19-2269; Rev 1; 1/03 EVALUATION KIT AVAILABLE

+3.3V, 2.5Gbps Low-Power **Limiting Amplifiers**

General Description

The MAX3272/MAX3272A 2.5Gbps limiting amplifiers accept a wide range of input voltages and provide a constant-level output voltage with controlled edge speeds. Additional features include power detectors with programmable loss-of-signal (LOS) indication, an optional squelch function that mutes the data output signal when the input voltage falls below a programmable threshold, and an output polarity selector. These parts exhibit excellent jitter performance and have low power dissipation.

The MAX3272/MAX3272A feature current-mode logic (CML) data outputs that are tolerant of inductive connectors, and are available in a 4mm x 4mm QFN package or in die form (MAX3272 only). Along with the MAX3271, the MAX3272/MAX3272A are ideal for lowpower, compact optical receivers.

Applications

Gigabit Ethernet Optical Receivers Fibre Channel Optical Receivers System Interconnects 2.5Gbps Optical Receivers SONET/SDH Receivers

Features

- Single +3.3V Power Supply
- 33mA Supply Current
- 5ps Deterministic Jitter
- 90ps Edge Speed
- Output Squelch Function
- Programmable Loss-of-Signal Function
- CML Output Interface
- 20-Pin 4mm × 4mm QFN Package
- Selectable Output Polarity

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX3272EGP	-40°C to +85°C	20 QFN
MAX3272E/D	-40°C to +85°C	Dice*
MAX3272AEGP	-40°C to +85°C	20 QFN

*Dice are designed and guaranteed to operate from -40°C to +85°C, but are tested only at $T_A = +25$ °C.

Pin Configuration appears at end of data sheet.



Typical Operating Circuit

MIXIM

Maxim Integrated Products 1

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ABSOLUTE MAXIMUM RATINGS

Power-Supply Voltage (V_{CC})....-0.5V to +6.0V Voltage at IN+, IN-(V_{CC} - 2.4V) to (V_{CC} + 0.5V) Voltage at SQUELCH, CAZ1, CAZ2, TH, CLOS-0.5V to (V_{CC} + 0.5V) Voltage at LOS, LOS-0.5V to +6.0V Voltage at LEVEL.....-0.5V to +2.0V

Continuous Current at	
CML Outputs (OUT+, OUT-)	
Continuous Power Dissipation	1600mW
Storage Ambient Temperature	
Range (T _{STG})	55°C to +150°C
Operating Junction Temperature	
Range (T _J)	55°C to +150°C
Die Attach Temperature	+400°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = +3.0V to +3.6V, T_A = -40°C to +85°C. Typical values are at V_{CC} = +3.3V and T_A = +25°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL		CONDITIONS	MIN	ТҮР	MAX	UNITS	
Supply Current	Icc	(Note 2)			33	44	mA	
Input Data Rate					2.5		Gbps	
Input Voltage Range	VIN	Differential		15		1200	mV _{P-P}	
Output Deterministic Jitter		(Notes 3, 4, 5)			5	27	psp-p	
Random Jitter		(Notes 4, 6)			3		ps _{RMS}	
Data Output Edge Speed		(Notes 3, 4)	$15mV_{P-P} < V_{IN} \le 30mV_{P-P}$		90	130	200	
(20% to 80%)		(NOLES 3, 4)	$30mV_{P-P} \le V_{IN} \le 1200mV_{P-P}$		90	115	ps	
Differential Input Resistance	R _{IN} IN+ to IN-				100	105	Ω	
Input-Referred Noise					220		μV_{RMS}	
CML Output Voltage	VOUT	LEVEL open, F	$R_{LOAD} = 50\Omega$	550	750	1200	mV_{P-P}	
Output Signal when Squelched		Outputs AC-co	oupled		2.2		mV _{P-P}	
Power-Supply Noise Rejection	PSNR	f ≤ 2MHz (Note	e 7)		30		dB	
Low Frequency Cutoff	foo	C _{AZ} = open			0.9		MHz	
Low Frequency Culon	foc	$C_{AZ} = 0.1 \mu F$			1.5		kHz	
Output Resistance	Rout	Single ended	to V _{CC}	42.5	50	57.5	Ω	
Single-Ended Output Return		≤ 2.5GHz			10		dB	
Loss		2.5GHz to 4.00	GHz		9		uВ	
Differential Input Return Loss		4.0GHz			10		dB	
OUTPOL Input Limits	VIL					0.8	V	
Oon OE input Einits	VIH			2.4			v	
LOS Hysteresis		(Notes 3, 4, 8)		2	3.3		dB	
LOS Assert/Deassert Time		C _{CLOS} = open (Notes 3, 9, 10			1			
LOS Assel (Deasselt Time		$C_{CLOS} = 0.01$	uF (Notes 3, 9, 10)	2.3	50	100	μs	
Low LOS Assert Level		$R_{TH} = 20 k\Omega$ (N	Notes 3, 10)	4.5	6.5		mV _{P-P}	
Low LOS Deassert Level		$R_{TH} = 20k\Omega$ (N	Notes 3, 10)		9.5	12.7	mV _{P-P}	
Medium LOS Assert Level		$R_{TH} = 1k\Omega$ (No	otes 3, 10)	7.8	12.9		mV _{P-P}	
Medium LOS Deassert Level		$R_{TH} = 1k\Omega$ (No	otes 3, 10)		17.4	22.4	mV _{P-P}	
High LOS Assert Level		$R_{TH} = 80\Omega$ (No	otes 3, 10)	24.3	48		mV _{P-P}	





ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +3.0V \text{ to } +3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.$ Typical values are at $V_{CC} = +3.3V$ and $T_A = +25^{\circ}C$, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
High LOS Deassert Level		R _{TH} = 80Ω (Notes 3, 10)		73	124.7	mV _{P-P}
LOS Output High Voltage		Sinking 30µA	2.4			V
LOS Output Low Voltage		Sourcing 1.2mA			0.4	V
Squelch Input Current					400	μA

Note 1: Dice are designed and guaranteed from -40°C to +85°C but are tested only at $T_A = +25$ °C.

Note 2: Supply current measurement excludes the current of the CML output stage (16mA typical). See Figure 1, *Power-Supply Current Measurement*.

Note 3: Guaranteed by design and characterization.

Note 4: Input edge speed is controlled using 4-pole, lowpass Bessel filters with bandwidth approximately 75% of the maximum data rate.

Note 5: Deterministic jitter is measured with a K28.5 pattern (0011 1110 1011 0000 0101). Deterministic jitter is the peak-to-peak deviation from ideal time crossings, measured at the zero-level crossings of the differential output per ANSI X3.230, Annex A.

Note 6: Random jitter is measured with the minimum input signal. For Fibre Channel and Gigabit Ethernet applications, the peak-to-peak random jitter is 14.1 times the RMS random jitter.

Note 7: Power-supply noise rejection (PSNR) is calculated by the equation $PSNR = 20 \log (\Delta V_{CC}/(\Delta V_{OUT}))$, where ΔV_{OUT} is the change in differential output voltage due to the power-supply noise, ΔV_{CC} . See Power-Supply Noise Rejection vs. Frequency in the *Typical Operating Characteristics*.

Note 8: Hysteresis is defined as: $20 \times \log(V_{LOS-DEASSERT}/V_{LOS-ASSERT})$.

Note 9: Response time to a 10dB change in input power. For the specification guaranteed, the power is assumed to switch back and forth between two levels (separated by 10dB and equidistant from assert and deassert levels) outside of the two hysteresis thresholds.

Note 10: All power-detect AC parameters are guaranteed with a 2²³ - 1 PRBS, 2.5Gbps input, with the longest possible run of 80CID.

Typical Operating Characteristics

(V_{CC} = +3.3V, T_A = +25°C, unless otherwise noted.)



