3-CHANNEL VIDEO AMPLIFIER FOR HIGH-RESOLUTION COLOR DISPLAY

DESCRIPTION

The M52307P/SP are semiconductor integrated circuits that have three channels of built-in amplifiers in the broad-band video amplifier series (M51392P, M51399P, M51387P) having a band of 130MHz.

Every channel is provided with a broad-band amplifier, main/ sub contrast control, and main/sub brightness control functions.

Accordingly, these ICs have an optional configuration for use in high-resolution color display monitors.

FEATURES

- The M52307P/SP use a new bi-polar wafer processing to realize low power dissipation so that three channels can be built in the amplifier. (Vcc = 12V, lcc = 63mA)
- Input : 0.7VP-P(Typ.)

Output : 4.5VP-P(Max.)

Frequency band: 130MHz (at 3VP-P)

- The main control adjusts 3 channels of contrast and brightness at the same time, and the sub control adjusts each channel independently.
- Since the feedback circuit is built in the IC, a stable DC level is obtained at the output pins of the IC.



RECOMMENDED OPERATING CONDITION

Supply Voltage range	V
Rated Supply Voltage12.0	V

APPLICATION

CRT display





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ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Ratings	Unit
Vcc	Supply voltage	13.0	· V
Pa	Power dissipation	2000	mW
Topr	Operating temperature	-20~85	°C
Tstg	Strage temperature	-40~150	°C
Vopr	Recommended operating supply voltage	12.0	V
Vopr'	Recommended operating supply voltage range	11.5~12.5	V
Surge	Electrostatic discharge	±200	V

ELECTRICAL CHARACTERISTICS (Ta=25°C, unless otherwise noted)

.

		Test		Input		·	Test conditions					Limits	,	
Symbol	Parameter	point	SW10 R-ch	SW6 G-ch	SW2 B-ch	V3	V13	V15	V26	SW14	Min.	Typ.	Max.	Unit
lcc	Circuit current	A	a _	a _	a _	12.0	12.0	5.0	. –	b SG6	45	72	110	mA
Vomax	Output dynamic range	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	12.0	12.0	Vari- able		a 	5.8	6.8	9.0	V р-р
Vimex	Maximum input voltage	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	12.0	6.0	Vari- able	-	a ' 	1.9	2.4	2.9	∨р-р
Gv	Maximum gain	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	12.0	12.0	Vī	` _	a 	13.0	17.0	20.0	dB
ΔGv	Relative maximum gain	-	-	-	-	-	-	-		-	0.8	1.0	1.2	-
VCR1	Contrast control charac- teristics (standard)	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	12.0	6.0	Vī		a 	4.1	7.4	10	dB
∆VCR1	Relative contrast control characteristics (standard)	-	-	-	-	-	-	-	-	-	0.8	1.0	1.2	-
VCR2	Contrast control charac- teristics (minimum)	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	12.0	3.5	Vī	-	а ~	5	30.0	70.0	mVp-p
∆VCR2	Relative contrast control characteristics (minimum)	· ·	-	-	-	-	. –	-	-	_	0.8	1.0	1.3	-
VSCR1	Sub contrast control characteristics (standard)	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	6.0	12.0	Vт	-	8	10.0	14.0	18.0	dB
∆VSCR1	Relative sub contrast control characteristics (standard)	-	-		-	-	-	-		-	0.8	.1.0	1.2	. –
VSCR2	Sub contrast control characteristics (minimum)	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	3.0	12.0	Vт	-	8 -	100.0	300.0	860.0	mVp-p
∆Vscr2	Relative contrast control characteristics (minimum)	-		-		-	-	-	_		0.8	-1.0	1.2	-
VCR3	Contrast/sub contrast control characteristics (Standard for both con- trast and sub contrast)	T.P20 T.P24 T.P28	b SG1	b SG1	b SG1	6.0	6.0	Vт	-	a _	900	1300	1700	mVp-p
AVCR3	Relative contrast/sub contrast control charac- teristics (Standard for both contrast and sub contrast)	-	-	-	-	-	-	-	-	-	0.8	1.0	1.2	-



