

LTC1262

## 12V, 30mA Flash Memory Programming Supply

## FEATURES

- Regulated 12V ±5% Output Voltage
- No Inductors
- Supply Voltage Range: 4.75V to 5.5V
- Guaranteed 30mA Output
- Low Power: I<sub>CC</sub> = 500µA
- I<sub>CC</sub> in Shutdown: 0.5µA
- 8-Pin PDIP or SO-8 Package

## **APPLICATIONS**

- 12V Flash Memory Programming Supplies
- Compact 12V Op Amp Supplies
- Battery-Powered Systems

## DESCRIPTION

The LTC<sup>®</sup>1262 is a regulated 12V, 30mA output DC/DC converter. It is designed to provide the 12V  $\pm$ 5% output necessary to program byte-wide flash memories. The output will provide up to 30mA from input voltages as low as 4.75V without using any inductors. Only four external capacitors are required to complete an extremely small surface mountable circuit.

The TTL compatible shutdown pin can be directly connected to a microprocessor and reduces the supply current to less than  $0.5\mu$ A. The LTC1262 offers improved shutdown current performance and requires fewer external components than competing solutions.

The LTC1262 is available in an 8-pin PDIP or SO-8 package.

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# TYPICAL APPLICATION



## SHDN 5V 5V/DIV 5V 5V/DIV 5V 0V 5V/DIV 5V 0V 1ms/DIV 1282 TAUL

### In/Out of Shutdown



## ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V <sub>CC</sub> ) 6	V
Input Voltage (SHDN) $-0.3V$ to V <sub>CC</sub> + 0.3	
Output Current (I <sub>OUT</sub> ) 50m	А
Operating Temperature Range	
LTC1262C0°C to 70°	С
LTC1262140°C to 85°	С
Storage Temperature Range65°C to 150°	С
Lead Temperature (Soldering, 10 sec) 300°	С

## PACKAGE/ORDER INFORMATION



Consult factory for Military grade parts.

# **ELECTRICAL CHARACTERISTICS**

 $V_{CC}$  = 4.75V to 5.5V,  $T_A$  = -40°C to 85°C, (Notes 2, 3), unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V <sub>OUT</sub>	Output Voltage	$\begin{array}{l} 0mA \leq I_{OUT} \leq 30mA,  V_{SHDN} = 0V,  0^{\circ}C \leq T_A \leq 70^{\circ}C \\ 0mA \leq I_{OUT} \leq 30mA,  V_{SHDN} = 0V,  -40^{\circ}C \leq T_A \leq 85^{\circ}C \end{array}$	•	11.4 11.2		12.6 12.6	V V
I <sub>CC</sub>	Supply Current	No Load, $V_{SHDN} = 0V$ , $0^{\circ}C \le T_A \le 70^{\circ}C$ No Load, $V_{SHDN} = 0V$ , $-40^{\circ}C \le T_A \le 85^{\circ}C$	•		0.5 0.5	1.0 1.3	mA mA
I <sub>SHDN</sub>	Shutdown Supply Current	No Load, V <sub>SHDN</sub> = V <sub>CC</sub>	•		0.5	10	μA
f <sub>OSC</sub>	Oscillator Frequency	V <sub>CC</sub> = 5V, I <sub>OUT</sub> = 30mA	•		300		kHz
	Power Efficiency	V <sub>CC</sub> = 5V, I <sub>OUT</sub> = 30mA	•		74		%
R <sub>SW</sub>	V <sub>CC</sub> to V <sub>OUT</sub> Switch Impedance	$V_{CC} = V_{SHDN} = 5V, I_{OUT} = 0mA$	•		0.18	2	kΩ
VIH	SHDN Input High Voltage		•	2.4			V
V <sub>IL</sub>	SHDN Input Low Voltage		•			0.8	V
	SHDN Input Current	$\begin{array}{l} V_{CC} = 5V,  V_{SHDN} = 0V,  0^{\circ}C \leq T_A \leq 70^{\circ}C \\ V_{CC} = 5V,  V_{SHDN} = 0V,  -40^{\circ}C \leq T_A \leq 85^{\circ}C \\ V_{CC} = 5V,  V_{SHDN} = 5V \end{array}$	•	-20 -35	-10 -10 0.06	-5 -5 10	μΑ μΑ μΑ
t <sub>ON</sub>	Turn-On Time	C1 = C2 = $0.22\mu$ F, C <sub>IN</sub> = C <sub>OUT</sub> = $4.7\mu$ F, (Figures 1, 2)			500		μs
t <sub>OFF</sub>	Turn-Off Time	C1 = C2 = $0.22\mu$ F, C <sub>IN</sub> = C <sub>OUT</sub> = $4.7\mu$ F, (Figures 1, 2)			3.3		ms