 $(V_{CC} = +3.3V, T_A = +25^{\circ}C, unless otherwise noted.)$

_Typical Operating Characteristics (continued)

VIN

VOUT

RANDOM JITTER vs. INPUT AMPLITUDE 10 9 8 RANDOM JITTER (psrms) 7 6 HYSTERESIS (5 4 3 2 1 0 100 1000 10,000 10 INPUT AMPLITUDE (mVp-p)



DATA OUTPUT EYE DIAGRAM (MINIMUM INPUT)





LOSS-OF-SIGNAL WITH SQUELCH

DATA OUTPUT EYE DIAGRAM (MAXIMUM INPUT)



INPUT RETURN LOSS vs. FREQUENCY 45 40 35 RETURN LOSS (dB) 30 25 20 INPUT F 15 10 5 0 10M 100M 10G 1G FREQUENCY (Hz)

OUTPUT RETURN LOSS vs. FREQUENCY















Typical Operating Characteristics (continued)





_Pin Description

PIN	NAME	FUNCTION
1, 4, 17	GND	Supply Ground
2	IN+	Noninverted Input Signal
3	IN-	Inverted Input Signal
5	TH	Loss-of-Signal Threshold Pin. Resistor to ground sets the LOS threshold.
6, 12, 15, 20	V _{CC}	Power Supply
7	CLOS	LOS Time-Constant Capacitor Connection. For SONET applications, $C_{CLOS} = 0.01 \mu$ F is recommended.
8	SQUELCH	Squelch Input. The squelch function is disabled when SQUELCH is not connected or set to TTL low level. When SQUELCH is set to TTL high level and LOS is asserted, the data outputs (OUT+, OUT-) are forced to static levels.
9	LOS	Noninverted Loss-of-Signal Output. LOS is asserted TTL high when the signal drops below the assert threshold set by the TH input. The MAX3272 does not have ESD protection on this pin. The MAX3272A has ESD protection on this pin.
10	LOS	Inverted Loss-of-Signal Output. LOS is asserted TTL low when the signal drops below the assert threshold set by the TH input. The MAX3272 does not have ESD protection on this pin. The MAX3272A has ESD protection on this pin.
11	LEVEL	Output Current Level. When this pin is not connected, the CML output current is approximately 16mA. When this pin is connected to ground, the output current increases to about 20mA.
13	OUT-	Inverted Data Output
14	OUT+	Noninverted Data Output
16	OUTPOL	Output Polarity Control Input. Connect to GND for an inversion of polarity through the limiting amplifier and connect to V_{CC} for normal operation.
18	CAZ2	Offset-Correction-Loop Capacitor Connection. A capacitor connected between this pin and CAZ1 extends the time constant of the offset correction loop. Typical value of C_{AZ} is 0.1µF.
19	CAZ1	Offset-Correction-Loop Capacitor Connection. A capacitor connected between this pin and CAZ2 extends the time constant of the offset correction loop. Typical value of C_{AZ} is 0.1µF.
EP	EXPOSED PAD	Connect the exposed paddle to board ground for optimal electrical and thermal performance.



Figure 1. Power-Supply Current Measurement

Detailed Description

Figure 2 is a functional diagram of the MAX3272/ MAX3272A, comprising a CML input buffer, power detector and loss-of- signal indicators, gain stage, offsetcorrection loop, and CML output buffer.

CML Input Buffer

The input buffer (Figure 3) provides 100Ω input impedance between IN+ and IN-. DC-coupling the inputs is not recommended; this prevents the DC offset-correction circuitry from functioning properly.

Power Detect and Loss-of-Signal Indicator

The MAX3272/MAX3272A are equipped with loss-of-signal (LOS) circuitry that indicates when the input signal is below a programmable threshold, set by resistor R_{TH} at the TH pin (see *Typical Operating Characteristics* for appropriate resistor selection). An averaging peakpower detector compares the input signal amplitude with this threshold and feeds the signal-detect information to the LOS outputs, which are internally terminated to 8k Ω (Figure 4).



Figure 2. Functional Diagram

///XI//I

_Interface Schematics



Figure 3. Input Circuit

Two control voltages VASSERT, and VDEASSERT, define the LOS assert and deassert levels. To prevent LOS chatter in the region of the programmed threshold, approximately 3.3dB of hysteresis is built into the LOS assert/deassert function. Once asserted, LOS is not deasserted until the input amplitude rises to the required level (VDEASSERT).