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				Input		Test conditions				Limits				
Symbol	Parameter	Test point	SW10 R-ch	· · ·	SW2 B-ch	∨3	V13	V15	V26	SW14	Min.	Тур.	Max.	Unit
VB1	Brightness control char- acteristics (maximum)	T.P20 T.P24 T.P28	a -	a 	a _	12.0	12.0	5.5	-	b SG6	3.7	4.3	4.9	v
ΔV81	Relative brightness con- trol characteristics (maximum)	-	-	-	-	-	-	-	_	-	-100.0	0.0	100.0	mV
VB2	Brightness control char- acteristics (standard)	T.P20 T.P24 T.P28	8	a 	a -	12.0	12.0	5.0	-	b SG6	3 .1	3.7	4.3	v
∆∨в2	Relative brightness con- trol characteristics (standard)	-	-	-	-	-	-	_	-	-	100.0	0.0	100.0	m∨
VB3	Brightness control char- acteristics (minimum)	T.P20 T.P24 T.P28	a ~	a 	a _	12.0	12.0	4.5	-	b SG6	2.6	3.2	3.9	VDC
∆∨вз	Relative brightness con- trol characteristics (minimum)	-	-		-	-	_		-	-	100.0	0.0	100.0	٣V
VSB1	Sub brightness control characteristics (maximum)	T.P20 T.P24 T.P28	a ~	a -	a -	12.0	12.0	5.0	4.0	b SG6	2.4	3.1	3.8	VDC
ΔVSB1	Sub brightness control characteristics (minimum)	T.P20 T.P24 T.P28	a -	a -	a 1	12.0	12.0	5.0	3.5	b SG6	2.3	3.0	3.7	VDC
FC1	Frequency charac- teristics I (f=50MHz Max.)	T.P20 T.P24 T.P28	b SG3	ь SG3	b SG3	12.0	7.5	VT	-	a _	-2	-1	3	dB
ΔFc1	Relative frequency char- acteristics I (f=50MHz Max.)	-	-	-	-	-	-	-	-	-	-1.0	0.0	1.0	dB
FC1'	Frequency charac- teristics I (f=130MHz Max.)	T.P20 T.P24 T.P28	b SG4	b SG4	ь SG4	12.0	7.5	Vт	-	a _	-3	-2	3	dB
∆Fc1'	Relative frequency char- acteristics I (f=130MHz Max.)	-	-	-	-	-	-	-	-	-	-1.0	0.0	1.0	dB
FC2	Frequency charac- teristics II (f=50MHz standard)	T.P20 T.P24 T.P28	b SG3	ь SG3	b SG3	12.0	6.5	Vт	-	a _	-1	0	3	dB
FC2'	Frequency charac- teristics II (f=130MHz standard)	T.P20 T.P24 T.P28	b SG4	b SG4	b SG4	12.0	6.5	Vτ	-	a -	-2.5	0	3	dB
Fc5	Frequency charac- teristics III (f=50MHz minimum)	T.P20 T.P24 T.P28	ь SG3	b SG3	b SG3	12.0	5.0	VT	-	a 	-0.5	0	2	dB
FC5'	Frequency charac- teristics III (f=130MHz minimum)	T.P20 T.P24 T.P28	b SG4	b SG4	b SG4	12.0	5.0	VT	-	a -	-0.5	0	2	dB
C.T.1	Crosstalk 1 (f=50MHz)	T.P20 T.P24 T.P28	b SG3	a -	a ~	12.0	12.0	VT	-	a -	-	-32	-20	dB
C.T.1'	Crosstalk I (f=130MHz)	T.P20 T.P24 T.P28	b SG4	a 	a _	12.0	12.0	V٦	-	a _	-	-22	-15	dB
C.T.2	Crosstalk II (f=50MHz)	T.P20 T.P24 T.P28	a -	b SG3	8 —	12.0	12.0	VT	-	a _	-	-32	-20	dB
C.T.2'	Crosstalk II (f=130MHz)	T.P20 T.P24 T.P28	a -	b SG4	a -	12.0	12.0	VT	-	a _	-	-22	-15	dB

ELECTRICAL CHARACTERISTICS (cont.)



