MC68HC05C9A MC68HCL05C9A MC68HSC05C9A

Advance Information

HCMOS
Microcontroller Unit



MC68HC05C9A MC68HCL05C9A MC68HSC05C9A

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MOTOROLA 3

Advance Information

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MC68HC05C9A — Rev. 5.0

List of Sections

Section 1. General Description19
Section 2. Memory
Section 3. Central Processor Unit (CPU)37
Section 4. Interrupts
Section 5. Resets
Section 6. Low-Power Modes55
Section 7. Input/Output Ports57
Section 8. Capture/Compare Timer
Section 9. Serial Communications Interface (SCI)73
Section 10. Serial Peripheral Interface (SPI)91
Section 11. Instruction Set103
Section 12. Electrical Specifications121
Section 13. Mechanical Specifications
Section 14. Ordering Information141
Appendix A. MC68HCL05C9A143
Appendix B. MC68HSC05C9A147
Appendix C. Self-Check Mode
Appendix D. M68HC05Cx Family Feature Comparisons

 $\mathsf{MC68HC05C9A} - \mathsf{Rev.}\ 5.0$

List of Sections

Table of Contents

Section 1. General Description

1.1	Contents
1.2	Introduction
1.3	Features
1.4	Mask Options
1.5	Software-Programmable Options
1.6	Functional Pin Descriptions
1.6.1	V _{DD} and V _{SS}
1.6.2	IRQ
1.6.3	OSC1 and OSC2
1.6.4	RESET27
1.6.5	TCAP
1.6.6	TCMP
1.6.7	PA0-PA7
1.6.8	PB0-PB7
1.6.9	PC0-PC728
1.6.10	PD0-PD5 and PD728
	Section 2. Memory
	•
2.1	Contents
2.2	Introduction
2.3	RAM31
2.4	ROM31
2.5	ROM Security
2.6	I/O Registers31

MC68HC05C9A — Rev. 5.0

	Section 3. Central Processor Unit (CPU)	
3.1	Contents	7
3.2	Introduction3	7
3.3	CPU Registers	8
3.3.1	Accumulator (A)	
3.3.2	Index Register (X)	
3.3.3 3.3.4	Program Counter (PC)	
3.3.5	Condition Code Register (CCR)4	
	Section 4. Interrupts	
4.1	Contents	
4.2	Introduction4	
4.3	Non-Maskable Software Interrupt (SWI)4	2
4.4	External Interrupt (IRQ or Port B)4	2
4.5	Timer Interrupt	4
4.6	SCI Interrupt	4
4.7	SPI Interrupt	4
	Section 5. Resets	
5.1	Contents	7
5.2	Introduction4	
5.3	Power-On Reset (POR)	8
5.4	RESET Pin	9
5.5	Computer Operating Properly (COP) Reset	0
5.5.1	COP Reset Register5	1
5.5.2	COP Control Register5	2
5.6	COP During Wait Mode	4
5.7	COP During Stop Mode	4
5.8	Clock Monitor Reset	4

Advance Information

 $\mathsf{MC68HC05C9A}-\mathsf{Rev.}\ 5.0$

	Section 6. Low-Power Modes
6.1	Contents
6.2	Introduction
6.3	Stop Mode
6.4	Wait Mode
	Coation 7 Innut/Outmut Donto
7 4	Section 7. Input/Output Ports
7.1	Contents
7.2	Introduction
7.3	Port A
7.4	Port B
7.5	Port C
7.6	Port D
	Section 8. Capture/Compare Timer
8.1	Content
8.2	Introduction
8.3	Timer Operation
8.3.1	Input Capture
8.3.2	Output Compare
8.4	Timer I/O Registers64
8.4.1	Timer Control Register
8.4.2	Timer Status Register66
8.4.3	Timer Registers
8.4.4	Alternate Timer Registers
8.4.5	Input Capture Registers
8.4.6	Output Compare Registers70
8.5	Timer During Wait Mode71
8.6	Timer During Stop Mode71
5	Section 9. Serial Communications Interface (SCI)

Content......73

9.1

9.2

9

Table of Contents

9.3	Features
9.4	SCI Receiver Features
9.5	SCI Transmitter Features
9.6	Functional Description76
9.7	Data Format
9.8 9.8.1 9.8.2	Receiver Wakeup Operation78Idle Line Wakeup.79Address Mark Wakeup.79
9.9	Receive Data In (RDI)
9.10	Start Bit Detection81
9.11	Transmit Data Out (TDO)
9.12.1	SCI I/O Registers
9.12.2	SCI Control Register 1
9.12.3 9.12.4	SCI Control Register 2
9.12.5	Baud Rate Register
	Section 10. Serial Peripheral Interface (SPI)
10.1	Content
	Introduction91
10.3	Features
10.4 10.4.1 10.4.2 10.4.3 10.4.4	SPI Signal Description .92 Master In/Slave Out (MISO) .93 Master Out/Slave In (MOSI) .93 Serial Clock (SCK) .94 Slave Select (SS) .94
10.5	Functional Description95
	SPI Registers

Advance Information

 $\mathsf{MC68HC05C9A}-\mathsf{Rev.}\ 5.0$

Section 11. Instruction Set

MC68HC05C9A — Rev. 5.0

Contents
Introduction
40-Pin Plastic Dual In-Line (DIP) Package (Case 711-03)
42-Pin Plastic Shrink Dual In-Line (SDIP) Package (Case 858-01)
44-Lead Plastic-Leaded Chip Carrier (PLCC) (Case 777-02)
44-Lead Quad Flat Pack (QFP) (Case 824A-01)
Section 14. Ordering Information
Contents
Introduction
MC Order Numbers141
Appendix A. MC68HCL05C9A
Contents
Introduction
Operating Temperature143
DC Electrical Characeristics
Appendix B. MC68HSC05C9A
Contents
Introduction147
Operating Temperature148
MC68HC05C9A — Rev. 5.0

Section 13. Mechanical Specifications

B.4 B.4.1 B.4.2	DC Electrical Characeristics
B.5 B.5.1 B.5.2 B.5.3 B.5.4	Control Timing
	Appendix C. Self-Check Mode
C.1	Contents
C.2	Introduction
C.3 C.3.1 C.3.2	Self-Check Mode

Appendix D. M68HC05Cx Family Feature Comparisons

MOTOROLA

Table of Contents

List of Figures

Figur	e Title	Page
1-1	Block Diagram	
1-2 1-3	Option Register	
1-4	42-Pin SDIP Pin Assignments	
1-5	44-Lead PLCC Pin Assignments	
1-6	44-Pin QFP Pin Assignments	
2-1	Memory Map	
2-2	I/O Register Summary	
2-3	Input/Output Registers	33
3-1	Programming Model	38
3-2	Interrupt Stacking Order	38
4-1	Interrupt Flowchart	45
5-1	Reset Sources	
5-2	Power-On Reset and RESET	
5-3	COP Block Diagram	
5-4	COP Reset Register (COPRST)	
5-5	COP Control Register (COPCR)	
6-1	Stop Recovery Timing Diagram	56
7-1	Port A I/O Circuit	58
7-2	Port B I/O Logic	60
8-1	Capture/Compare Timer Block Diagram	
8-2	Timer Control Register (TCR)	
8-3	Timer Status Register (TSR)	
8-4	Timer Registers (TRH and TRL)	
8-5	Alternate Timer Registers (ATRH and ATRL)	
8-6 8-7	Input Capture Registers (ICRH and ICRL) Output Compare Registers (OCRH and OCRL)	
0-1	Output Compare Registers (OCRT and OCRL)	/0

	Figure	e Title	Page
	9-1	Serial Communications Interface Block Diagram	75
	9-2	Rate Generator Division	77
	9-3	Data Format	
	9-4	SCI Examples of Start Bit Sampling Techniques	
	9-5	SCI Sampling Technique Used on All Bits	80
	9-6	SCI Artificial Start Following a Frame Error	
	9-7	SCI Start Bit Following a Break	
	9-8	SCI Data Register (SCDR)	83
	9-9	SCI Control Register 1 (SCCR1)	
	9-10	SCI Control Register 2 (SCCR2)	
	9-11	SCI Status Register (SCSR)	
	9-12	Baud Rate Register (BAUD)	89
	10-1	Data Clock Timing Diagram	93
	10-2	Serial Peripheral Interface Block Diagram	95
	10-3	Serial Peripheral Interface Master-Slave Interconnection	
	10-4	SPI Control Register (SPCR)	97
	10-5	SPI Status Register	99
	10-6	SPI Data Register (SPDR)	101
	12-1	Test Load	123
	12-2	Maximum Supply Current versus Internal	
		Clock Frequency, V _{DD} = 5.5 V	127
	12-3	Maximum Supply Current versus Internal	
		Clock Frequency, V _{DD} = 3.6 V	127
	12-4	TCAP Timing Relationships	129
	12-5	External Interrupt Timing	130
	12-6	Stop Recovery Timing Diagram	130
	12-7	Power-On Reset Timing Diagram	131
	12-8	External Reset Timing	131
	12-9	SPI Master Timing Diagram	134
	12-10	SPI Slave Timing Diagram	135
	13-1	40-Pin Plastic DIP Package (Case 711-03)	138
	13-2	42-Pin Plastic SDIP Package (Case 858-01)	138
	13-3	44-Lead PLCC (Case 777-02)	139
	13-4	44-Lead QFP (Case 824A-01)	140
	C-1	Self-Check Circuit Schematic	157
Advance Information		MC68HC05C9A -	– Rev. 5.0

List of Tables

Table	Title	Page
4-1	Vector Addresses for Interrupts and Resets	43
5-1	COP Timeout Period	53
9-1 9-2	Baud Rate Generator Clock Prescaling	
10-1	SPI Clock Rate Selection	99
11-1 11-2 11-3 11-4 11-5 11-6 11-7	Register/Memory Instructions Read-Modify-Write Instructions Jump and Branch Instructions Bit Manipulation Instructions Control Instructions Instruction Set Summary Opcode Map	108 110 111 111
14-1	MC Order Numbers	141
B-1	High-Speed Operating Temperature Range	148
C-1	Self-Check Circuit LED Codes	158
D-1	M68HC05Cx Feature Comparison	160

 $\mathsf{MC68HC05C9A} - \mathsf{Rev.}\ 5.0$

List of Tables

Section 1. General Description

1.1 Contents

1.2	Introduction
1.3	Features
1.4	Mask Options
1.5	Software-Programmable Options
1.6	Functional Pin Descriptions
1.6.1	V _{DD} and V _{SS}
1.6.2	ĪRQ26
1.6.3	OSC1 and OSC227
1.6.4	RESET27
1.6.5	TCAP
1.6.6	TCMP27
1.6.7	PA0-PA727
1.6.8	PB0-PB7
1.6.9	PC0–PC7
1.6.10	PD0–PD5 and PD7

1.2 Introduction

The MC68HC05C9A HCMOS (high-density complementary metal-oxide semiconductor) microcontroller is a member of the M68HC05 Family. The MC68HC05C9A memory map consists of 15,936 bytes of user ROM and 352 bytes of RAM. The MC68HC05C9A includes a serial communications interface, a serial peripheral interface, and a 16-bit capture/compare timer.

MC68HC05C9A — Rev. 5.0

1.3 Features

Features of the MC68HC05C9A include:

- M68HC05 CPU
- Mask programmable interrupt capability on port B
- Software programmable external interrupt sensitivity
- 15,936 bytes of read-only memory (ROM)
- 352 bytes of random-access memory (RAM)
- Memory mapped input/output (I/O)
- 31 bidirectional I/O lines with high current sink and source on PC7
- Asynchronous serial communications interface (SCI)
- Synchronous serial peripheral interface (SPI)
- 16-Bit capture/compare timer
- Computer operating properly (COP) watchdog timer and clock monitor
- Power-saving wait and stop modes
- On-chip crystal oscillator connections
- Single 3.0 volts to 5.5 volts power supply requirement
- ROM contents security⁽¹⁾ feature
- Available packages:
 - 40-pin dual in-line (DIP)
 - 44-pin plastic leaded chip carrier (PLCC)
 - 44-pin quad flat pack (QFP)
 - 42-pin plastic shrink dual in-line (SDIP) packages

1.4 Mask Options

Eight mask options are available to select external interrupt capability (including an internal pullup device) on each of the port B pins.

1. No security feature is absolutely secure. However, Motorola's strategy is to make reading or copying the ROM difficult for unauthorized users.

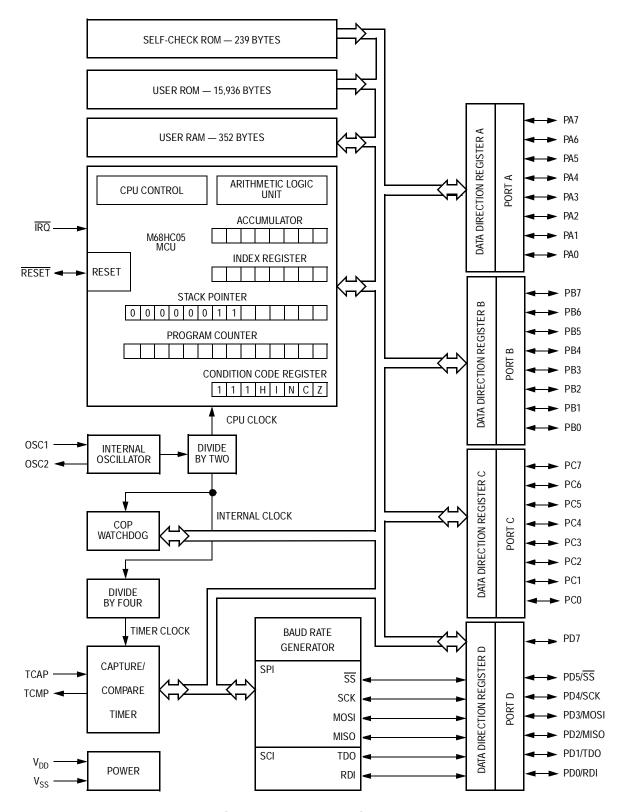


Figure 1-1. Block Diagram

MC68HC05C9A — Rev. 5.0 Advance Information

1.5 Software-Programmable Options

The option register (OR), shown in **Figure 1-2**, contains the programmable bits for these options:

- Map two different areas of memory between RAM and ROM, one of 48 bytes and one of 128 bytes
- Edge-triggered only or edge- and level-triggered external interrupt (IRQ pin and any port B pin configured for interrupt)

This register must be written to by user software during operation of the microcontroller.