The  $\bullet$  denotes specifications which apply over the full operating temperature range.

**Note 1:** Absolute maximum ratings are those values beyond which the life of the device may be impaired.

**Note 2:** All currents into device pins are positive; all currents out of device pins are negative. All voltages are referenced to ground unless otherwise specified.

**Note 3:** All typicals are given at  $V_{CC} = 5V$ ,  $T_A = 25^{\circ}C$ .



## TYPICAL PERFORMANCE CHARACTERISTICS



## **PIN FUNCTIONS**

C1<sup>-</sup> (Pin 1): C1 Negative Input. Connect a  $0.22\mu$ F capacitor C1 between C1<sup>+</sup> and C1<sup>-</sup>.

**C1<sup>+</sup> (Pin 2):** C1 Positive Input. Connect a  $0.22\mu$ F capacitor C1 between C1<sup>+</sup> and C1<sup>-</sup>.

C2  $^-$  (Pin 3): C2 Negative Input. Connect a 0.22µF capacitor C2 between C2<sup>+</sup> and C2<sup>-</sup>.

C2+ (Pin 4): C2 Positive Input. Connect a  $0.22\mu$ F capacitor C2 between C2+ and C2<sup>-</sup>.

 $V_{CC}$  (Pin 5): Positive Supply Input Where 4.75V  $\leq V_{CC} \leq$  5.5V. Connect a 4.7µF bypass capacitor C<sub>IN</sub> to ground.

 $V_{OUT}$  (Pin 6): 12V Output. Connect a 4.7µF bypass capacitor C\_{OUT} to ground. When in the shutdown mode V\_{OUT} = V\_{CC}.

GND (Pin 7): Ground.

**SHDN (Pin 8):** Logic Level Shutdown Pin. Application of a logic low at SHDN pin will place the regulator in normal operation. With no external connection, or with SHDN tied to  $V_{CC}$ , the device will be put into shutdown mode. Connect to GND for normal operation. In shutdown mode the charge pump is turned off and  $V_{OUT} = V_{CC}$ .



# **BLOCK DIAGRAM**



## TIMING DIAGRAMS









## **APPLICATIONS INFORMATION**

## Operation

The LTC1262 uses a charge pump tripler to generate 12V from a  $V_{CC}$  of 5V. The charge pump operates when clocked by a 300kHz oscillator. When the oscillator output is low, C1 and C2 are connected between V<sub>CC</sub> and GND, charging them to  $V_{CC}$ . When the oscillator output goes high, C1 and C2 are stacked in series with the bottom plate of C1 pulled to  $V_{CC}$ . The top plate of C2 is switched to charge  $C_{OUT}$  and  $V_{OUT}$  rises.  $V_{OUT}$  is regulated to within 5% of 12V by an oscillator pulse gating scheme. A resistor divider senses  $V_{OUT}$ . When the output of the divider ( $V_{DIV}$ ) is less than the output of a bandgap (V<sub>BGAP</sub>) by the hysteresis voltage (V<sub>HYST</sub>) of the comparator, oscillator pulses are applied to the charge pump to raise  $V_{OUT}$ . When  $V_{DIV}$  is above  $V_{BGAP}$ by V<sub>HYST</sub>, the oscillator pulses are prevented from clocking the charge pump.  $V_{OUT}$  drops until  $V_{DIV}$  is below  $V_{BGAP}$ by V<sub>HYST</sub> again. The gates of all internal switches are driven between V<sub>OUT</sub> and GND. An internal diode ensures that the LTC1262 will start up under load by charging  $C_{OUT}$ to one diode drop below  $V_{CC}$ .