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		Test		Input			Tes	t condi	tions		Limits			{
Symbol	Parameter	point	SW10 R-ch	SW6 G-ch	SW2 B-ch	٧З	V13	V15	V26	SW14	Min.	Түр.	Max.	Unit
С.Т.З	Crosstalk III (f=50MHz)	T.P20 T.P24 T.P28	a 	â 	b SG3	12.0	12.0	۲V	-	a -	+	-32	-20	đB
C.T.3'	Crosstalk III (f=130MHz)	T.P20 T.P24 T.P28	a 	- a 	b SG4	12.0	12.0	Vτ	-	8	-	-22	-15	dB
Tr	Pulse characteristics (T.P20 T.P24 T.P28	b SG5	b SG5	b SG5	12.0	7.0	3.0	-	b SG6	-	2	- 4	nsec
Tf	Pulse characteristics II	T.P20 T.P24 T.P28	b SG5	b SG5	b SG5	12.0	7.0	3.0	-	b SG6	-	3	6	nsec
V14th	Clamp pulse threshold voltage	T.P20 T.P24 T.P28	a _	a 	a -	12.0	12.0	3.0	-	b SG6	0.7	1.5	2.5	VDC
W14	Clamp pulse minimum width	T.P20 T.P24 T.P28	a _	a 	a 	12.0	12.0	3.0	-	b SG6	-	0.3	1.0	µsec
V27	Hold voltage	T.P19 T.P23 T.P27	a _	a _	a -	12.0	12.0	3.0	-	b SG6	4.6	5.2	5.8	VDC
Росн	Pedestal voltage temperature characteristics 1	T.P20 T.P24 T.P28	b SG7	b SG7	b SG7	12.0	12.0	3.5	-	b SG6	-0.3	0	0.3	VDC
PdCL	Pedestal voltage temperature characteristics 2	T.P20 T.P24 T.P28	b SG7	b SG7	b SG7	12.0	12.0	3.5	-	b SG6	-0.3	0	0.3	VDC

ELECTRICAL CHARACTERISTICS (cont.)

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ELECTRICAL CHARACTERISTICS TEST METHOD

- Note1. The switch (SW) numbers for the signal input pin and pulse input pin have already been given in Attached Table 1; therefore, only the switch numbers for the external power supply will be given in the following notes.
- Note2. SUB BRIGHT voltages V18, V22, V26 are normally set at the same value, which are represented by V26 in Attached Table 1.

SUB CONTRAST voltages V3, V7, V11 are also set at the same value, which are represented by V3 in Attached Table 1.

Icc 1 Fix SW18, SW22, SW26 on side "b."

2 The other conditions are as shown in Attached Table1 When SW1 is fixed on side "a," lcc is measured, using ampere meter A.

Vomax

- 1 Fix SW18, SW22, SW26 on side "b."
 - 2 V15 is set up in the following order:
 - a) SG1 is input to pin (1) (pins (6), (2)). V15 voltage is gradually increased, and when the upper side of the TP20 (TP24 and TP28) output waveform becomes distorted, V15 is read, which is taken as VTR1 (VTG1, VTB1).

In contrast to the above, when voltage V15 is gradually reduced, and the bottom side of TP20 (TP24, TP28) output waveform becomes distorted, V15 is read, which is taken as VTR2 (VTg2, VTB2).

TP20 output waveform (This is also the same with TP24 and TP28.).