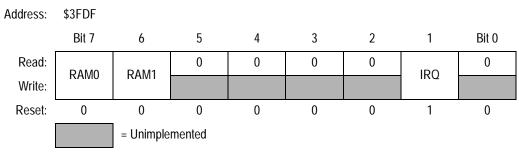


Figure 1-2. Option Register

RAM0 — Random-Access Memory Control Bit 0

This read/write bit selects between RAM or ROM in location \$0020 to \$004F. This bit can be read or written at any time.

1 = RAM selected

0 = ROM selected

RAM1— Random-Access Memory Control Bit 1

This read/write bit selects between RAM or ROM in location \$0100 to \$017F. This bit can be read or written at any time.

1 = RAM selected

0 = EPROM selected

IRQ — Interrupt Request Bit

This bit selects between an edge-triggered only or edge- and leveltriggered external interrupt. This bit is set by reset, but can be cleared by software. This bit can be written only once.

1 = Edge and level interrupt option selected

0 = Edge-only interrupt option selected

1.6 Functional Pin Descriptions

Figure 1-3, **Figure 1-4**, **Figure 1-5**, and **Figure 1-6** show the pin assignments for the available packages. A functional description of the pins follows.

NOTE: A line over a signal name indicates an active low signal. For example, RESET is active high and RESET is active low.

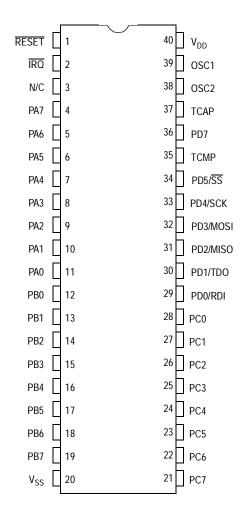


Figure 1-3. 40-Pin PDIP Pin Assignments

NOTE: If MC68HC705C9A devices are to be used in the same socket, pin 3 should be tied to V_{DD} .

MC68HC05C9A — Rev. 5.0

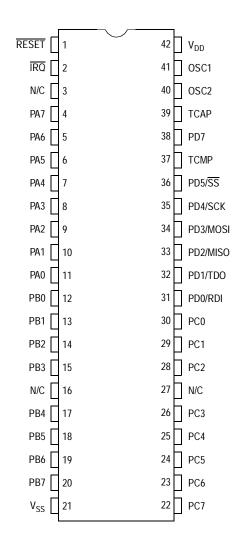


Figure 1-4. 42-Pin SDIP Pin Assignments

NOTE: If MC68HC705C9A devices are to be used in the same socket, pin 3 should be tied to V_{DD} .

Advance Information

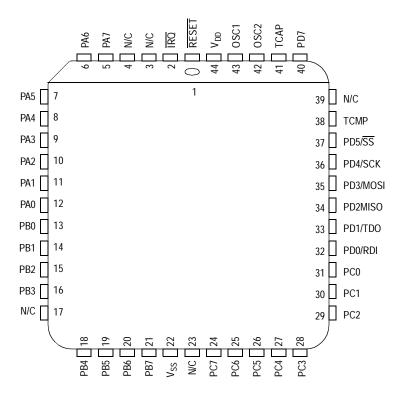


Figure 1-5. 44-Lead PLCC Pin Assignments

NOTE: The 44-pin PLCC pin assignment diagram is for compatibility with the MC68HC705C9A. However, if MC68HC705C9A devices are to be used in the same socket, pin 3 should be tied to V_{DD} .

For compatibility with MC68HC05C4A/C8A/C12A devices in 44-pin PLCC, tie pins 17 and 18 together, and tie pins 39 and 40 together.

For compatibility with MC68HC705C8A 44-pin PLCC device, three sets of pins should be tied together: pins 17 and 18, pins 39 and 40, and pins 3, 4, and 44.

MC68HC05C9A — Rev. 5.0

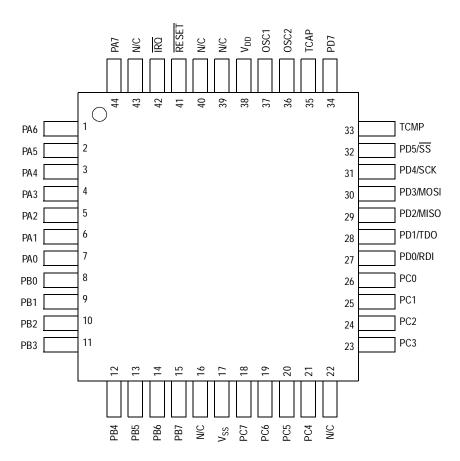


Figure 1-6. 44-Pin QFP Pin Assignments

NOTE: If MC68HC705C9A devices are to be used in the same socket, pin 43 should be tied to V_{DD} .

1.6.1 V_{DD} and V_{SS}

Power is supplied to the MCU using these two pins. V_{DD} is the positive supply and V_{SS} is ground.

1.6.2 IRQ

This interrupt pin has an option that provides two different choices of interrupt triggering sensitivity. The \overline{IRQ} pin contains an internal Schmitt trigger as part of its input to improve noise immunity. Refer to **Section 4. Interrupts** for more detail.

1.6.3 OSC1 and OSC2

These pins provide control input for an on-chip clock oscillator circuit. A crystal or ceramic resonator connected to these pins provides a system clock. The internal frequency is one-half the crystal frequency.

1.6.4 **RESET**

As an input pin, this active low RESET pin is used to reset the MCU to a known startup state by pulling RESET low. As an output pin, the RESET pin indicates that an internal MCU reset has occurred. The RESET pin contains an internal Schmitt trigger as part of its input to improve noise immunity. Refer to Section 5. Resets for more detail.

1.6.5 TCAP

This pin controls the input capture feature for the on-chip programmable timer. The TCAP pin contains an internal Schmitt trigger as part of its input to improve noise immunity. Refer to **Section 8. Capture/Compare Timer** for more detail.

1.6.6 TCMP

The TCMP pin provides an output for the output compare feature of the on-chip programmable timer. Refer to **Section 8. Capture/Compare Timer** for more detail.

1.6.7 PA0-PA7

These eight I/O lines comprise port A. The state of each pin is software programmable and all port A pins are configured as inputs during reset. Refer to **Section 7. Input/Output Ports** for more detail.

General Description

1.6.8 PB0-PB7

These eight I/O lines comprise port B. The state of each pin is software programmable and all port B pins are configured as inputs during reset. Port B has mask option register enabled pullup devices and interrupt capability selectable for any pin. Refer to **Section 7. Input/Output Ports** for more detail.

1.6.9 PC0-PC7

These eight I/O lines comprise port C. The state of each pin is software programmable and all port C pins are configured as inputs during reset. PC7 has high current sink and source capability. Refer to **Section 7**. **Input/Output Ports** for more detail.

1.6.10 PD0-PD5 and PD7

These seven I/O lines comprise port D. The state of each pin is software programmable and all port D pins are configured as inputs during reset. Refer to **Section 7. Input/Output Ports** for more detail.

Section 2. Memory

2.1 Contents

2.2	Introduction	. 29
2.3	RAM	.31
2.4	ROM	.31
2.5	ROM Security	.31
2.6	I/O Registers	.31

2.2 Introduction

The microcontroller unit (MCU) has a 16-Kbyte memory map. The memory map consists of:

- Input/output (I/O), control, and status registers
- User random-access memory (RAM)
- User read-only memory (ROM)
- Self-check ROM
- Reset and interrupt vectors

See Figure 2-1 and Figure 2-2.

Two control bits in the option register (\$3FDF) allow the user to switch between RAM and ROM at any time in two special areas of the memory map, \$0020–\$004F (48 bytes) and \$0100–\$017F (128 bytes).

MC68HC05C9A — Rev. 5.0

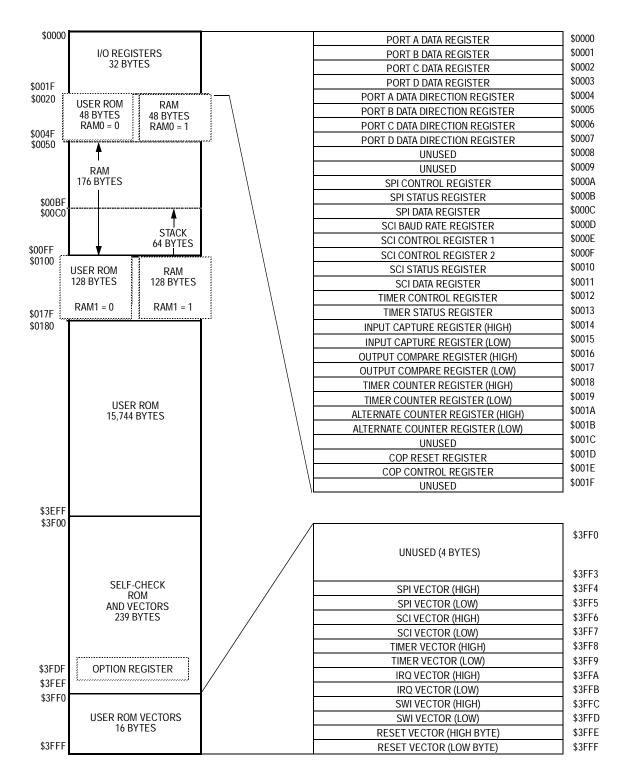


Figure 2-1. Memory Map

2.3 **RAM**

The main user RAM consists of 176 bytes at \$0050–\$00FF. This RAM area is always present in the memory map and includes a 64-byte stack area. The stack pointer can access 64 bytes of RAM in the range \$00FF down to \$00C0.

NOTE:

Using the stack area for data storage or temporary work locations requires care to prevent it from being overwritten due to stacking from an interrupt or subroutine call.

Two additional RAM areas are available at \$0020–\$004F (48 bytes) and \$0100–\$017F (128 bytes) (see **Figure 2-1** and **Figure 2-2**.) These may be accessed at any time by setting the RAM0 and RAM1 bits, respectively, in the option register.

Refer to **1.5 Software-Programmable Options** for additional information.

2.4 **ROM**

The user ROM consists of 48 bytes of page zero ROM from \$0020 to \$004F, 15,872 bytes of ROM from \$0100 to \$3EFF, and 16 bytes of user vectors from \$3FF0 to \$3FFF.

2.5 ROM Security

A security feature has been incorporated into the MC68HC05C9A to help prevent external access to the contents of the ROM in any mode of operation.

2.6 I/O Registers

Except for the option register, all I/O, control and status registers are located within one 32-byte block in page zero of the address space (\$0000–\$001F). A summary of these registers is shown in **Figure 2-2**. More detail about the contents of these registers is given in **Figure 2-3**.

Address	Register Name
\$0000	Port A Data Register
\$0001	Port B Data Register
\$0002	Port C Data Register
\$0003	Port D Data Register
\$0004	Port A Data Direction Register
\$0005	Port B Data Direction Register
\$0006	Port C Data Direction Register
\$0007	Port D Data Direction Register
\$0008	Unused
\$0009	Unused
\$000A	Serial Peripheral Control Register
\$000B	Serial Peripheral Status Register
\$000C	Serial Peripheral Data Register
\$000D	Baud Rate Register
\$000E	Serial Communications Control Register 1
\$000F	Serial Communications Control Register 2
\$0010	Serial Communications Status Register
\$0011	Serial Communications Data Register
\$0012	Timer Control Register
\$0013	Timer Status Register
\$0014	Input Capture Register High
\$0015	Input Capture Register Low
\$0016	Output Compare Register High
\$0017	Output Compare Register Low
\$0018	Timer Register High
\$0019	Timer Register Low
\$001A	Alternate Timer Register High
\$001B	Alternate Timer Register Low
\$001C	Unused
\$001D	COP Reset Register
\$001E	COP Control Register
\$001F	Reserved

Figure 2-2. I/O Register Summary

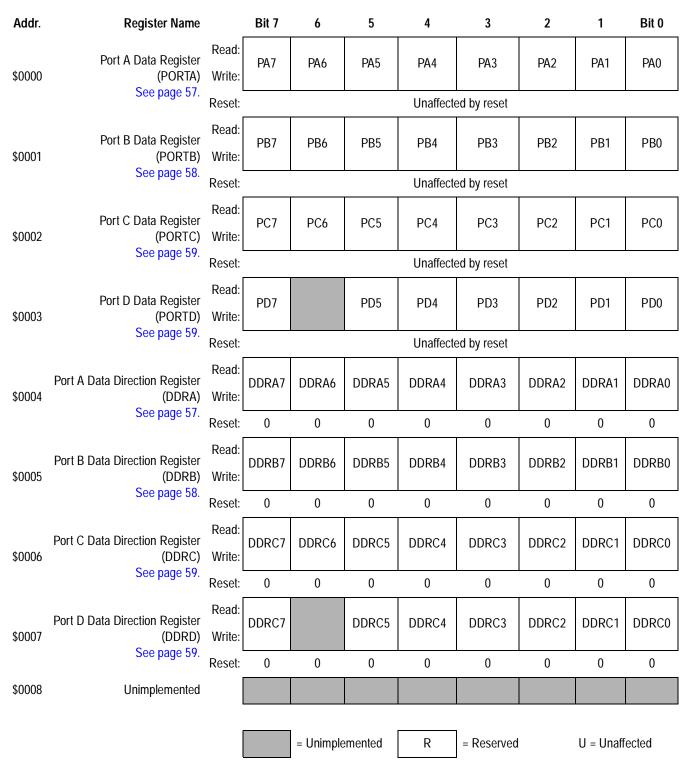


Figure 2-3. Input/Output Registers (Sheet 1 of 4)

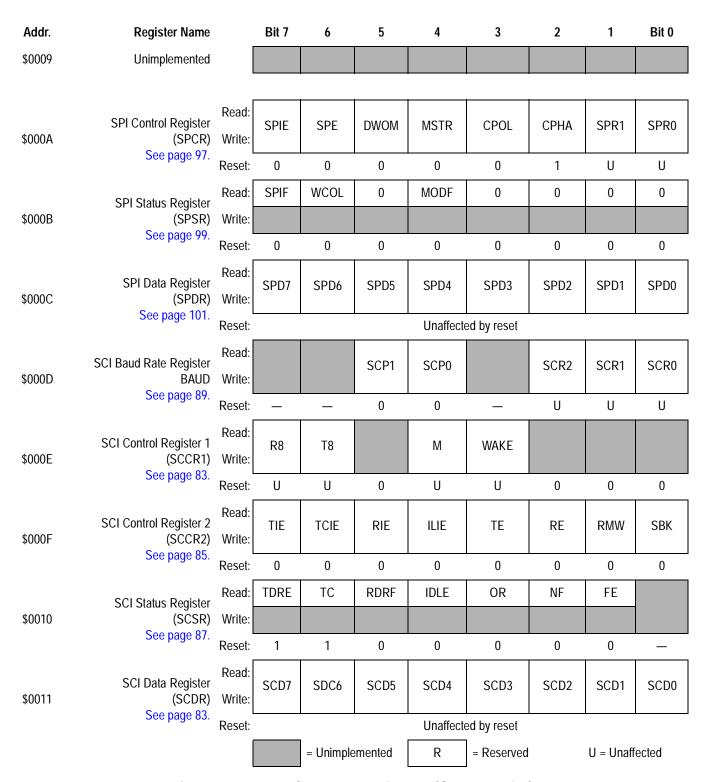


Figure 2-3. Input/Output Registers (Sheet 2 of 4)