To reduce supply current the LTC1262 may be put into shutdown mode by floating the SHDN pin or taking it to V<sub>CC</sub>. In this mode the bandgap, comparator, oscillator and resistor divider are switched off to reduce supply current to typically 0.5µA. At the same time an internal switch shorts V<sub>OUT</sub> to V<sub>CC</sub>; V<sub>OUT</sub> takes 3.3ms to reach 5.1V (see t<sub>OFF</sub> in Figure 1). When the SHDN pin is low, the LTC1262 exits shutdown and the charge pump operates to raise V<sub>OUT</sub> to 12V. V<sub>OUT</sub> takes 500µs to reach the lower regulation limit of 11.4V (see t<sub>ON</sub> in Figure 1).

## **Choice of Capacitors**

The LTC1262 is tested with the capacitors shown in Figure 2. C1 and C2 are  $0.22\mu$ F ceramic capacitors and C<sub>IN</sub> and C<sub>OUT</sub> are  $4.7\mu$ F tantalum capacitors. Refer to Table 1 if other choices are desired.

CERAMIC	TANTALUM	ALUMINUM
0.22µF to 1µF	Not	Not
	Recommended	Recommended
2μF (Min)	4.7µF (Min)	10μF (Min)
1μF (Min)	4.7μF (Min)	10µF (Min)
	0.22μF to 1μF 2μF (Min)	0.22µF to 1µF Not Recommended 2µF (Min) 4.7µF (Min)

C1 and C2 should be ceramic capacitors with values in the range of 0.22 $\mu$ F to 1 $\mu$ F. Higher values provide better load regulation. Tantalum capacitors are not recommended as the higher ESR of these capacitors degrades performance when the load current is above 25mA with V<sub>CC</sub> = 4.75V.

 $C_{IN}$  and  $C_{OUT}$  can be ceramic, tantalum or electrolytic capacitors. The ESR of  $C_{OUT}$  introduces steps in the  $V_{OUT}$  waveform whenever the charge pump charges  $C_{OUT}$ . This tends to increase  $V_{OUT}$  ripple. Ceramic or tantalum capacitors are recommended for  $C_{OUT}$  if minimum ripple is desired. The LTC1262 does not require a  $0.1\mu F$  capacitor between  $V_{CC}$  and  $V_{OUT}$  for stability.

## **Maximum Load Current**

The LTC1262 will source up to 50mA continuously without any damage to itself. **Do not short the V<sub>OUT</sub> pin to ground**. If the V<sub>OUT</sub> pin is shorted to ground, irreversible damage to the device will result.



# TYPICAL APPLICATION



5V to 3.3V/10A Converter

Burst Mode is a trademark of Linear Technology Corporation.



## **PACKAGE DESCRIPTION** Dimensions in inches (millimeters) unless otherwise noted.



N8 Package

\*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

> S8 Package 8-Lead Plastic Small Outline (Narrow 0.150) (LTC DWG # 05-08-1610)



\*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

\*\*DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE



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SO8 0996

# TYPICAL APPLICATION





#### Output Voltage vs Output Current for Two Paralleled Devices

## **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LT1106*	Micropower Step-Up DC/DC Converter for PCMCIA Card Flash Memory	PCMCIA Card Power Control, 9µA I <sub>SHDN</sub> , Small SMT Components, Requires External Inductor
LT1109-12	Micropower Low Cost DC/DC Converter Adjustable and Fixed 12V	Three-Lead Z Package, Requires External Inductor
LT1109A-12	Micropower DC/DC Converter Flash Memory VPP Generator Adjustable and Fixed 12V	Requires External Inductor
LT1301	Micropower High Efficiency 5V/12V Step-Up DC/DC Converter for Flash Memory	$7\mu A~I_{SHDN},~SMT$ Inductor and Capacitors
LT1309	500kHz Micropower DC/DC Converter for Flash Memory	Small SMT Inductor and Capacitors, 6µA I <sub>SHDN</sub>

\* See also LT1312/LT1313 PCMCIA VPP drivers/regulators, LT1314/LT1315 PCMCIA switch matrix and the LTC1470/LTC1471/LTC1472 Protected V<sub>CC</sub> and VPP switching matrices

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