T.P20 Output Waveform (This is also the same with T.P24 and T.P28.)

b) Accordingly, VT (VTR, VTG, VTB) is found by the following:

$$V_{TR}(V_{TG}, V_{TB}) = \frac{V_{TR1}(V_{TG1}, V_{TB1}) + (V_{TR2}(V_{TG1}, V_{TB1}))}{2}$$

This equation should be used properly, depending

When TP20 is measured, VTR1 should be used, and when TP24 and TP28 are measured, VTG1 and VTB should be used respectively.

- 3 After VTR (VTG, VTB) is set, gradually increase the amplitude of SG1 from 700mV, and measure the amplitude of the output waveform when the upper/lower output waveforms of TP20 (TP24, TP28) start distortion at the same time.
- Vimax From the condition in NOTE 2 above, change V13 to 6.0V as given in Attached Table 1, gradually increase the amplitude of the input signal from 700 mVP-P, and read the input signal amplitude when the output signal starts to be distorted.

Gv ∆Gv

- 1 Fix SW18, SW22 and SW26 on side "b," and also set the conditions as shown in Attached Table 1.
- 2 Input SG1 to pin10 (Pin (6), Pin (2)) and read the amplitude of TP20 (TP24, TP28) output at this time: it should be taken as Vor1 (Vog1, Vog1).
- 3 The maximum gain GV is determined by: $GV = 20 \text{ LOG } \frac{\text{VORI(VOG1, VOBI) } [V_{P-P}]}{0.7 }$
- 4 The relative maximum gain ∆Gv is calculated as follows:

 $\Delta Gv = Vor1/Vog1, Vog1/Vob1, Vob1/Vor1$

VCR1 AVCR1

- 1 The conditions are the same as in Attached Table 1 except that V13 is set at 6.0V.
- 2 Read the amplitude of TP20 (TP24, TP28) output at this time: it should be taken as VoR2 (Vog2, Vog2).
- 3 The contrast control characteristics VCR1 and relative contrast control characteristics Δ VCR1 are calculated as follows:

$$VCR1 = 20 \ LOG \frac{VORI(VOG1, VOBI)}{0.7} \ [VP-P]$$

 $\Delta V_{CR1} = V_{OR2}/V_{OG2}, V_{OG2}/V_{OB2}, V_{OB2}/V_{OR2}$

VCR2 Δ VCR2

- 1 The conditions are the same as in Attached Table I except that V13 is set at 3.0V.
- 2 Read the amplitude of TP20 (TP24, TP28) output at this time: it should be taken as VOR3 (VOG3, VOB3), which shall be VCR2.
- 3 The relative contrast control characteristics ΔVCR2 is:

on the output pin.		 ∆Vcr2 = Vor3/	Vog3, Vog3/Vob3,	Vовз/Vояз
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VSCR1 AVSCR1

- 1 The conditions are the same as in Attached Table 1 except that V3, V7 and V11 are set at 6.0V.
- 2 Read the amplitude of TP20 (TP24, TP28) output at this time: it should be taken as VOR4 (Vog4, Vob4).
- 3 The sub contrast control characteristics Vscn1 and relative sub contrast control characteristics AVscR1 are found by:

$$V_{SCR1} = 20 \ LOG \frac{V_{OR4}(V_{OG4}, V_{OB4}) \ [V_{P-P}]}{0.7 \ [V_{P-P}]}$$

 $\Delta V_{CR4} = V_{OR4}/V_{OG4}, V_{OG4}/V_{OB4}, V_{OB4}/V_{OR4}$

VSCR2 AVSCR2

- 1 The conditions are the same as in Attached Table 1 except that V3, V7 and V11 are set at 3.0 V.
- 2 Read the amplitude of TP20 (TP24, TP28) output at this time: it should be taken as VOR5 (VOG5, VOB5), which shall be VSCR2.
- 3 The relative sub contrast control characteristics ΔVCR2 is:

 $\Delta V_{CR2} = V_{OR5}/V_{OG5}, V_{OG5}/V_{OB5}, V_{OB5}/V_{OR5}$

VCR3 AVCR3

- 1. The conditions are the same as in Attached Table 1 except that V13, V3, V7 and V11 are set at 6.0 V.
- 2 Read the amplitude of TP20 (TP24, TP28) output at this time: it should be taken as Vore (Voge, VOB6).
- 3 The gain and relative gain when the contrast and sub contrast are standard are determined by:

$$V_{SCR1} = 20 \ LOG \frac{V_{OR4}(V_{OG4}, V_{OB4}) \ [V_{P}P]}{0.7 \ [V_{P}P]}$$

 $\Delta V_{SCR1} = V_{OR4}/V_{OG4}, V_{OG4}/V_{OB4}, V_{OB4}/V_{OR4}$

- 1 Fix SW18, SW22 and SW26 on side "b," and set the conditions as given in Attached Table 1.
- 2 Measure the output of TP20 (TP24, TP28) at this time with a voltmeter: it should be taken as VOR7 (Vog7, Vog7). This value is VB1.
- 3 Also calculate the difference between each channel from VOR7, VOG7 and VOB7.

The relative brightness control characteristics VB1 is found by:

 $\Delta V_{B1} = V_{OR7} - V_{OG7}$ [mV]

= VOG7 - VOB7

= Vob7 - Vor7

1 Fix SW18, SW22 and SW26 on side "b," and set

V82 ∆V82

- the conditions as given in Attached Table1. 2 Measure the output of TP20 (TP24, TP28) at this
- time, using a voltmeter: it should be taken as VOR7' (VOG7', VOB7'). This value is VB2.
- 3 Also calculate the difference between each channel from Vorz', Vogz' and Vorz'. The relative brightness control characteristics ΔVB2
 - is: $\Delta V_{B2} = V_{OR7'} - V_{OG7'}$ [mV]
 - = Vog7' Vob7' = Vob7' - Vor7'

VB3 ДVB3

- 1 Fix SW18, SW22 and SW26 on side "b" and set the conditions as given in Attached Table 1.
- 2 Measure the output of TP20 (TP24, TP28) at this time with a voltmeter: it should be taken as VOR7" (VOG7", VOB7"). This value is VB3.
- 3 Also calculate the difference between each channel from Vorr, Vogr and Vorr. The relative brightness control characteristics ΔVB3
 - is found by:

 $\Delta V_{B3} = V_{OR7"} - V_{OG7"}$ [mV]

> = VOG7" - VOB7" = VOB7" - VOR7"

VSB1 ∆VSB1

The conditions are the same as in NOTE10 above except that SW18, SW22 and SW26 are fixed on side "a" and SUB BRIGHT (V18, V22, V26) is set at 4.0V or 3.5V.

FC1AFC1 FC1'AFC1'

- 1 Fix SW18, SW22 and SW26 on side "b" and set the conditions as given in Attached Table 1.
- 2 Use SG3 and SG4. According to the procedure shown in NOTE 4 above, however, measure the amplitude of output waveform on TP20 (TP24, TP28).
- 3 When these measured amplitudes are output amplitudes VOR1, (VOG1, VOB1), VOR8 (VOG8, VOB8) and VOR9 (VOG9, VOB9) at SG1, SG3 and SG4 inputs respectively, the frequency characteristics FC1, FC1' are calculated as follows:
 - VOR8(VOG8, VOB8) [VP-P] Fc1 = 20 LOG VOR1(VOG1, VOB1) [VP.P] VOR9(VOG9, VOB9) [VP-P] Fc1' = 20 LOG VOR1(VOG1, VOB1) [VP.P]
- However, NOTE10-(3) above is excluded.



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- 4 For relative frequency characteristics ΔFC1, ΔFC1', calculate the difference between FC1 and FC1' for each channel.
- Fc2Fc2' The conditions are the same as given in NOTE 14 above except that CONTRAST (V13) is reduced to 6.5V.
- FcsFcs' The conditions are the same as given in NOTE 14 above except that CONTRAST (V13) is reduced to 4.5 V.