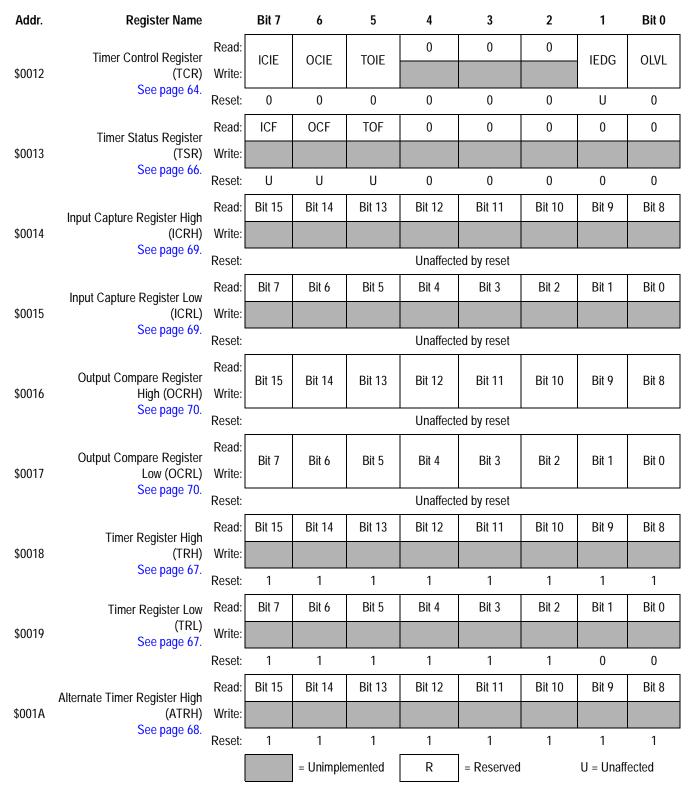


Figure 2-3. Input/Output Registers (Sheet 3 of 4)

MC68HC05C9A — Rev. 5.0 Advance Information

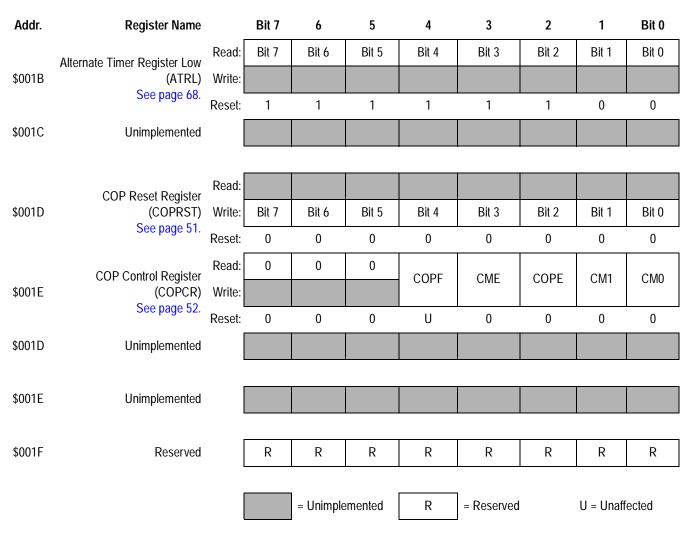


Figure 2-3. Input/Output Registers (Sheet 4 of 4)

Section 3. Central Processor Unit (CPU)

3.1 Contents

3.2	Introduction
3.3	CPU Registers
3.3.1	Accumulator (A)39
3.3.2	Index Register (X)
3.3.3	Program Counter (PC)39
3.3.4	Stack Pointer (SP)39
3.3.5	Condition Code Register (CCR)

3.2 Introduction

This section contains information describing the basic programmer's model and the registers contained in the central processor unit (CPU).

MC68HC05C9A — Rev. 5.0

3.3 CPU Registers

The microcontroller unit (MCU) contains five registers as shown in the programming model of **Figure 3-1**. The interrupt stacking order is shown in **Figure 3-2**.

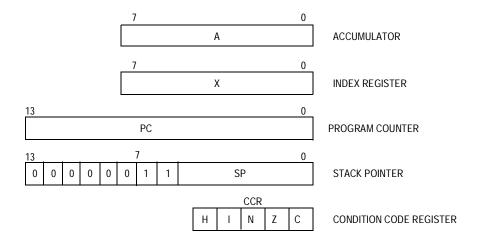


Figure 3-1. Programming Model

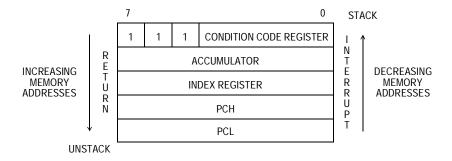


Figure 3-2. Interrupt Stacking Order

3.3.1 Accumulator (A)

The accumulator is a general-purpose 8-bit register used to hold operands and results of arithmetic calculations or data manipulations.

3.3.2 Index Register (X)

The index register is an 8-bit register used for the indexed addressing value to create an effective address. The index register may also be used as a temporary storage area.

3.3.3 Program Counter (PC)

The program counter is a 14-bit register that contains the address of the next byte to be fetched.

3.3.4 Stack Pointer (SP)

The stack pointer contains the address of the next free location on the stack. During an MCU reset or the reset stack pointer (RSP) instruction, the stack pointer is set to location \$0FF. The stack pointer is then decremented as data is pushed onto the stack and incremented as data is pulled from the stack.

When accessing memory, the eight most significant bits are permanently set to 00000011. These eight bits are appended to the six least significant register bits to produce an address within the range of \$00FF to \$00C0. Subroutines and interrupts may use up to 64 (decimal) locations. If 64 locations are exceeded, the stack pointer wraps around and loses the previously stored information. A subroutine call occupies two locations on the stack; an interrupt uses five locations.

Central Processor Unit (CPU)

3.3.5 Condition Code Register (CCR)

The CCR is a 5-bit register in which four bits are used to indicate the results of the instruction just executed, and the fifth bit indicates whether interrupts are masked. These bits can be individually tested by a program, and specific actions can be taken as a result of their state. Each bit is explained here.

Half Carry (H)

This bit is set during ADD and ADC operations to indicate that a carry occurred between bits 3 and 4.

Interrupt (I)

When this bit is set, the timer, serial communications interface (SCI), serial peripheral interface (SPI), and external interrupt are masked (disabled). If an interrupt occurs while this bit is set, the interrupt is latched and processed as soon as the interrupt bit is cleared.

Negative (N)

When set, this bit indicates that the result of the last arithmetic, logical, or data manipulation was negative.

Zero (Z)

When set, this bit indicates that the result of the last arithmetic, logical, or data manipulation was 0.

Carry/Borrow (C)

When set, this bit indicates that a carry or borrow out of the arithmetic logical unit (ALU) occurred during the last arithmetic operation. This bit is also affected during bit test and branch instructions and during shifts and rotates.

Section 4. Interrupts

4.1 Contents

4.2	Introduction	.41
4.3	Non-Maskable Software Interrupt (SWI)	. 42
4.4	External Interrupt (IRQ or Port B)	. 42
4.5	Timer Interrupt	.44
4.6	SCI Interrupt	.44
4.7	SPI Interrupt	.44

4.2 Introduction

The MC68HC05C9A microcontroller unit (MCU) can be interrupted by five different sources: four maskable hardware interrupts, and one non-maskable software interrupt:

- External signal on the IRQ pin or port B pins
- 16-bit programmable timer
- Serial communications interface (SCI)
- Serial peripheral interface (SPI)
- Software interrupt instruction (SWI)

Interrupts cause the processor to save register contents on the stack and to set the interrupt mask (I bit) to prevent additional interrupts. The return from interrupt (RTI) instruction causes the register contents to be recovered from the stack and normal processing to resume.

MC68HC05C9A — Rev. 5.0

Unlike RESET, hardware interrupts do not cause the current instruction execution to be halted, but are considered pending until the current instruction is complete.

NOTE: The current instruction is the one already fetched and being operated on.

When the current instruction is complete, the processor checks all pending hardware interrupts. If interrupts are not masked (CCR I bit clear) and if the corresponding interrupt enable bit is set, the processor proceeds with interrupt processing; otherwise, the next instruction is fetched and executed.

If an external interrupt and a timer, SCI, or SPI interrupt are pending at the end of an instruction execution, the external interrupt is serviced first. The SWI is executed the same as any other instruction, regardless of the I-bit state.

Table 4-1 shows the relative priority of all the possible interrupt sources. **Figure 4-1** shows the interrupt processing flow.

4.3 Non-Maskable Software Interrupt (SWI)

The SWI is an executable instruction and a non-maskable interrupt. It is executed regardless of the state of the I bit in the CCR. If the I bit is 0 (interrupts enabled), SWI executes after interrupts which were pending when the SWI was fetched, but before interrupts generated after the SWI was fetched. The interrupt service routine address is specified by the contents of memory locations \$3FFC and \$3FFD.

4.4 External Interrupt (IRQ or Port B)

If the interrupt mask bit (I bit) of the CCR is set, all maskable interrupts (internal and external) are disabled. Clearing the I bit enables interrupts. The interrupt request is latched immediately following the falling edge of IRQ. It is then synchronized internally and serviced as specified by the contents of \$3FFA and \$3FFB.

Table 4-1. Vector Addresses for Interrupts and Resets

Function	Source	Local Mask	Global Mask	Priority (1 = Highest)	Vector Address
	Power-on reset	None	None	1	\$3FFE-\$3FFF
Reset	RESET pin				
	COP watchdog				
Software interrupt (SWI)	User code	None	None	Same priority as instruction	\$3FFC-\$3FFD
External interrupt	IRQ pin	- None	l bit	2	\$3FFA-\$3FFB
External interrupt	Port B pins				
	ICF bit	ICIE bit	I bit	3	\$3FF8-\$3FF9
Timer interrupts	OCF bit	OCIE bit			
	TOF bit	TOIE bit			
	TDRE bit	TCIE bit	l bit	4	\$3FF6-\$3FF7
	TC bit				
SCI interrupts	RDRF bit	RIE bit			
	OR bit	NIE DIL			
	IDLE bit	ILIE bit			
CDI interrupte	SPIF bit	SPIE bit	l bit	5	\$3FF4-\$3FF5
SPI interrupts	MODF bit	SFIE VIL			

When any of the port B pullups are enabled, each pin becomes an additional external interrupt source which is executed identically to the \overline{IRQ} pin. Port B interrupts follow the same edge/edge-level selection as the \overline{IRQ} pin. The branch instructions BIL and BIH also respond to the port B interrupts in the same way as the \overline{IRQ} pin. See **7.4 Port B**.

Either a level-sensitive and edge-sensitive trigger or an edge-sensitiveonly trigger operation is selectable. The sensitivity is software-controlled by the IRQ bit in the option register (\$3FDF).

NOTE:

The internal interrupt latch is cleared in the first part of the interrupt service routine; therefore, one external interrupt pulse can be latched and serviced as soon as the I bit is cleared.

MC68HC05C9A — Rev. 5.0

4.5 Timer Interrupt

Three different timer interrupt flags cause a timer interrupt whenever they are set and enabled. The interrupt flags are in the timer status register (TSR), and the enable bits are in the timer control register (TCR). Any of these interrupts will vector to the same interrupt service routine, located at the address specified by the contents of memory locations \$3FF8 and \$3FF9.

4.6 SCI Interrupt

Five different SCI interrupt flags cause an SCI interrupt whenever they are set and enabled. The interrupt flags are in the SCI status register (SCSR), and the enable bits are in the SCI control register 2 (SCCR2). Any of these interrupts will vector to the same interrupt service routine, located at the address specified by the contents of memory locations \$3FF6 and \$3FF7.

4.7 SPI Interrupt

Two different SPI interrupt flags cause an SPI interrupt whenever they are set and enabled. The interrupt flags are in the SPI status register (SPSR), and the enable bits are in the SPI control register (SPCR). Either of these interrupts will vector to the same interrupt service routine, located at the address specified by the contents of memory locations \$3FF4 and \$3FF5.

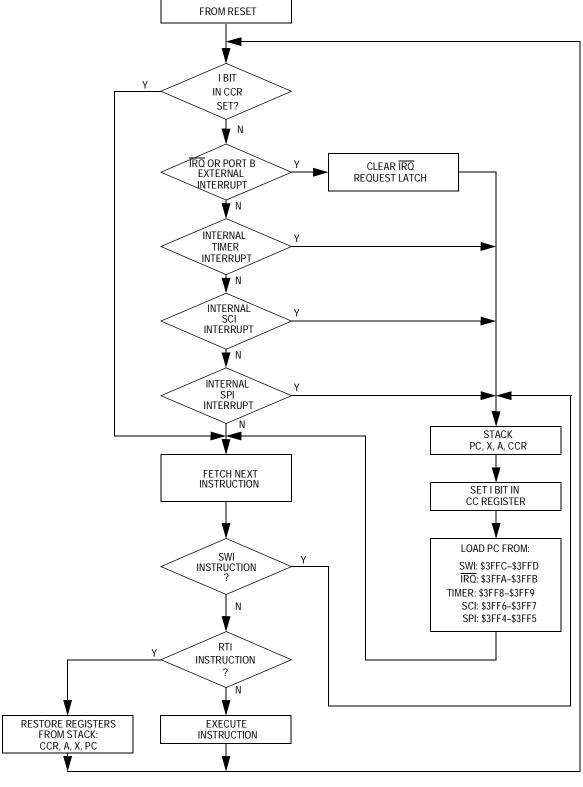


Figure 4-1. Interrupt Flowchart

Interrupts

Section 5. Resets

5.1 Contents

5.2	Introduction
5.3	Power-On Reset (POR)
5.4	RESET Pin
	Computer Operating Properly (COP) Reset .50 COP Reset Register .51 COP Control Register .52
5.6	COP During Wait Mode54
5.7	COP During Stop Mode
5.8	Clock Monitor Reset

5.2 Introduction

The MC68HC05C9A microcontroller unit (MCU) can be reset four ways:

- Initial power-on reset function
- Active low input to the RESET pin
- Computer operating properly (COP)
- Clock monitor

A reset immediately stops the operation of the instruction being executed, initializes some control bits, and loads the program counter with a user-defined reset vector address. **Figure 5-1** is a block diagram of the reset sources.

