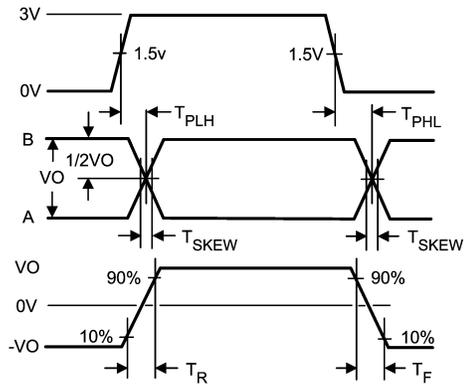
FIGURE 5. Test Circuit for Receiver Propagation Delay



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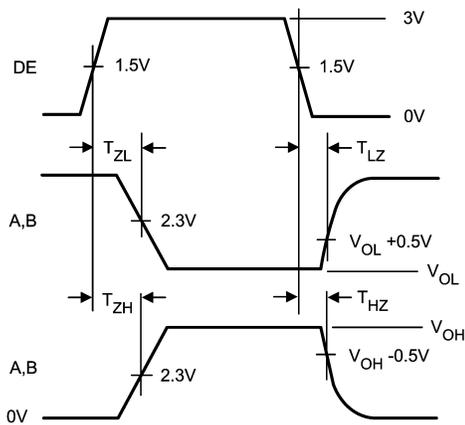
FIGURE 6. Test Circuit for Receiver Enable / Disable

# Switching Characteristics



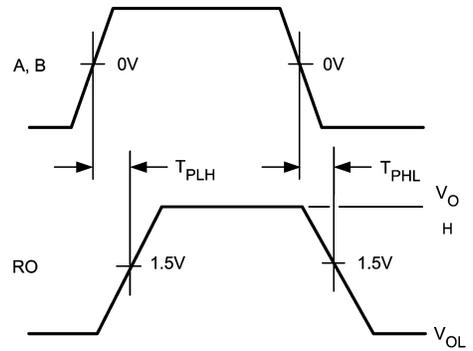
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FIGURE 7. Driver Propagation Delay, Rise / Fall Time



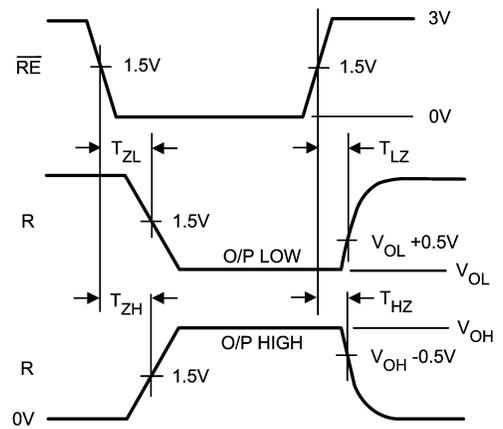
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FIGURE 8. Driver Enable / Disable Time



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FIGURE 9. Receiver Propagation Delay



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FIGURE 10. Receiver Enable / Disable Time

## Application Information

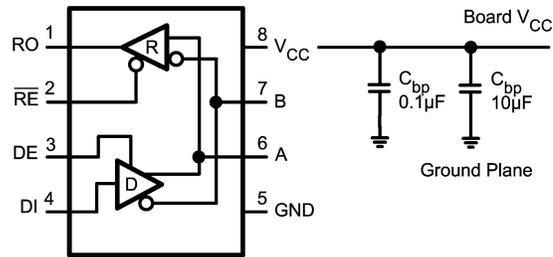
### POWER LINE NOISE FILTERING

A factor to consider in designing power and ground is noise filtering. A noise filtering circuit is designed to prevent noise generated by the integrated circuit (IC) as well as noise entering the IC from other devices. A common filtering method is to place by-pass capacitors ( $C_{bp}$ ) between the power and ground lines.

Placing a by-pass capacitor ( $C_{bp}$ ) with the correct value at the proper location solves many power supply noise problems. Choosing the correct capacitor value is based upon the desired noise filtering range. Since capacitors are not

ideal, they may act more like inductors or resistors over a specific frequency range. Thus, many times two by-pass capacitors may be used to filter a wider bandwidth of noise. It is highly recommended to place a larger capacitor, such as  $10\mu\text{F}$ , between the power supply pin and ground to filter out low frequencies and a  $0.1\mu\text{F}$  to filter out high frequencies.