C.T.1 C.T.1'

- 1 Fix SW18, SW22, and SW26 on side "b" and set the conditions as given in Attached Table 1.
- 2 Input SG2 (or SG4) to pin 10 (R-ch) only and measure the amplitude of output waveforms on TP20 (TP24, TP28) at that time: these measurements should be taken as VOR, VOG and VOB.
- 3 The crosstalk CTI is determined by:

C.T.1= 20 LOG
$$\frac{V_{OG OR} V_{OB} [V_{P-P}]}{V_{OR} [V_{P-P}]}$$
 [dB]
(C.T.1')

- C.T.2 C.T.2'
 - Change the input pin from pin (10 (R-ch) to pin 6 (G-ch), and read the output in the same manner as in NOTE 17 above.
 - 2 The crosstalk C.T.2 is determined by: C.T.2= 20 LOG VOR OR VOB [VP.P] (C.T.2') [dB]
- C.T.3 C.T.3'
 - Change the input pin from pin (1) (R-ch) to pin
 (B-ch), and read the output in the same manner as in NOTE 17.

C.T.3= 20 LOG
$$\frac{V_{OR OR} V_{OG} [V_{P-P}]}{V_{OB} [V_{P-P}]}$$
 [dB]
(C.T.3')

- 1 Fix SW18, SW22 and SW26 on side "b" and set the conditions as given in Attached Table 1.
- 2 Measure the rise time Tr1 and fall time Tf1 between 10 and 90% of the input pulse with an active probe.
- 3 Next, measure the rise time Tr2 and fall time Tf2 between 10 and 90% of the output pulse with an active probe.
- 4 The pulse characteristics Tr, Tf are found by:

$$Tr (nsec) = -\sqrt{(Tr_2)^2 - (Tr_1)^2}$$

$$Tf (nsec) = -\sqrt{(Tf_2)^2 - (Tf_1)^2}$$



V14th.

- 1 Fix SW18, SW22 and SW26 on side "b" and set the conditions as given in Attached Table 1.
- 2 While monitoring the output (approx. 2VDC) at this time, lower the SG6 level gradually and measure the SG6 level when the output reaches 0V.
- W14. While monitoring the output under the conditions given in NOTE 21 above, decrease the SG6 pulse width gradually.

Also measure the SG6 pulse width when the output becomes 0V.

W27.

- 1 Fix SW18, SW22 and SW26 on side "b."
- 2 Read TP19, TP23 and TP27 with a voltmeter.

PDCH. PDCL.

- 1 Fix SW18, SW22 and SW26 on side "b" and set the conditions as given in Attached Table1.
- 2 Measure the pedestal voltage at a room temperature: it should be taken as Poc1.
- 3 Next, measure the pedestal voltage at -20 and 85°C: these voltages should be taken as Poc2 and Poc3.
- 4 PDCH = $PDC_1 PDC_2$
 - PDCH = PDC1 PDC2PDCL = PDC1 - PDC3



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GND

GND

12

11

100 µ # 10.01 / V11 #

Vcc

13

עייע דוד

 π

14

€ SW14 ẩ∫b

SG6⊙ \$50

 $\frac{1}{m}$

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GND

GND

8

7

 π

Vcc

9

10

a b

¢sw10

GND

GND

4

3

 $\frac{1}{2}$

Vcc

5

6

100 µ #

a b

}sw6

TEST CIRCUIT

Vcc

1

٩

12V SW1 a

47 µ

2

þ

}sw2

å

SG1 SG2 SG3 ⇔ ≵50 SG4 SG5 777 777

μ#∓⁺ +0.01μ *π* π

Units Resistance: Ω Capacitance: F



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TYPICAL CHARACTERISTICS













Vcc 12V Main contrast 12V Main brightness 3.2V Sub brightness 0V Input signal level 0.7Vp-p