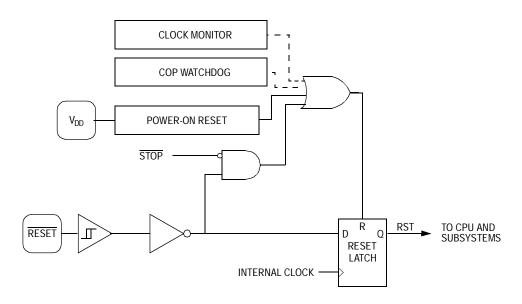


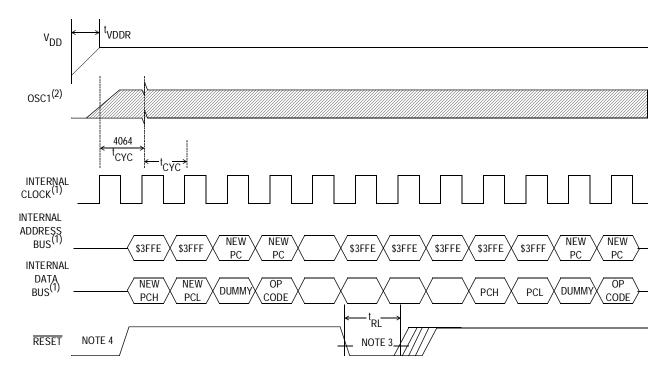
Figure 5-1. Reset Sources

5.3 Power-On Reset (POR)

A power-on reset (POR) occurs when a positive transition is detected on V_{DD} . The power-on reset is strictly for power turn-on conditions and should not be used to detect a drop in the power supply voltage. There is a 4064 internal processor clock cycle (t_{CYC}) oscillator stabilization delay after the oscillator becomes active. The RESET pin will output a logic 0 during the 4064-cycle delay. If the RESET pin is low after the end of this 4064-cycle delay, the MCU will remain in the reset condition until RESET is driven high externally.

5.4 RESET Pin

The MCU is reset when a logic 0 is applied to the $\overline{\text{RESET}}$ input for a period of one and one-half machine cycles (t_{RL}). However, to differentiate between an external reset and an internal reset (generated from the COP or clock monitor), any externally driven reset must be active (logic 0) for at least eight t_{cvc} .



Notes:

- 1. Internal timing signal and bus information are not available externally.
- 2. OSC1 line is not meant to represent frequency. It is meant to represent only time.
- 3. The next rising edge of the internal processor clock following the rising edge of RESET initiates the reset sequence.
- 4. RESET outputs V_{OL} during 4064 power-on reset cycles.

Figure 5-2. Power-On Reset and RESET

5.5 Computer Operating Properly (COP) Reset

This device includes a watchdog COP feature which guards against program run-away failures. A timeout of the COP timer generates a COP reset. The COP watchdog is a software error detection system that automatically times out and resets the MCU if not cleared periodically by a program sequence.

The COP is controlled with two registers, one to reset the COP timer and the other to enable and control COP and clock monitor functions.

Figure 5-3 shows a block diagram of the COP.

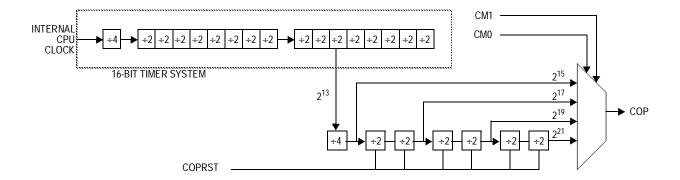


Figure 5-3. COP Block Diagram

5.5.1 COP Reset Register

The COP reset register (COPRST), shown in **Figure 5-4**, is a write-only register used to reset the COP.

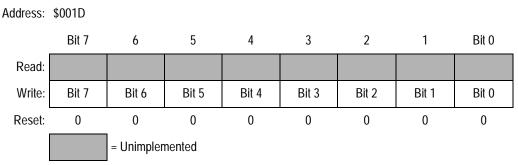


Figure 5-4. COP Reset Register (COPRST)

The sequence required to reset the COP timer is:

- Write \$55 to the COP reset register
- Write \$AA to the COP reset register

Both write operations must occur in the order listed, but any number of instructions may be executed between the two write operations provided that the COP does not time out between the two writes. The elapsed time between software resets must not be greater than the COP timeout period. If the COP should time out, a system reset will occur and the device will be re-initialized in the same fashion as a power-on reset or reset.

Reading this register does not return valid data.

5.5.2 COP Control Register

The COP control register (COPCR), shown in **Figure 5-5**, performs these functions:

- Enables clock monitor function
- Enables COP function
- Selects timeout duration of COP timer

And flags these conditions:

- COP timeout
- Clock monitor reset

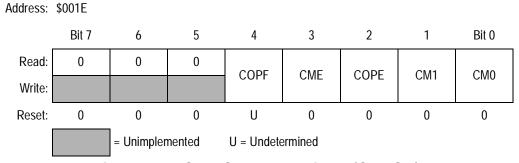


Figure 5-5. COP Control Register (COPCR)

COPF — Computer Operating Properly Flag

Reading the COP control register clears COPF.

- 1 = COP or clock monitor reset has occurred.
- 0 = No COP or clock monitor reset has occurred.

CME — Clock Monitor Enable Bit

This bit is readable any time, but may be written only once.

- 1 = Clock monitor enabled
- 0 = Clock monitor disabled

COPE — COP Enable Bit

This bit is readable any time. COPE, CM1, and CM0 together may be written with a single write, only once, after reset. This bit is cleared by reset.

1 = COP enabled

0 = COP disabled

CM1 — COP Mode Bit 1

Used in conjunction with CM0 to establish the COP timeout period, this bit is readable any time. COPE, CM1, and CM0 together may be written with a single write, only once, after reset. This bit is cleared by reset. See **Table 5-1** for timeout period options.

CM0 — COP Mode Bit 0

Used in conjunction with CM1 to establish the COP timeout period, this bit is readable any time. COPE, CM1, and CM0 together may be written with a single write, only once, after reset. This bit is cleared by reset. See **Table 5-1** for timeout period options.

Bits 7-5 — Not Used

These bits always read as 0.

Table 5-1. COP Timeout Period

CM1	СМО	f _{OP} /2 ¹⁵ Divide By	Timeout Period (f _{OSC} = 2.0 MHz)	Timeout Period (f _{OSC} = 4.0 MHz)
0	0	1	32.77 ms	16.38 ms
0	1	4	131.07 ms	65.54 ms
1	0	16	524.29 ms	262.14 ms
1	1	64	2.097 s	1.048 s

5.6 COP During Wait Mode

The COP will continue to operate normally during wait mode. The software must pull the device out of wait mode periodically and reset the COP to prevent a system reset.

5.7 COP During Stop Mode

Stop mode disables the oscillator circuit and thereby turns the clock off for the entire device. The COP counter will be reset when stop mode is entered. If a reset is used to exit stop mode, the COP counter will be reset after the 4064 cycles of delay after stop mode. If an IRQ is used to exit stop mode, the COP counter will not be reset after the 4064-cycle delay and will have that many cycles already counted when control is returned to the program.

In the event that an inadvertent STOP instruction is executed, the COP will not provide a reset. The clock monitor function provides protection for this situation.

5.8 Clock Monitor Reset

The clock monitor circuit can provide a system reset if the clock stops for any reason, including stop mode. When the CME bit in the COP control register is set, the clock monitor detects the absence of the internal bus clock for a certain period of time. The timeout period is dependent on the processing parameters and varies from 5 μs to 100 μs , which implies that systems using a bus clock rate of 200 kHz or less should not use the clock monitor.

If a slow or absent clock is detected, the clock monitor causes a system reset. The reset is issued to the external system via the bidirectional RESET pin for four bus cycles if the clock is slow or until the clocks recover in the case where the clocks are absent.

Section 6. Low-Power Modes

6.1 Contents

6.2	Introduction	55
6.3	Stop Mode	55
6 4	Wait Mode	56

6.2 Introduction

This section describes the low-power stop and wait modes.

6.3 Stop Mode

The STOP instruction places the microcontroller unit (MCU) in its lowest-power consumption mode. In stop mode, the internal oscillator is turned off, halting all internal processing, including timer operation.

During stop mode, the TCR bits are altered to remove any pending timer interrupt request and to disable any further timer interrupts. The timer prescaler is cleared. The I bit in the condition code register (CCR) is cleared to enable external interrupts. All other registers and memory remain unaltered. All input/output (I/O) lines remain unchanged. The processor can be brought out of stop mode only by an external interrupt or reset. See **Figure 6-1**.

MC68HC05C9A — Rev. 5.0

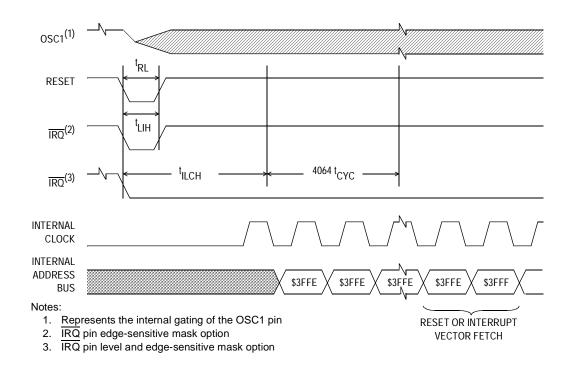


Figure 6-1. Stop Recovery Timing Diagram

6.4 Wait Mode

The WAIT instruction places the MCU in a low-power consumption mode, but wait mode consumes more power than stop mode. All central processor unit (CPU) action is suspended, but the timer, serial communications interface (SCI), serial peripheral interface (SPI), and the oscillator remain active. Any interrupt or reset will cause the MCU to exit wait mode.

During wait mode, the I bit in the CCR is cleared to enable interrupts. All other registers, memory, and I/O lines remain in their previous state. The timer, SCI, and SPI may be enabled to allow a periodic exit from the wait mode.

Advance Information

MOTOROLA

Section 7. Input/Output Ports

7.1 Contents

7.2	Introduction
7.3	Port A
7.4	Port B
7.5	Port C
7.6	Port D

7.2 Introduction

This section briefly describes the 31 input/output (I/O) lines arranged as one 7-bit and three 8-bit ports. All of these port pins are programmable as either inputs or outputs under software control of the data direction registers.

NOTE:

To avoid a glitch on the output pins, write data to the I/O port data register before writing a 1 to the corresponding data direction register.

7.3 Port A

Port A is an 8-bit bidirectional port which does not share any of its pins with other subsystems. The port A data register is at \$0000 and the data direction register (DDR) is at \$0004. The contents of the port A data register are indeterminate at initial power-up and must be initialized by user software. Reset does not affect the data registers, but clears the data direction registers, thereby returning the ports to inputs. Writing a 1 to a DDR bit sets the corresponding port bit to output mode. A block diagram of the port logic is shown in **Figure 7-1**.

MC68HC05C9A — Rev. 5.0

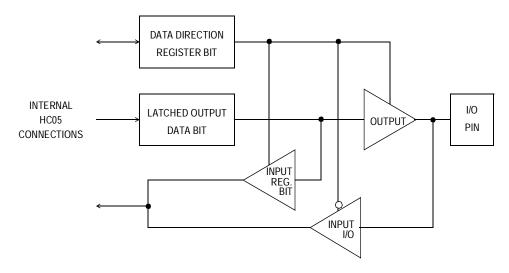


Figure 7-1. Port A I/O Circuit

7.4 Port B

Port B is an 8-bit bidirectional port. The port B data register is at \$0001 and the data direction register (DDR) is at \$0005. The contents of the port B data register are indeterminate at initial powerup and must be initialized by user software. Reset does not affect the data registers, but clears the data direction registers, thereby returning the ports to inputs. Writing a one to a DDR bit sets the corresponding port pin to output mode. Each of the port B pins has an optional external interrupt capability that can be enabled by mask option.

The interrupt option also enables a pullup device when the pin is configured as an input. The edge or edge- and level-sensitivity of the \overline{IRQ} pin will also pertain to the enabled port B pins. Care needs to be taken when using port B pins that have the pullup enabled. Before switching from an output to an input, the data should be preconditioned to a 1 to prevent an interrupt from occurring. The port B logic is shown in Figure 7-2.

7.5 Port C

Port C is an 8-bit bidirectional port. The port C data register is at \$0002 and the data direction register (DDR) is at \$0006. The contents of the port C data register are indeterminate at initial powerup and must be initialized by user software. Reset does not affect the data registers, but clears the data direction registers, thereby returning the ports to inputs. Writing a 1 to a DDR bit sets the corresponding port bit to output mode. PC7 has a high current sink and source capability. Figure 7-1 is also applicable to port C.

7.6 Port D

Port D is a 7-bit bidirectional port. Four of its pins are shared with the SPI subsystem and two more are shared with the SCI subsystem. The port D data register is at \$0003 and the data direction register is at \$0007. The contents of the port D data register are indeterminate at initial powerup and must be initialized by user software. During reset all seven bits become valid input ports because the DDR bits are cleared and the special function output drivers associated with the SCI and SPI subsystems are disabled, thereby returning the ports to inputs. Writing a 1 to a DDR bit sets the corresponding port bit to output mode.