Gain Stage

The high-bandwidth gain stage provides approximately 42dB of gain.

Offset-Correction Loop

Due to the high gain of the amplifier, the MAX3272/ MAX3272A are susceptible to DC offsets in the signal path. In communications systems using NRZ data with a 50% duty cycle, pulse-width distortion present in the signal or generated by the transimpedance amplifier appears as input offset and is removed by the offsetcancellation loop. An external capacitor is required between CAZ1 and CAZ2 to decouple the offset-cancellation loop and determine the lower 3dB frequency of the signal path.



Figure 4a. LOS Output Circuit for MAX3272



Figure 4b. LOS Output Circuit for MAX3272A

CML Output Buffer

The MAX3272/MAX3272A CML output circuit (Figure 5) provides high tolerance to impedance mismatches and inductive connectors. The output current can be set to two levels using the LEVEL pin. When LEVEL is unconnected, the output current is approximately 16mA. Connecting LEVEL to ground sets the output current to approximately 20mA. The squelch function is enabled when the SQUELCH pin is set to a TTL high. This function holds OUT+ and OUT- to a static level whenever the input signal amplitude drops below the loss-of-signal threshold. This circuit is also equipped with a polarity selector, programmed by the OUTPOL pin. When





Figure 5. CML Output Circuit

this pin is connected to V_{CC} , no inversion will occur. When connected to ground, the output signal will be inverted.

___Design Procedure

Program the LOS Assert Threshold

External resistor R_{TH} programs the loss-of-signal threshold. See the LOS Threshold vs. R_{TH} graph in the the *Typical Operating Characteristics* section to select the appropriate resistor.

Select the Coupling Capacitors

When AC-coupling, input and output coupling capacitors (C_{IN} and C_{OUT}) should be selected to minimize the receiver's deterministic jitter. Jitter is decreased as the input low-frequency cutoff (f_{IN}) is decreased:

$f_{IN} = 1 / [2\pi(50)(C_{IN})]$

For ATM/SONET or other applications using scrambled NRZ data, select (C_{IN}, C_{OUT}) \geq 0.1µF, which provides f_{IN} < 32kHz. For Fibre Channel, Gigabit Ethernet, or other applications using 8B/10B data coding, select (C_{IN}, C_{OUT}) \geq 0.01µF, which provides f_{IN} < 320kHz. Refer to application note HFAN-1.1: *Choosing AC-Coupling Capacitors*.

Select the Offset-Correction Capacitor

The capacitor between CAZ1 and CAZ2 determines the time constant of the signal path DC offset-cancellation loop. To maintain stability, it is important to keep a one-decade separation between f_{IN} and the low-frequency cutoff (f_{OC}) associated with the DC offset-cancellation circuit. For ATM/SONET or other applications using scrambled NRZ data, f_{IN} < 32kHz, so f_{OCMAX} < 3.2kHz. Therefore, C_{AZ} = 0.1 μ F (f_{OC} = 2kHz). For Fibre Channel or Gigabit Ethernet applications, leave pins CAZ1 and CAZ2 open.

Program the LOS Time Constant

External capacitor C_{CLOS} programs the LOS assert and deassert times. When inputting data with many consecutive identical digits (CIDs), a longer time constant may be advantageous, so LOS does not flag incorrectly. In this case, connect the CLOS pin to a 0.01μ F capacitor to set the assert time in the range of 2µs to 100µs. For scrambled data where the mark density is kept at 50%, a shorter time constant may be desirable. Leave the CLOS pin open for a shorter time constant of about 1µs.

Applications Information

Optical Hysteresis

In an optical receiver, the electrical power change at the limiting amplifier is 2 times the optical power change.

As an example, if a receiver's optical input power (x) increases by a factor of two, and the preamplifier is linear, then the voltage input to the limiting amplifier also increases by a factor of two.

The optical power change is $10\log(2x/x) = 10\log(2) = +3dB$.