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] 117 (mV)



1 (nsec)

FALL TIME

1.75 (nsec)



Square wave Input amplitude (level) 0.70 (Vpp) Trin 1.85 (nsec)

1

2.65 (nsec)

Output amplitude (level) 4.0 (Vpp) Trout 2.65 (nsec) Vcc = 12V Main contrast 7.5V Sub contrast 12V

Output pin

Main brightness 3.2V

Sub brightness - OPEN

680

1 (nsec)

Output signal



Output signal

Output amplitude (level) 4.0 (Vpp) Tfout 4.0 (nsec) Tfout Vcc = 12V Main contrast 7.5V Sub contrast 12V Main brightness 3.2V Sub brightness OPEN

$$\begin{array}{c} \bigcirc & \text{Output pin} \\ & & 680 \\ & & & \\ &$$

 $=\sqrt{4^2-1.75^2}$



 $Tr = \sqrt{(Trout)^2 - (Trin)^2}$



670 (mV)

Units Resistance: Ω Capacitance: F

MITSUBISHI ICs (TV)

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3-CHANNEL VIDEO AMPLIFIER FOR HIGH-RESOLUTION COLOR DISPLAY

APPLICATION EXAMPLE 1





APPLICATION EXAMPLE 2

3-CHANNEL VIDEO AMPLIFIER FOR HIGH-RESOLUTION COLOR DISPLAY

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MITSUBISHI ICs (TV)



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3-CHANNEL VIDEO AMPLIFIER FOR HIGH-RESOLUTION COLOR DISPLAY

Pin No.	Name	Voltage and wave information	Peripheral circuit of pins	Remarks
1	Vcc (B-ch)			The voltages applied to
5	Vcc (G-ch)	12V		three channels should b
9	Vcc (R-ch)			equal to each other.
		· · · · · · · · · · · · · · · · · · ·		
• .			Vcc	÷
		•	₹24.7k	
				<i>t</i>
2 6 10	B-IN			
6	G-IN	2.9V	▼ .	
10	R-IN			. · ·
			Š 3.6k	11
	,		1 · · · · · · · · · · · · · · · · · · ·	
			GND	
			······································	
	· · ·			
			Vcc	
				· · ·
0				
9	B SUB CONTRAST	4.01/		
3 7 11	G SUB CONTRAST	4.0V	§ 72k Q	
9	R SUB CONTRAST		P P P	
			0.12mA	
4, Ø	GND (B-ch)	•		
8, Ø	GND (G-ch)	GND		
Q, D	GND (R-ch)			
		<u></u>		
			4	
				u -
13	CONTRAST	6.9V		
			₹72k	
			0.4mA	



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3-CHANNEL VIDEO AMPLIFIER FOR HIGH-RESOLUTION COLOR DISPLAY





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DESCRIPTION OF PIN (cont.)

Pin No.	Name	Voltage and wave information	Peripheral circuit of pins	Remarks
୦୦ ୧୬	R OUT G OUT B OUT	Variable		A resistor is required at the GND side. Choose any resistance value under 15mA according t the driving capability required.

PRECAUTIONS FOR APPLICATION M52307P/SP Usage Information

- 1) Clamp Pulse Input
- The circuit in the diagram on the right shows the configuration of the clamp pulse input. The input is: $VTH = 2.2V - Diode \times 1 = 1.5 V$
- 2.2V or more voltage is limited by LIM.
- Accordingly, the recommended voltage is as shown in the diagram on the right.
- For the pulse width:
- 1.0µsec Min. at 15kHz
- 0.5µsec Min. at 30kHz
- 0.3µsec Min. at 64kHz

are recommended.



Note that in general, wiring inside the unit for the clamp pulse is long and it is often produced from the high voltage side or connected indirectly to an external pin; therefore, it is liably exposed to a strong surge input.