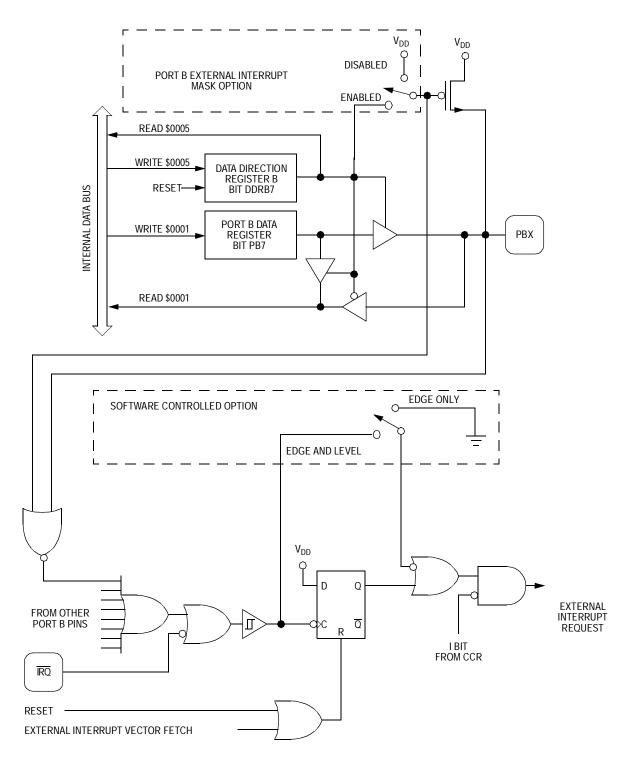


Figure 7-2. Port B I/O Logic

Section 8. Capture/Compare Timer

8.1 Content

8.2	Introduction
8.3	Timer Operation
8.3.1	Input Capture
8.3.2	Output Compare
8.4	Timer I/O Registers64
8.4.1	Timer Control Register
8.4.2	Timer Status Register
8.4.3	Timer Registers
8.4.4	Alternate Timer Registers
8.4.5	Input Capture Registers69
8.4.6	Output Compare Registers
8.5	Timer During Wait Mode71
8.6	Timer During Stop Mode71

8.2 Introduction

This section describes the operation of the 16-bit capture/compare timer. **Figure 8-1** shows the structure of the capture/compare subsystem.

MC68HC05C9A — Rev. 5.0

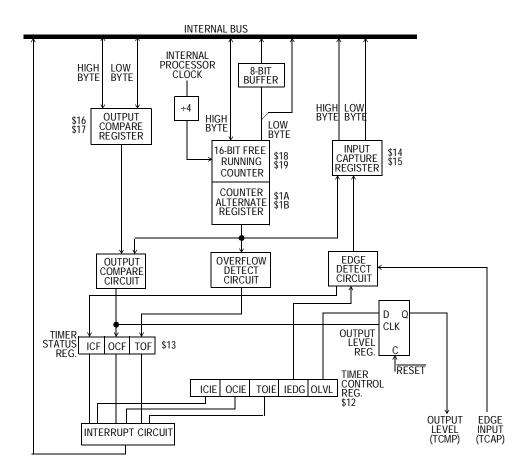


Figure 8-1. Capture/Compare Timer Block Diagram

8.3 Timer Operation

The core of the capture/compare timer is a 16-bit free-running counter. The counter provides the timing reference for the input capture and output compare functions. The input capture and output compare functions provide a means to latch the times at which external events occur, to measure input waveforms, and to generate output waveforms and timing delays. Software can read the value in the 16-bit free-running counter at any time without affecting the counter sequence.

Because of the 16-bit timer architecture, the input/output (I/O) registers for the input capture and output compare functions are pairs of 8-bit registers.

Because the counter is 16 bits long and preceded by a fixed divide-by-4 prescaler, the counter rolls over every 262,144 internal clock cycles. Timer resolution with a 4-MHz crystal is $2 \mu s$.

8.3.1 Input Capture

The input capture function is a means to record the time at which an external event occurs. When the input capture circuitry detects an active edge on the TCAP pin, it latches the contents of the timer registers into the input capture registers. The polarity of the active edge is programmable.

Latching values into the input capture registers at successive edges of the same polarity measures the period of the input signal on the TCAP pin. Latching values into the input capture registers at successive edges of opposite polarity measures the pulse width of the signal.

8.3.2 Output Compare

The output compare function is a means of generating an output signal when the 16-bit counter reaches a selected value. Software writes the selected value into the output compare registers. On every fourth internal clock cycle the output compare circuitry compares the value of the counter to the value written in the output compare registers. When a match occurs, the timer transfers the programmable output level bit (OLVL) from the timer control register to the TCMP pin.

The programmer can use the output compare register to measure time periods, to generate timing delays, or to generate a pulse of specific duration or a pulse train of specific frequency and duty cycle on the TCMP pin.

8.4 Timer I/O Registers

These I/O registers control and monitor timer operation:

- Timer control register (TCR)
- Timer status register (TSR)
- Timer registers (TRH and TRL)
- Alternate timer registers (ATRH and ATRL)
- Input capture registers (ICRH and ICRL)
- Output compare registers (OCRH and OCRL)

8.4.1 Timer Control Register

The timer control register (TCR), shown in **Figure 8-2**, performs these functions:

- Enables input capture interrupts
- Enables output compare interrupts
- Enables timer overflow interrupts
- Controls the active edge polarity of the TCAP signal
- Controls the active level of the TCMP output

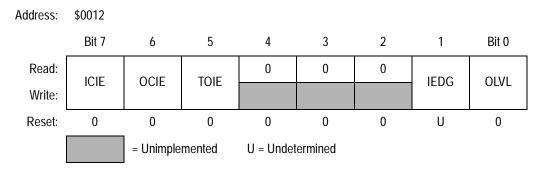


Figure 8-2. Timer Control Register (TCR)

ICIE — Input Capture Interrupt Enable Bit

This read/write bit enables interrupts caused by an active signal on the TCAP pin. Reset clears the ICIE bit.

- 1 = Input capture interrupts enabled
- 0 = Input capture interrupts disabled

OCIE — Output Compare Interrupt Enable Bit

This read/write bit enables interrupts caused by an active signal on the TCMP pin. Reset clears the OCIE bit.

- 1 = Output compare interrupts enabled
- 0 = Output compare interrupts disabled

TOIE — Timer Overflow Interrupt Enable Bit

This read/write bit enables interrupts caused by a timer overflow. Reset clears the TOIE bit.

- 1 = Timer overflow interrupts enabled
- 0 = Timer overflow interrupts disabled

IEDG — Input Edge Bit

The state of this read/write bit determines whether a positive or negative transition on the TCAP pin triggers a transfer of the contents of the timer register to the input capture register. Resets have no effect on the IEDG bit.

- 1 = Positive edge (low to high transition) triggers input capture.
- 0 = Negative edge (high to low transition) triggers input capture.

OLVL — Output Level Bit

The state of this read/write bit determines whether a logic 1 or logic 0 appears on the TCMP pin when a successful output compare occurs. Reset clears the OLVL bit.

- 1 = TCMP goes high on output compare.
- 0 = TCMP goes low on output compare.

8.4.2 Timer Status Register

The timer status register (TSR), shown in **Figure 8-3**, contains flags to signal these conditions:

- An active signal on the TCAP pin, transferring the contents of the timer registers to the input capture registers
- A match between the 16-bit counter and the output compare registers, transferring the OLVL bit to the TCMP pin
- A timer roll over from \$FFFF to \$0000

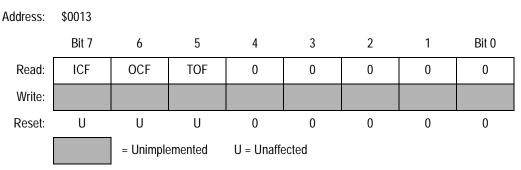


Figure 8-3. Timer Status Register (TSR)

ICF — Input Capture Flag

The ICF bit is set automatically when an edge of the selected polarity occurs on the TCAP pin. Clear the ICF bit by reading the timer status register with ICF set and then reading the low byte (\$0015) of the input capture registers. Resets have no effect on ICF.

OCF — Output Compare Flag

The OCF bit is set automatically when the value of the timer registers matches the contents of the output compare registers. Clear the OCF bit by reading the timer status register with OCF set and then reading the low byte (\$0017) of the output compare registers. Resets have no effect on OCF.

TOF — Timer Overflow Flag

The TOF bit is set automatically when the 16-bit counter rolls over from \$FFFF to \$0000. Clear the TOF bit by reading the timer status register with TOF set, and then reading the low byte (\$0019) of the timer registers. Resets have no effect on TOF.

8.4.3 Timer Registers

The timer registers (TRH and TRL), shown in **Figure 8-4**, contain the current high and low bytes of the 16-bit counter. Reading TRH before reading TRL causes TRL to be latched until TRL is read. Reading TRL after reading the timer status register clears the timer overflow flag (TOF). Writing to the timer registers has no effect.

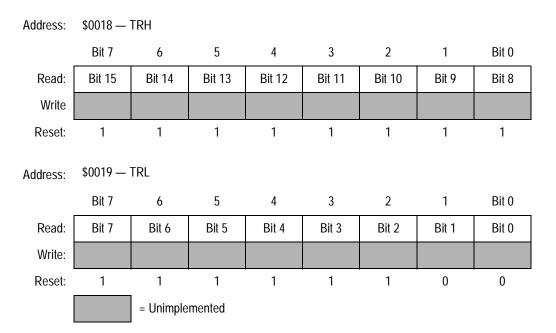


Figure 8-4. Timer Registers (TRH and TRL)

67

8.4.4 Alternate Timer Registers

The alternate timer registers (ATRH and ATRL), shown in **Figure 8-5**, contain the current high and low bytes of the 16-bit counter. Reading ATRH before reading ATRL causes ATRL to be latched until ATRL is read. Reading ATRL has no effect on the timer overflow flag (TOF). Writing to the alternate timer registers has no effect.

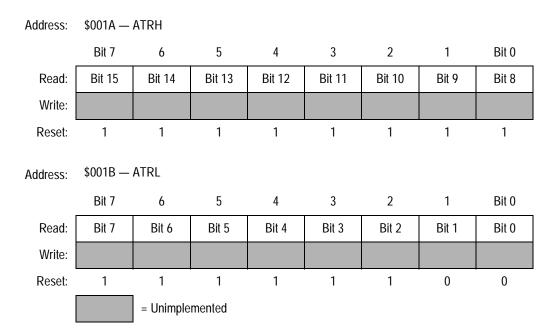


Figure 8-5. Alternate Timer Registers (ATRH and ATRL)

NOTE: To prevent interrupts from occurring between readings of ATRH and ATRL, set the interrupt flag in the condition code register before reading ATRH, and clear the flag after reading ATRL.

8.4.5 Input Capture Registers

When a selected edge occurs on the TCAP pin, the current high and low bytes of the 16-bit counter are latched into the input capture registers. Reading ICRH before reading ICRL inhibits further capture until ICRL is read. Reading ICRL after reading the status register clears the input capture flag (ICF). Writing to the input capture registers has no effect.

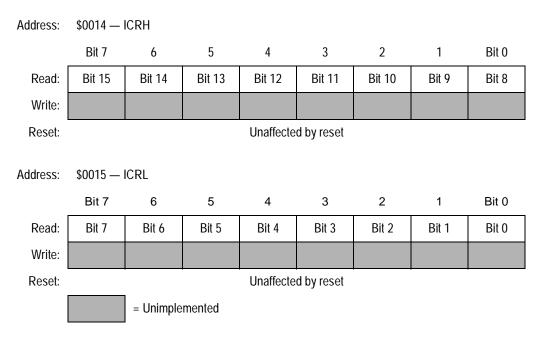


Figure 8-6. Input Capture Registers (ICRH and ICRL)

NOTE: To prevent interrupts from occurring between readings of ICRH and ICRL, set the interrupt flag in the condition code register before reading ICRH, and clear the flag after reading ICRL.

8.4.6 Output Compare Registers

When the value of the 16-bit counter matches the value in the output compare registers, the planned TCMP pin action takes place. Writing to OCRH before writing to OCRL inhibits timer compares until OCRL is written. Reading or writing to OCRL after the timer status register clears the output compare flag (OCF).

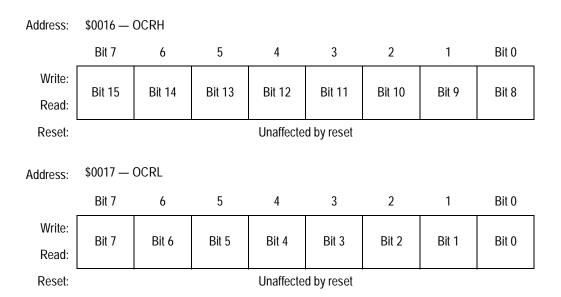


Figure 8-7. Output Compare Registers (OCRH and OCRL)

To prevent OCF from being set between the time it is read and the time the output compare registers are updated, use this procedure:

- Disable interrupts by setting the I bit in the CCR.
- 2. Write to OCRH. Compares are now inhibited until OCRL is written.
- 3. Clear bit OCF by reading timer status register (TSR).
- 4. Enable the output compare function by writing to OCRL.
- 5. Enable interrupts by clearing the I bit in the CCR.

8.5 Timer During Wait Mode

The central processor unit (CPU) clock halts during wait mode, but the timer remains active. If interrupts are enabled, a timer interrupt will cause the processor to exit wait mode.

8.6 Timer During Stop Mode

In stop mode, the timer stops counting and holds the last count value if STOP is exited by an interrupt. If STOP is exited by reset, the counters are forced to \$FFFC. During STOP, if at least one valid input capture edge occurs at the TCAP pins, the input capture detect circuit is armed. This does not set any timer flags or wake up the microcontroller unit (MCU). But if an interrupt is used to exit stop mode, there is an active input capture flag and data from the first valid edge that occurred during the stop mode. If reset is used to exit stop mode, then no input capture flag or data remains, even if a valid input capture edge occurred.

Capture/Compare Timer

Section 9. Serial Communications Interface (SCI)

9.1 Content

9.2	Introduction
9.3	Features
9.4	SCI Receiver Features
9.5	SCI Transmitter Features
9.6	Functional Description
9.7	Data Format
9.8	Receiver Wakeup Operation
9.8.1	Idle Line Wakeup
9.8.2	Address Mark Wakeup
9.9	Receive Data In (RDI)
9.10	Start Bit Detection
9.11	Transmit Data Out (TDO)
9.12	SCI I/O Registers
9.12.1	SCI Data Register
9.12.2	SCI Control Register 1
9.12.3	SCI Control Register 2
9.12.4	SCI Status Register87
9.12.5	Baud Rate Register89

MC68HC05C9A — Rev. 5.0

9.2 Introduction

This section describes the on-chip asynchronous serial communications interface (SCI). The SCI allows full-duplex, asynchronous, RS232 or RS422 serial communication between the microcontroller unit (MCU) and remote devices, including other MCUs. The transmitter and receiver of the SCI operate independently, although they use the same baud rate generator.