At the limiting amplifier, the electrical power change is:

$$10\log \frac{(2V_{IN})^2 / R_{IN}}{V_{IN}^2 / R_{IN}} = 10\log(2^2) = 20\log(2) = +6dB$$

The MAX3272 typical voltage hysteresis is 3.3dB. This provides an optical hysteresis of 1.65dB.

Wire Bonding Die

For high-current density and reliable operation, the MAX3272 uses gold metallization. Make connections to the dice with gold wire only, and use ball-bonding techniques (wedge bonding is not recommended). Die pad dimensions are 94.4 microns by 94.4 microns. Die thickness is 15 mils (0.375mm).

Pad Coordinates



		la coordinates
PAD	NAME	COORDINATES (µm)
1	GND	47, 836
2	IN+	47, 603
3	IN-	47, 425
4	GND	47, 237
5	TH	47, 47
6	V _{CC}	255, -154
7	CLOS	436, -154
8	SQUELCH	645, -154
9	LOS	850, -154
10	LOS	1063, -154
11	LEVEL	1331, 37
12	V _{CC}	1331, 212
13	OUT-	1331, 421
14	OUT+	1331, 573
15	V _C C	1331, 780
16	OUTPOL	1119, 1042
17	GND	957, 1042
18	CAZ2	773, 1042
19	CAZ1	583, 1042
20	N.C.	422, 1042
21	Vcc	268, 1042

Coordinates are for the center of the pad.

Coordinate 0, 0 is the lower left corner of the passivation opening for pad 5.



Chip Information

TRANSISTOR COUNT: 726 PROCESS: SiGe Bipolar SUBSTRATE: Insulator, Connect to GND DIE SIZE: 1.68mm × 1.57mm DIE THICKNESS: 15 mils

MAX3272/MAX3272A

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <u>www.maxim-ic.com/packages</u>.)



Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <u>www.maxim-ic.com/packages</u>.)

					C	ommon e	DIMENSI	ONS												
PKG	1	2L 4x4		1	6L 4x4		2	0L 4x4		2	4L 4x4									
SYMBOL	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.								
A	0.80	0.90	1.00	0.80	0.90	1.00	0.80	0.90	1.00	0.80	0.90	1.00								
A1	0.00	0.01	0.05	0.00	0.01	0.05	0.00	0.01	0.05	0.00	0.01	0.05		FXP	JSED	PA1	n va			2
A2	0.00	0.65	0.80	0.00	0.65	0.80	0.00	0.65	0.80	0.00	0.65	0.80	-			D2			E2	<u> </u>
A3		0.20 REF	-		0.20 REF			0.20 REI	F		0.20 REF	-	PKG.		MĪN.					MAX.
b	0.28	0.33	0.40	0.23	0.28	0.35	0.18	0.23	0.30	0.18	0.23	0.30					MAX.		NDM.	
D	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	G124		1.95	2.10	2.25	1.95	2.10	2.25
D1		3.75 BSC	>		3.75 BSC			3.75 BS	¢		3.75 BS0	2	G164		1.95	2.10	2,25	1.95	2.10	2.25
E	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	G204		1.95	2.10	2.25	1.95	2.10	2.25
E1		3.75 BSC			3.75 BSC			3.75 BS			3.75 BS0		G204		1.55	1.70	1.85	1.55	1.70	1.85
e		0.80 BSC			0.65 BSC			0.50 BS			0.50 BS		G244	4-1	1.95	2.10	2.25	1.95	2.10	2.25
k	0.25	-	-	0.25	-	-	0.25	-	-	0.25	-	-	G244	14-2	2.45	2.60	2.75	2.45	2.60	2.75
L	0.50	0.60	0.75	0.50	0.60	0.75	0.50	0.60	0.75	0.30	0.40	0.55								
N		12			16			20			24									
ND		3			4			5			6									
		3			4			5			6									
NE		-	-		<u> </u>			-			-									
P	0.00	0.42	0.60	0.00	0.42	0.60	0.00	0.42	0.60	0.00	0.42	0.60								
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