A protective circuit as shown in the diagram on the right is therefore recommended.



2) Main/Sub Brightness Control



The diagram above shows the principle. The M51387P includes the sub brightness pin directly in the signal feedback loop; however, this IC incorporates the sub brightness pin indirectly, and no signal is produced on the sub brightness pin.

2-1) Main Brightness Pin

Use the main brightness pin within the range from 1V to 6 V.

This control characteristic is as shown in the chart below.

MAIN BRIGHTNESS CONTROL CHARACTERISTICS



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2-2) Sub Brightness Pin

As described above, note that the internal circuit configuration of the sub brightness pin completely differs from that of the M51387P.

 a) If SUB BRIGHTNESS is not changed: Connect all control pins to VCC. However, if external disturbing voltage intrudes into this pin due to wiring on PC board for applications, which may adversely affect IC output, consider the use of an additional by-pass capacitor as well.

b) If SUB BRIGHTNESS is changed:





SUB BRIGHTNESS VOLTAGE (V)

2–3) Holding Capacitor Capacitance

IC requires 1,000P or more (when fH = 15kHz). However, this capacitance varies, depending on the holding duration (time other than clamping): as the holding duration is longer, a larger capacitance is required.

In view of IC applications, the response is quicker as the capacitance is smaller, and as it is larger, the response will become more stable.

Consequently, set this capacitance optionally according to the signal or clamp pulse contents (pulse condition at vertical sync timing in particular).

M52307P Crosstalk

Testing Conditions:

Main contrast pin voltage	12V
Sub contrast pin voltage	12V
Main brightness pin voltage	5V
Sub brightness pin voltage	Open
Input signal 0.7VP-P sine wave	

			Inpu	t Frequen	су				
		10MHz 50MHz 75MHz 100MHz U							
CT1	R→G	-45	-29	-23	-18	dB			
	R→B	-60	-38	-30	-20	dB			
CT2	G→R	-60	-34	-23	-18	dB			
	G→B	-45	-26	-20	-18	dB			
СТЗ	B→R	-65	-35	-23	-19	dB			
	B→G	-60	-40	29	-26	dB			

For crosstalk CT1, input a signal only to pin 10(R-ch) and take the output waveform amplitudes at pins (29, (29, 29)), at that time as VOR, Vog and Vog.

$$CT1 = 20LOG_{10} \quad \frac{V_{OG} \text{ or } V_{OB}}{V_{OB}} \quad [dB]$$

For crosstalk CT2, the conditions are the same as with CT1 above except that the input pin is changed to pin (G) (G-ch).

$$CT2= 20LOG_{10} \quad \frac{V_{OR OR} V_{OB}}{V_{OG}} \quad [dB]$$

For crosstalk CT3, the conditions are the same as CT1 above except that the input pin is changed to pin 2 (B-ch).

$$CT3= 20LOG_{10} \quad \frac{V_{OR} OR V_{OG}}{V_{OB}} [dB]$$

PC Board Fabricating Precautions

Since a broad-band amplifier is built in this IC and oscillation may occur due to the shape of PC board wiring; note the following points:

- Make the output pin and resistor wiring as short as possible.
- Make the output pin load capacitance as small as possible.
- Install a by-pass capacitor on the Vcc GND, DC line near or around the pin.
- For Vcc, use a stable power supply. (Independent use of four units is more preferable.)
- Insertion of 10 or more Ω resistor between the output pin and circuit in the next stage makes it hard to oscillate.
- Insertion of a coil or resistor, such as to 16 pin Vcc produces an effect, depending on PC board.
- Also pay attention to a leak signal from the power amplifier.
- Make GND as wide as possible; basically, plane grounding is required.
- Also ground the hold capacitance to stable GND, which is as near to the pin as possible.

IC Operating Precautions

 It is recommended that the IC be used between pedestal voltage 2V and 3V. (Optimal distortion)

• Connect each input pin of this IC with a sufficiently low