9.3 Features

Features of the SCI include:

- Standard mark/space non-return-to-zero format
- Full-duplex operation
- 32 programmable baud rates
- Programmable 8-bit or 9-bit character length
- Separately enabled transmitter and receiver
- Two receiver wakeup methods:
 - Idle line wakeup
 - Address mark wakeup
- Interrupt-driven operation capability with five interrupt flags:
 - Transmitter data register empty
 - Transmission complete
 - Transmission data register full
 - Receiver overrun
 - Idle receiver input
- Receiver framing error detection
- 1/16 bit-time noise detection

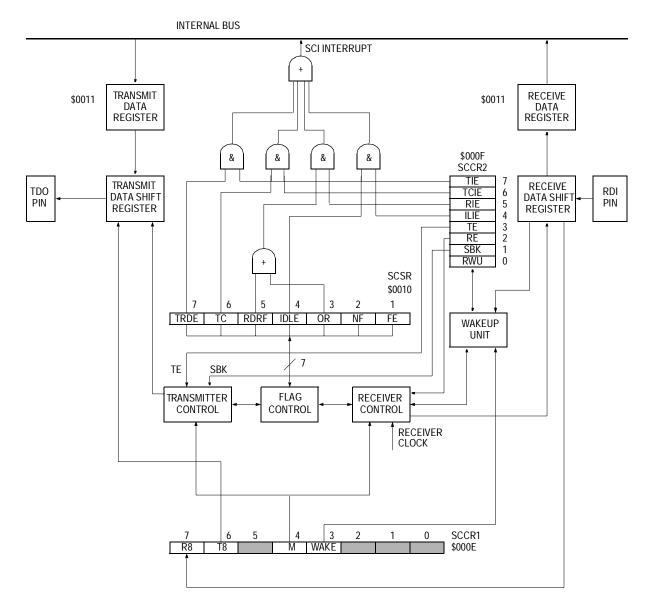


Figure 9-1. Serial Communications Interface Block Diagram

NOTE: The serial communications data register (SCI SCDR) is controlled by the internal R/W signal. It is the transmit data register when written to and the receive data register when read.

9.4 SCI Receiver Features

Features of the SCI receiver include:

- Receiver wakeup function (idle line or address bit)
- Idle line detection
- Framing error detection
- Noise detection
- Overrun detection
- Receiver data register full flag

9.5 SCI Transmitter Features

Features of the SCI transmitter include:

- Transmit data register empty flag
- Transmit complete flag
- Send break

9.6 Functional Description

A block diagram of the SCI is shown in **Figure 9-1**. Option bits in serial control register1 (SCCR1) select the wakeup method (WAKE bit) and data word length (M bit) of the SCI. SCCR2 provides control bits that individually enable the transmitter and receiver, enable system interrupts, and provide the wakeup enable bit (RWU) and the send break code bit (SBK). Control bits in the baud rate register (BAUD) allow the user to select one of 32 different baud rates for the transmitter and receiver.

Data transmission is initiated by writing to the serial communications data register (SCDR). Provided the transmitter is enabled, data stored in the SCDR is transferred to the transmit data shift register. This transfer of data sets the transmit data register empty flag (TDRE) in the SCI status register (SCSR) and generates an interrupt (if transmitter interrupts are enabled). The transfer of data to the transmit data shift

register is synchronized with the bit rate clock (see **Figure 9-2**). All data is transmitted least significant bit first. Upon completion of data transmission, the transmission complete flag (TC) in the SCSR is set (provided no pending data, preamble, or break is to be sent) and an interrupt is generated (if the transmit complete interrupt is enabled). If the transmitter is disabled, and the data, preamble, or break (in the transmit data shift register) has been sent, the TC bit will be set also. This will also generate an interrupt if the transmission complete interrupt enable bit (TCIE) is set. If the transmitter is disabled during a transmission, the character being transmitted will be completed before the transmitter gives up control of the TDO pin.

When SCDR is read, it contains the last data byte received, provided that the receiver is enabled. The receive data register full flag bit (RDRF) in the SCSR is set to indicate that a data byte has been transferred from the input serial shift register to the SCDR; this will cause an interrupt if the receiver interrupt is enabled. The data transfer from the input serial shift register to the SCDR is synchronized by the receiver bit rate clock. The OR (overrun), NF (noise), or FE (framing) error flags in the SCSR may be set if data reception errors occurred.

An idle line interrupt is generated if the idle line interrupt is enabled and the IDLE bit (which detects idle line transmission) in SCSR is set. This allows a receiver that is not in the wakeup mode to detect the end of a message, or the preamble of a new message, or to re-synchronize with the transmitter. A valid character must be received before the idle line condition or the IDLE bit will not be set and idle line interrupt will not be generated.

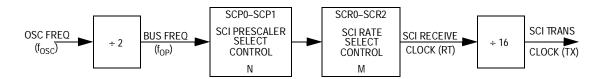


Figure 9-2. Rate Generator Division

9.7 Data Format

Receive data or transmit data is the serial data that is transferred to the internal data bus from the receive data input pin (RDI) or from the internal bus to the transmit data output pin (TDO). The non-return-to-zero (NRZ) data format shown in **Figure 9-3** is used and must meet these criteria:

- The idle line is brought to a logic 1 state prior to transmission/ reception of a character.
- A start bit (logic 0) is used to indicate the start of a frame.
- The data is transmitted and received least significant bit first.
- A stop bit (logic 1) is used to indicate the end of a frame. A frame consists of a start bit, a character of eight or nine data bits, and a stop bit.
- A break is defined as the transmission or reception of a low (logic 0) for at least one complete frame time.

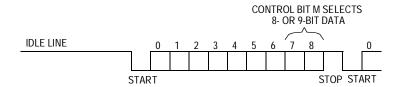


Figure 9-3. Data Format

9.8 Receiver Wakeup Operation

The receiver logic hardware also supports a receiver wakeup function which is intended for systems having more than one receiver. With this function a transmitting device directs messages to an individual receiver or group of receivers by passing addressing information as the initial byte(s) of each message. The wakeup function allows receivers not addressed to remain in a dormant state for the remainder of the unwanted message. This eliminates any further software overhead to service the remaining characters of the unwanted message and thus improves system performance.

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The receiver is placed in wakeup mode by setting the receiver wakeup bit (RWU) in the SCCR2 register. While RWU is set, all of the receiver-related status flags (RDRF, IDLE, OR, NF, and FE) are inhibited (cannot become set).

NOTE:

The idle line detect function is inhibited while the RWU bit is set. Although RWU may be cleared by a software write to SCCR2, it would be unusual to do so.

Normally, RWU is set by software and is cleared automatically in hardware by one of these methods: idle line wakeup or address mark wakeup.

9.8.1 Idle Line Wakeup

In idle line wakeup mode, a dormant receiver wakes up as soon as the RDI line becomes idle. Idle is defined as a continuous logic high level on the RDI line for 10 (or 11) full bit times. Systems using this type of wakeup must provide at least one character time of idle between messages to wake up sleeping receivers, but must not allow any idle time between characters within a message.

9.8.2 Address Mark Wakeup

In address mark wakeup, the most significant bit (MSB) in a character is used to indicate whether it is an address (logic 1) or data (logic 0) character. Sleeping receivers will wake up whenever an address character is received. Systems using this method for wakeup would set the MSB of the first character of each message and leave it clear for all other characters in the message. Idle periods may be present within messages and no idle time is required between messages for this wakeup method.

9.9 Receive Data In (RDI)

Receive data is the serial data that is applied through the input line and the SCI to the internal bus. The receiver circuitry clocks the input at a rate equal to 16 times the baud rate. This time is referred to as the RT rate in **Figure 9-4** and as the receiver clock in **Figure 9-6**.

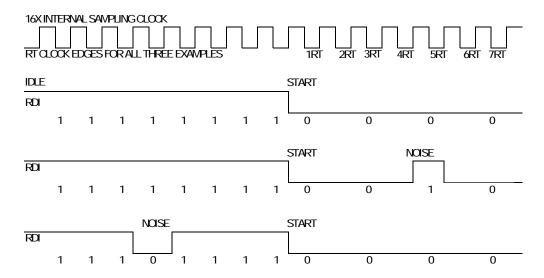


Figure 9-4. SCI Examples of Start Bit Sampling Techniques

The receiver clock generator is controlled by the baud rate register; however, the SCI is synchronized by the start bit, independent of the transmitter.

Once a valid start bit is detected, the start bit, each data bit, and the stop bit are sampled three times at RT intervals 8RT, 9RT, and 10RT (1RT is the position where the bit is expected to start), as shown in **Figure 9-5**. The value of the bit is determined by voting logic which takes the value of the majority of the samples. A noise flag is set when all three samples on a valid start bit or data bit or the stop bit do not agree.

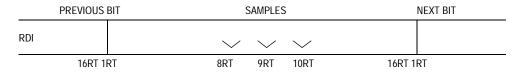


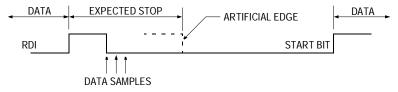
Figure 9-5. SCI Sampling Technique Used on All Bits

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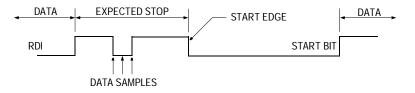
9.10 Start Bit Detection

When the input (idle) line is detected low, it is tested for three more sample times (referred to as the start edge verification samples in **Figure 9-4**). If at least two of these three verification samples detect a logic 0, a valid start bit has been detected; otherwise, the line is assumed to be idle. A noise flag is set if all three verification samples do not detect a logic 0. Thus, a valid start bit could be assumed with a set noise flag present.

If a framing error has occurred without detection of a break (10 0s for 8-bit format or 11 0s for 9-bit format), the circuit continues to operate as if there actually was a stop bit, and the start edge will be placed artificially. The last bit received in the data shift register is inverted to a logic 1, and the three logic 1 start qualifiers (shown in **Figure 9-4**) are forced into the sample shift register during the interval when detection of a start bit is anticipated (see **Figure 9-6**); therefore, the start bit will be accepted no sooner than it is anticipated.



a) Case 1: Receive Line Low During Artificial Edge



b) Case 2: Receive Line High During Expected Start Edge

Figure 9-6. SCI Artificial Start Following a Frame Error

If the receiver detects that a break (RDRF = 1, FE = 1, receiver data register = \$003B) produced the framing error, the start bit will not be artificially induced and the receiver must actually detect a logic 1 before the start bit can be recognized (see Figure 9-7).

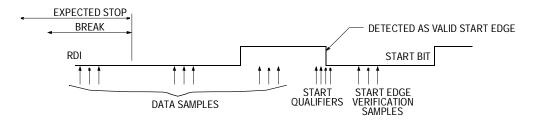


Figure 9-7. SCI Start Bit Following a Break

9.11 Transmit Data Out (TDO)

Transmit data is the serial data from the internal data bus that is applied through the SCI to the output line. Data format is as discussed in **9.7 Data Format** and shown in **Figure 9-3**. The transmitter generates a bit time by using a derivative of the RT clock, thus producing a transmission rate equal to 1/16th that of the receiver sample clock.

9.12 SCI I/O Registers

These I/O registers control and monitor SCI operation:

- SCI data register (SCDR)
- SCI control register 1 (SCCR1)
- SCI control register 2 (SCCR2)
- SCI status register (SCSR)

9.12.1 SCI Data Register

The SCI data register (SCDR), shown in **Figure 9-8**, is the buffer for characters received and for characters transmitted.



Figure 9-8. SCI Data Register (SCDR)

9.12.2 SCI Control Register 1

The SCI control register 1 (SCCR1), shown in **Figure 9-9**, has these functions:

- Stores ninth SCI data bit received and ninth SCI data bit transmitted
- Controls SCI character length
- Controls SCI wakeup method

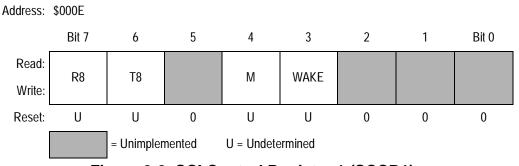


Figure 9-9. SCI Control Register 1 (SCCR1)

R8 — Bit 8 (Received)

When the SCI is receiving 9-bit characters, R8 is the ninth bit of the received character. R8 receives the ninth bit at the same time that the SCDR receives the other eight bits. Resets have no effect on the R8 bit.

MC68HC05C9A - Rev. 5.0

T8 — Bit 8 (Transmitted)

When the SCI is transmitting 9-bit characters, T8 is the ninth bit of the transmitted character. T8 is loaded into the transmit shift register at the same time that the SCDR is loaded into the transmit register. Resets have no effect on the T8 bit.

M — Character Length Bit

This read/write bit determines whether SCI characters are 8 bits long or 9 bits long. The ninth bit can be used as an extra stop bit, as a receiver wakeup signal, or as a mark or space parity bit. Resets have no effect on the M bit.

1 = 9-bit SCI characters

0 = 8-bit SCI characters

WAKE — Wakeup Method Bit

This read/write bit determines which condition wakes up the SCI: a logic 1 (address mark) in the most significant bit (MSB) position of a received character or an idle condition on the PD0/RDI pin. Resets have no effect on the WAKE bit.

1 = Address mark wakeup

0 = Idle line wakeup

9.12.3 SCI Control Register 2

SCI control register 2 (SCCR2), shown in **Figure 9-10**, has these functions:

- Enables the SCI receiver and SCI receiver interrupts
- Enables the SCI transmitter and SCI transmitter interrupts
- Enables SCI receiver idle interrupts
- Enables SCI transmission complete interrupts
- Enables SCI wakeup
- Transmits SCI break characters

Address: \$000F

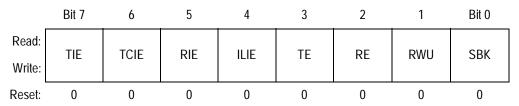


Figure 9-10. SCI Control Register 2 (SCCR2)

TIE — Transmit Interrupt Enable Bit

This read/write bit enables SCI interrupt requests when the TDRE flag becomes set. Resets clear the TIE bit.

- 1 = TDRE interrupt requests enabled
- 0 = TDRE interrupt requests disabled

TCIE — Transmission Complete Interrupt Enable Bit

This read/write bit enables SCI interrupt requests when the TC flag becomes set. Resets clear the TCIE bit.

- 1 = TC interrupt requests enabled
- 0 = TC interrupt requests disabled

RIE — Receiver Interrupt Enable Bit

This read/write bit enables SCI interrupt requests when the RDRF flag or the OR flag becomes set. Resets clear the RIE bit.

- 1 = RDRF interrupt requests enabled
- 0 = RDRF interrupt requests disabled

ILIE — Idle Line Interrupt Enable Bit

This read/write bit enables SCI interrupt requests when the IDLE bit becomes set. Resets clear the ILIE bit.