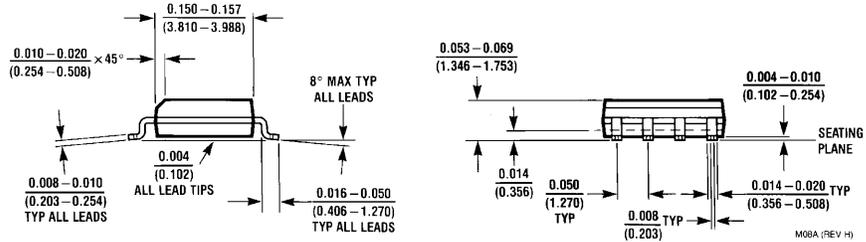
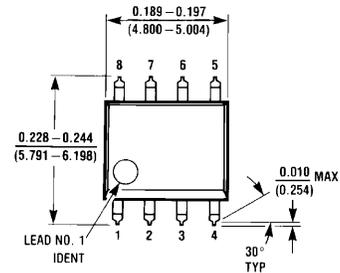
By-pass capacitors must be mounted as close as possible to the IC to be effective. Long leads produce higher impedance at higher frequencies due to stray inductance. Thus, this will reduce the by-pass capacitor's effectiveness. Surface mounted chip capacitors are the best solution because they have lower inductance.



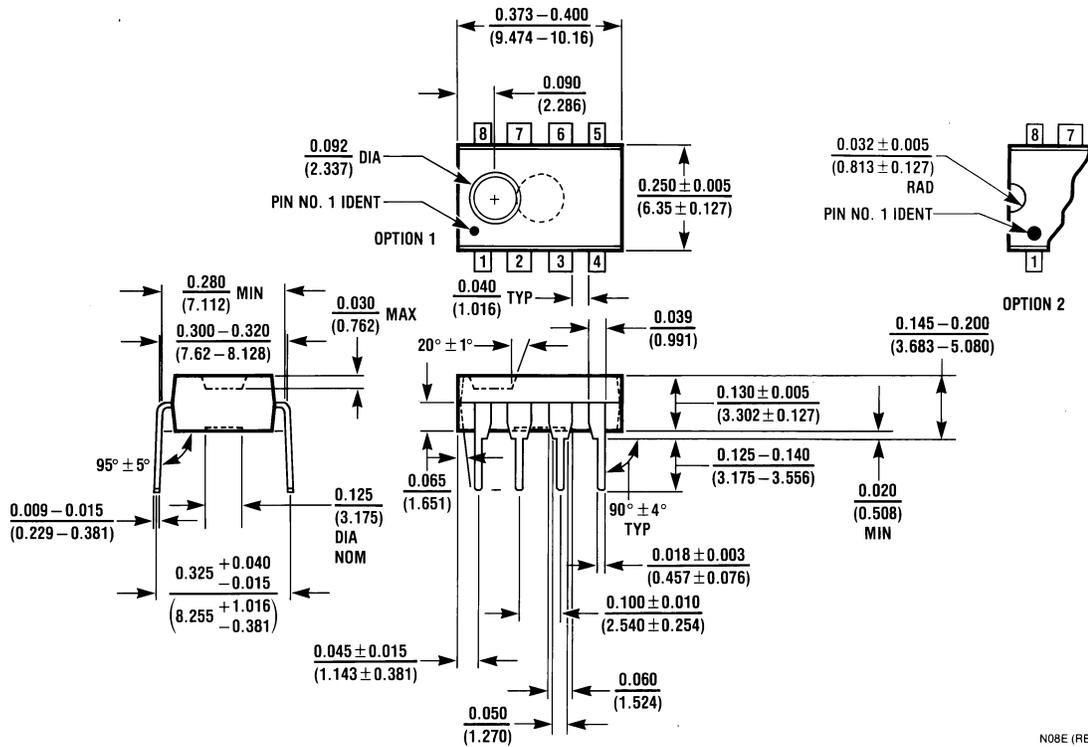
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FIGURE 11. Placement of by-pass Capacitors,  $C_{bp}$

**Physical Dimensions** inches (millimeters) unless otherwise noted



**8-Pin SOIC**  
NS Package Number M08A



**8-Pin DIP**  
NS Package Number N08E

## Notes

### LIFE SUPPORT POLICY

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