- 1 = IDLE interrupt requests enabled
- 0 = IDLE interrupt requests disabled

TE — Transmitter Enable Bit

Setting this read/write bit begins the transmission by sending a preamble of 10 or 11 logic 1s from the transmit shift register to the PD1/TDO pin. Resets clear the TE bit.

- 1 = Transmission enabled
- 0 = Transmission disabled

RE — Receiver Enable Bit

Setting this read/write bit enables the receiver. Clearing the RE bit disables the receiver and receiver interrupts but does not affect the receiver interrupt flags. Resets clear the RE bit.

- 1 = Receiver enabled
- 0 = Receiver disabled

RWU — Receiver Wakeup Enable Bit

This read/write bit puts the receiver in a standby state. Typically, data transmitted to the receiver clears the RWU bit and returns the receiver to normal operation. The WAKE bit in SCCR1 determines whether an idle input or an address mark brings the receiver out of standby state. Reset clears the RWU bit.

- 1 = Standby state
- 0 = Normal operation

SBK — Send Break Bit

Setting this read/write bit continuously transmits break codes in the form of 10-bit or 11-bit groups of logic 0s. Clearing the SBK bit stops the break codes and transmits a logic 1 as a start bit. Reset clears the SBK bit.

- 1 = Break codes being transmitted
- 0 = No break codes being transmitted

9.12.4 SCI Status Register

The SCI status register (SCSR), shown in **Figure 9-11**, contains flags to signal these conditions:

- Transfer of SCDR data to transmit shift register complete
- Transmission complete
- Transfer of receive shift register data SCDR complete
- Receiver input idle
- Noisy data
- Framing error

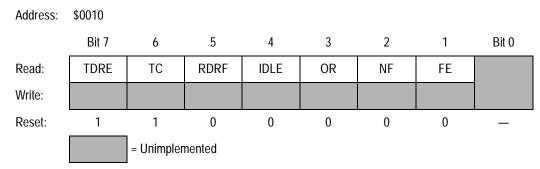


Figure 9-11. SCI Status Register (SCSR)

TDRE — Transmit Data Register Empty Flag

This clearable, read-only flag is set when the data in the SCDR transfers to the transmit shift register. TDRE generates an interrupt request if the TIE bit in SCCR2 is also set. Clear the TDRE bit by reading the SCSR with TDRE set and then writing to the SCDR. Reset sets the TDRE bit. Software must initialize the TDRE bit to logic 0 to avoid an instant interrupt request when turning the transmitter on.

- 1 = SCDR data transferred to transmit shift register
- 0 = SCDR data not transferred to transmit shift register

TC — Transmission Complete Flag

This clearable, read-only flag is set when the TDRE bit is set, and no data, preamble, or break character is being transmitted. TDRE generates an interrupt request if the TCIE bit in SCCR2 is also set. Clear the TC bit by reading the SCSR with TC set, and then writing to

MC68HC05C9A — Rev. 5.0

the SCDR. Reset sets the TC bit. Software must initialize the TC bit to logic 0 to avoid an instant interrupt request when turning the transmitter on.

- 1 = No transmission in progress
- 0 = Transmission in progress

RDRF — Receive Data Register Full Flag

This clearable, read-only flag is set when the data in the receive shift register transfers to the SCI data register. RDRF generates an interrupt request if the RIE bit in the SCCR2 is also set. Clear the RDRF bit by reading the SCSR with RDRF set and then reading the SCDR.

- 1 = Received data available in SCDR
- 0 = Received data not available in SCDR

IDLE — Receiver Idle Flag

This clearable, read-only flag is set when 10 or 11 consecutive logic 1s appear on the receiver input. IDLE generates an interrupt request if the ILIE bit in the SCCR2 is also set. Clear the ILIE bit by reading the SCSR with IDLE set and then reading the SCDR.

- 1 = Receiver input idle
- 0 = Receiver input not idle

OR — Receiver Overrun Flag

This clearable, read-only flag is set if the SCDR is not read before the receive shift register receives the next word. OR generates an interrupt request if the RIE bit in the SCCR2 is also set. The data in the shift register is lost, but the data already in the SCDR is not affected. Clear the OR bit by reading the SCSR with OR set and then reading the SCDR.

- 1 = Receive shift register full and RDRF = 1
- 0 = No receiver overrun

NF — Receiver Noise Flag

This clearable, read-only flag is set when noise is detected in data received in the SCI data register. Clear the NF bit by reading the SCSR and then reading the SCDR.

- 1 = Noise detected in SCDR
- 0 = No noise detected in SCDR

FE — Receiver Framing Error Flag

This clearable, read-only flag is set when there is a logic 0 where a stop bit should be in the character shifted into the receive shift register. If the received word causes both a framing error and an overrun error, the OR flag is set and the FE flag is not set. Clear the FE bit by reading the SCSR and then reading the SCDR.

1 = Framing error

0 = No framing error

9.12.5 Baud Rate Register

The baud rate register (BAUD), shown in **Figure 9-12**, selects the baud rate for both the receiver and the transmitter.



Figure 9-12. Baud Rate Register (BAUD)

SCP1 — SCP0-SCI Prescaler Select Bits

These read/write bits control prescaling of the baud rate generator clock, as shown in **Table 9-1**. Reset clears both SCP1 and SCP0.

Table 9-1. Baud Rate Generator Clock Prescaling

SCP1 and SCP0	Baud Rate Generator Clock
0 0	Internal clock ÷ 1
0 1	Internal clock ÷ 3
1 0	Internal clock ÷ 4
1 1	Internal clock ÷ 13

MC68HC05C9A — Rev. 5.0

SCR2 — SCR0-SCI Baud Rate Select Bits

These read/write bits select the SCI baud rate, as shown in **Table 9-2**. Resets have no effect on the SCR2–SCR0 bits.

Table 9-2. Baud Rate Selection

SCR2, SCR1, and SCR0	SCI Baud Rate (Baud)
0 0 0	Prescaled clock ÷ 1
0 0 1	Prescaled clock ÷ 2
0 1 0	Prescaled clock ÷ 4
0 1 1	Prescaled clock ÷ 8
1 0 0	Prescaled clock ÷ 16
1 0 1	Prescaled clock ÷ 32
1 1 0	Prescaled clock ÷ 64
1 1 1	Prescaled clock ÷ 128

Section 10. Serial Peripheral Interface (SPI)

10.1 Content

10.2	Introduction91
10.3	Features
10.4	SPI Signal Description
10.4.1	Master In Slave Out (MISO)
10.4.2	Master Out Slave In (MOSI)
10.4.3	Serial Clock (SCK)94
10.4.4	Slave Select (SS)
10.5	Functional Description
10.6	SPI Registers
10.6.1	Serial Peripheral Control Register97
10.6.2	Serial Peripheral Status Register
10.6.3	Serial Peripheral Data I/O Register101

10.2 Introduction

The serial peripheral interface (SPI) is an interface built into the device which allows several M68HC05 microcontroller units (MCU), or M68HC05 MCU plus peripheral devices, to be interconnected within a single printed circuit board. In an SPI, separate wires are required for data and clock. In the SPI format, the clock is not included in the data stream and must be furnished as a separate signal. An SPI system may be configured in one containing one master MCU and several slave MCUs or in a system in which an MCU is capable of being a master or a slave.

MC68HC05C9A — Rev. 5.0

Serial Peripheral Interface (SPI)

10.3 Features

SPI features include:

- Full-duplex, 4-wire synchronous transfers
- Master or slave operation
- Bus frequency divided by 2 (maximum) master bit frequency
- Bus frequency (maximum) slave bit frequency
- Four programmable master bit rates
- Programmable clock polarity and phase
- End-of-transmission interrupt flag
- Write collision flag protection
- Master-master mode fault protection capability

10.4 SPI Signal Description

The four basic signals (MOSI, MISO, SCK, and \overline{SS}) are described here. Each signal function is described for both the master and slave modes.

NOTE:

Any SPI output line has to have its corresponding data direction register bit set. If this bit is clear, the line is disconnected from the SPI logic and becomes a general-purpose input line. When the SPI is enabled, any SPI input line is forced to act as an input regardless of what is in the corresponding data direction register bit.

92

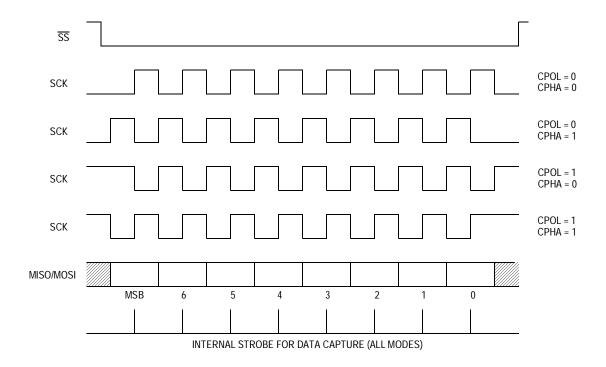


Figure 10-1. Data Clock Timing Diagram

10.4.1 Master In/Slave Out (MISO)

The MISO line is configured as an input in a master device and as an output in a slave device. It is one of the two lines that transfer serial data in one direction, with the most significant bit sent first. The MISO line of a slave device is placed in the high-impedance state if the slave is not selected.

10.4.2 Master Out/Slave In (MOSI)

The MOSI line is configured as an output in a master device and as an input in a slave device. It is one of the two lines that transfer serial data in one direction with the most significant bit sent first.

MC68HC05C9A — Rev. 5.0

Serial Peripheral Interface (SPI)

10.4.3 Serial Clock (SCK)

The master clock is used to synchronize data movement both in and out of the device through its MOSI and MISO lines. The master and slave devices are capable of exchanging a byte of information during a sequence of eight clock cycles. Since SCK is generated by the master device, this line becomes an input on a slave device.

As shown in **Figure 10-1**, four possible timing relationships may be chosen by using control bits CPOL and CPHA in the serial peripheral control register (SPCR). Both master and slave devices must operate with the same timing. The master device always places data on the MOSI line a half cycle before the clock edge (SCK), in order for the slave device to latch the data.

Two bits (SPR0 and SPR1) in the SPCR of the master device select the clock rate. In a slave device, SPR0 and SPR1 have no effect on the operation of the SPI.

10.4.4 Slave Select (SS)

The slave select (SS) input line is used to select a slave device. It has to be low prior to data transactions and must stay low for the duration of the transaction. The SS line on the master must be tied high. In master mode, if the SS pin is pulled low during a transmission, a mode fault error flag (MODF) is set in the SPSR. In master mode the SS pin can be selected as a general-purpose output by writing a 1 in bit 5 of the port D data direction register, thus disabling the mode fault circuit.

When CPHA = 0, the shift clock is the OR of SS with SCK. In this clock phase mode, \overline{SS} must go high between successive characters in an SPI message. When CPHA = 1, \overline{SS} may be left low for several SPI characters. In cases where there is only one SPI slave MCU, its \overline{SS} line could be tied to V_{SS} as long as CPHA = 1 clock modes are used.

10.5 Functional Description

Figure 10-2 shows a block diagram of the serial peripheral interface circuitry. When a master device transmits data to a slave via the MOSI line, the slave device responds by sending data to the master device via the master's MISO line. This implies full duplex transmission with both data out and data in synchronized with the same clock signal. Thus, the byte transmitted is replaced by the byte received and eliminates the need for separate transmit-empty and receive-full status bits. A single status bit (SPIF) is used to signify that the input/output (I/O) operation has been completed.

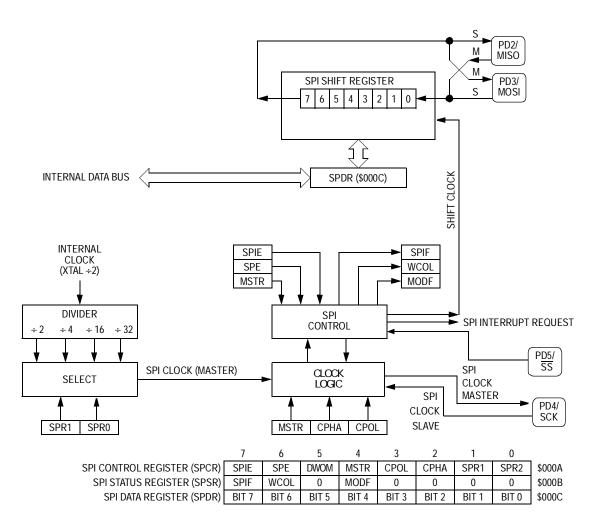


Figure 10-2. Serial Peripheral Interface Block Diagram

MC68HC05C9A — Rev. 5.0

Serial Peripheral Interface (SPI)

The SPI is double buffered on read, but not on write. If a write is performed during data transfer, the transfer occurs uninterrupted, and the write will be unsuccessful. This condition will cause the write collision (WCOL) status bit in the SPSR to be set. After a data byte is shifted, the SPIF flag of the SPSR is set.

In the master mode, the SCK pin is an output. It idles high or low, depending on the CPOL bit in the SPCR, until data is written to the shift register, at which point eight clocks are generated to shift the eight bits of data and then SCK goes idle again.

In a slave mode, the slave select start logic receives a logic low at the SS pin and a clock at the SCK pin. Thus, the slave is synchronized with the master. Data from the master is received serially at the MOSI line and loads the 8-bit shift register. After the 8-bit shift register is loaded, its data is parallel transferred to the read buffer. During a write cycle, data is written into the shift register, then the slave waits for a clock train from the master to shift the data out on the slave's MISO line.

Figure 10-3 illustrates the MOSI, MISO, SCK, and SS master-slave interconnections.

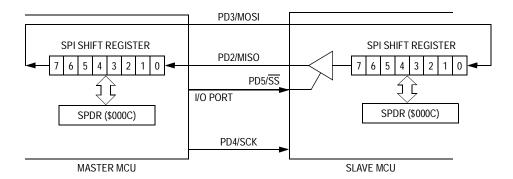


Figure 10-3. Serial Peripheral Interface
Master-Slave Interconnection

Advance Information MC68HC05C9A — Rev. 5.0

10.6 SPI Registers

This subsection describes the three registers in the SPI which provide control, status, and data storage functions. These registers are:

- Serial peripheral control register (SPCR)
- Serial peripheral status register (SPSR)
- Serial peripheral data I/O register (SPDR)

10.6.1 Serial Peripheral Control Register

The SPI control register (SPCR), shown in **Figure 10-4**, controls these functions:

- Enables SPI interrupts
- Enables the SPI system
- Selects between standard CMOS or open drain outputs for port D
- Selects between master mode and slave mode
- Controls the clock/data relationship between master and slave
- Determines the idle level of the clock pin

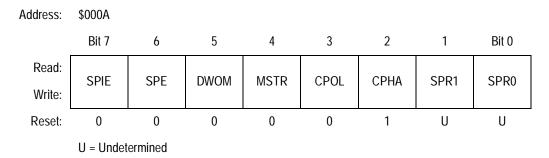


Figure 10-4. SPI Control Register (SPCR)

SPIE — Serial Peripheral Interrupt Enable Bit

This read/write bit enables SPI interrupts. Reset clears the SPIE bit.

1 = SPI interrupts enabled

0 = SPI interrupts disabled

MC68HC05C9A — Rev. 5.0

MOTOROLA

Serial Peripheral Interface (SPI)

SPE — Serial Peripheral System Enable Bit

This read/write bit enables the SPI. Reset clears the SPE bit.

- 1 = SPI system enabled
- 0 = SPI system disabled

DWOM — Port D Wire-OR Mode Option Bit

This read/write bit disables the high side driver transistors on port D outputs so that port D outputs become open-drain drivers. DWOM affects all seven port D pins together.

- 1 = Port D outputs act as open-drain outputs.
- 0 = Port D outputs are normal CMOS outputs.

MSTR — Master Mode Select Bit

This read/write bit selects master mode operation or slave mode operation. Reset clears the MSTR bit.

- 1 = Master mode
- 0 = Slave mode

CPOL — Clock Polarity Bit

When the clock polarity bit is cleared and data is not being transferred, a steady state low value is produced at the SCK pin of the master device. Conversely, if this bit is set, the SCK pin will idle high. This bit is also used in conjunction with the clock phase control bit to produce the desired clock-data relationship between master and slave. See **Figure 10-1**.

CPHA — Clock Phase Bit

The clock phase bit, in conjunction with the CPOL bit, controls the clock-data relationship between master and slave. The CPOL bit can be thought of as simply inserting an inverter in series with the SCK line. The CPHA bit selects one of two fundamentally different clocking protocols. When CPHA = 0, the shift clock is the OR of SCK with \overline{SS} . As soon as \overline{SS} goes low, the transaction begins and the first edge on SCK invokes the first data sample. When CPHA=1, the \overline{SS} pin may be thought of as a simple output enable control. See **Figure 10-1**.

SPR1 and SPR0 — SPI Clock Rate Select Bits

These read/write bits select one of four master mode serial clock rates, as shown in **Table 10-1**. They have no effect in slave mode.

Table 10-1. SPI Clock Rate Selection

SPR1 and SPR0	SPI Clock Rate
0 0	Internal clock ÷ 2
0 1	Internal clock ÷ 4
1 0	Internal clock ÷ 16
1 1	Internal clock ÷ 32

10.6.2 Serial Peripheral Status Register

The SPI status register (SPSR), shown in **Figure 10-5**, contains flags to signal these conditions:

- SPI transmission complete
- Write collision
- Mode fault

\$000B Address: Bit 7 3 2 1 6 5 4 Bit 0 WCOL Read: **SPIF** 0 **MODF** 0 0 0 0 Write: 0 0 0 0 0 0 0 0 Reset: = Unimplemented

Figure 10-5. SPI Status Register

SPIF — SPI Transfer Complete Flag

The serial peripheral data transfer flag bit is set upon completion of data transfer between the processor and external device. If SPIF goes high, and if SPIE is set, a serial peripheral interrupt is generated. Clearing the SPIF bit is accomplished by reading the SPSR (with

MC68HC05C9A — Rev. 5.0

Serial Peripheral Interface (SPI)

SPIF set) followed by an access of the SPDR. Following the initial transfer, unless SPSR is read (with SPIF set) first, attempts to write to SPDR are inhibited.

WCOL — Write Collision Bit

The write collision bit is set when an attempt is made to write to the serial peripheral data register while data transfer is taking place. If CPHA is 0, a transfer is said to begin when \overline{SS} goes low and the transfer ends when \overline{SS} goes high after eight clock cycles on SCK. When CPHA is 1, a transfer is said to begin the first time SCK becomes active while \overline{SS} is low and the transfer ends when the SPIF flag gets set. Clearing the WCOL bit is accomplished by reading the SPSR (with WCOL set) followed by an access to SPDR.

MODF — Mode Fault Flag

The mode fault flag indicates that there may have been a multi-master conflict for system control and allows a proper exit from system operation to a reset or default system state. The MODF bit is normally clear, and is set only when the master device has its SS pin pulled low. Setting the MODF bit affects the internal serial peripheral interface system in these ways:

- 1. An SPI interrupt is generated if SPIE = 1.
- 2. The SPE bit is cleared. This disables the SPI.
- 3. The MSTR bit is cleared, thus forcing the device into the slave mode.

Clearing the MODF bit is accomplished by reading the SPSR (with MODF set), followed by a write to the SPCR. Control bits SPE and MSTR may be restored by user software to their original state during this clearing sequence or after the MODF bit has been cleared. It is also necessary to restore DDRD after a mode fault.

Bits 5 and 3–0 — Not Implemented

These bits always read 0.

10.6.3 Serial Peripheral Data I/O Register

The serial peripheral data I/O register (SPDR), shown in **Figure 10-6**, is used to transmit and receive data on the serial bus. Only a write to this register will initiate transmission/reception of another byte and this will only occur in the master device. At the completion of transmitting a byte of data, the SPIF status bit is set in both the master and slave devices.

When the user reads the serial peripheral data I/O register, a buffer is actually being read. The first SPIF must be cleared by the time a second transfer of the data from the shift register to the read buffer is initiated or an overrun condition will exist. In cases of overrun, the byte which causes the overrun is lost.

A write to the serial peripheral data I/O register is not buffered and places data directly into the shift register for transmission.



Figure 10-6. SPI Data Register (SPDR)

Serial Peripheral Interface (SPI)

Section 11. Instruction Set

11.1 Contents

11.2	Introduction
11.3	Addressing Modes
11.3.1	Inherent104
11.3.2	Immediate
11.3.3	Direct
11.3.4	Extended
11.3.5	
11.3.6	11 11, 11
11.3.7	
11.3.8	Relative
11.4	Instruction Types107
11.4.1	Register/Memory Instructions
11.4.2	Read-Modify-Write Instructions108
11.4.3	Jump/Branch Instructions
11.4.4	Bit Manipulation Instructions
11.4.5	Control Instructions
11.5	Instruction Set Summary112
11.6	Opcode Map

11.2 Introduction

The microcontroller unit (MCU) instruction set has 62 instructions and uses eight addressing modes. The instructions include all those of the M146805 CMOS (complementary metal-oxide semiconductor) Family plus one more: the unsigned multiply (MUL) instruction. The MUL instruction allows unsigned multiplication of the contents of the accumulator (A) and the index register (X). The high-order product is stored in the index register, and the low-order product is stored in the accumulator.

MC68HC05C9A — Rev. 5.0

11.3 Addressing Modes

The central processor unit (CPU) uses eight addressing modes for flexibility in accessing data. The addressing modes provide eight different ways for the CPU to find the data required to execute an instruction. The eight addressing modes are:

- Inherent
- Immediate
- Direct
- Extended
- Indexed, no offset
- Indexed, 8-bit offset
- Indexed, 16-bit offset
- Relative

11.3.1 Inherent

Inherent instructions are those that have no operand, such as return from interrupt (RTI) and stop (STOP). Some of the inherent instructions act on data in the CPU registers, such as set carry flag (SEC) and increment accumulator (INCA). Inherent instructions require no operand address and are one byte long.

11.3.2 Immediate

Immediate instructions are those that contain a value to be used in an operation with the value in the accumulator or index register. Immediate instructions require no operand address and are two bytes long. The opcode is the first byte, and the immediate data value is the second byte.

MOTOROLA

11.3.3 Direct

Direct instructions can access any of the first 256 memory locations with two bytes. The first byte is the opcode, and the second is the low byte of the operand address. In direct addressing, the CPU automatically uses \$00 as the high byte of the operand address.

11.3.4 Extended

Extended instructions use three bytes and can access any address in memory. The first byte is the opcode; the second and third bytes are the high and low bytes of the operand address.

When using the Motorola assembler, the programmer does not need to specify whether an instruction is direct or extended. The assembler automatically selects the shortest form of the instruction.

11.3.5 Indexed, No Offset

Indexed instructions with no offset are 1-byte instructions that can access data with variable addresses within the first 256 memory locations. The index register contains the low byte of the effective address of the operand. The CPU automatically uses \$00 as the high byte, so these instructions can address locations \$0000–\$00FF.

Indexed, no offset instructions are often used to move a pointer through a table or to hold the address of a frequently used random-access memory (RAM) or input/output (I/O) location.

11.3.6 Indexed, 8-Bit Offset

Indexed, 8-bit offset instructions are 2-byte instructions that can access data with variable addresses within the first 511 memory locations. The CPU adds the unsigned byte in the index register to the unsigned byte following the opcode. The sum is the effective address of the operand. These instructions can access locations \$0000–\$01FE.

Indexed 8-bit offset instructions are useful for selecting the kth element in an n-element table. The table can begin anywhere within the first 256 memory locations and could extend as far as location 510 (\$01FE). The k value is typically in the index register, and the address of the beginning of the table is in the byte following the opcode.

11.3.7 Indexed, 16-Bit Offset

Indexed, 16-bit offset instructions are 3-byte instructions that can access data with variable addresses at any location in memory. The CPU adds the unsigned byte in the index register to the two unsigned bytes following the opcode. The sum is the effective address of the operand. The first byte after the opcode is the high byte of the 16-bit offset; the second byte is the low byte of the offset.

Indexed, 16-bit offset instructions are useful for selecting the kth element in an n-element table anywhere in memory.

As with direct and extended addressing, the Motorola assembler determines the shortest form of indexed addressing.

11.3.8 Relative

Relative addressing is only for branch instructions. If the branch condition is true, the CPU finds the effective branch destination by adding the signed byte following the opcode to the contents of the program counter. If the branch condition is not true, the CPU goes to the next instruction. The offset is a signed, two's complement byte that gives a branching range of –128 to +127 bytes from the address of the next location after the branch instruction.

When using the Motorola assembler, the programmer does not need to calculate the offset, because the assembler determines the proper offset and verifies that it is within the span of the branch.

11.4 Instruction Types

The MCU instructions fall into these five categories:

- Register/memory instructions
- Read-modify-write instructions
- Jump/branch instructions
- Bit manipulation instructions
- Control instructions

11.4.1 Register/Memory Instructions

These instructions operate on central processor unit (CPU) registers and memory locations. Most of them use two operands. One operand is in either the accumulator or the index register. The CPU finds the other operand in memory.

Table 11-1. Register/Memory Instructions

Instruction	Mnemonic
Add Memory Byte and Carry Bit to Accumulator	ADC
Add Memory Byte to Accumulator	ADD
AND Memory Byte with Accumulator	AND
Bit Test Accumulator	BIT
Compare Accumulator	CMP
Compare Index Register with Memory Byte	CPX
EXCLUSIVE OR Accumulator with Memory Byte	EOR
Load Accumulator with Memory Byte	LDA
Load Index Register with Memory Byte	LDX
Multiply	MUL
OR Accumulator with Memory Byte	ORA
Subtract Memory Byte and Carry Bit from Accumulator	SBC
Store Accumulator in Memory	STA
Store Index Register in Memory	STX
Subtract Memory Byte from Accumulator	SUB

MC68HC05C9A — Rev. 5.0 Advance Information

11.4.2 Read-Modify-Write Instructions

These instructions read a memory location or a register, modify its contents, and write the modified value back to the memory location or to the register.

NOTE: Do not use read-modify-write operations on write-only registers.

Table 11-2. Read-Modify-Write Instructions

Instruction	Mnemonic
Arithmetic Shift Left (Same as LSL)	ASL
Arithmetic Shift Right	ASR
Bit Clear	BCLR ⁽¹⁾
Bit Set	BSET ⁽¹⁾
Clear Register	CLR
Complement (One's Complement)	COM
Decrement	DEC
Increment	INC
Logical Shift Left (Same as ASL)	LSL
Logical Shift Right	LSR
Negate (Two's Complement)	NEG
Rotate Left through Carry Bit	ROL
Rotate Right through Carry Bit	ROR
Test for Negative or Zero	TST ⁽²⁾

^{1.} Unlike other read-modify-write instructions, BCLR and BSET use only direct addressing.

^{2.} TST is an exception to the read-modify-write sequence because it does not write a replacement value.

11.4.3 Jump/Branch Instructions

Jump instructions allow the CPU to interrupt the normal sequence of the program counter. The unconditional jump instruction (JMP) and the jump-to-subroutine instruction (JSR) have no register operand. Branch instructions allow the CPU to interrupt the normal sequence of the program counter when a test condition is met. If the test condition is not met, the branch is not performed.

The BRCLR and BRSET instructions cause a branch based on the state of any readable bit in the first 256 memory locations. These 3-byte instructions use a combination of direct addressing and relative addressing. The direct address of the byte to be tested is in the byte following the opcode. The third byte is the signed offset byte. The CPU finds the effective branch destination by adding the third byte to the program counter if the specified bit tests true. The bit to be tested and its condition (set or clear) is part of the opcode. The span of branching is from –128 to +127 from the address of the next location after the branch instruction. The CPU also transfers the tested bit to the carry/borrow bit of the condition code register.

Table 11-3. Jump and Branch Instructions

Instruction	Mnemonic
Branch if Carry Bit Clear	BCC
Branch if Carry Bit Set	BCS
Branch if Equal	BEQ
Branch if Half-Carry Bit Clear	внсс
Branch if Half-Carry Bit Set	BHCS
Branch if Higher	BHI
Branch if Higher or Same	BHS
Branch if IRQ Pin High	BIH
Branch if IRQ Pin Low	BIL
Branch if Lower	BLO
Branch if Lower or Same	BLS
Branch if Interrupt Mask Clear	ВМС
Branch if Minus	ВМІ
Branch if Interrupt Mask Set	BMS
Branch if Not Equal	BNE
Branch if Plus	BPL
Branch Always	BRA
Branch if Bit Clear	BRCLR
Branch Never	BRN
Branch if Bit Set	BRSET
Branch to Subroutine	BSR
Unconditional Jump	JMP
Jump to Subroutine	JSR

Advance Information MC68HC05C9A — Rev. 5.0

11.4.4 Bit Manipulation Instructions

The CPU can set or clear any writable bit in the first 256 bytes of memory, which includes I/O registers and on-chip RAM locations. The CPU can also test and branch based on the state of any bit in any of the first 256 memory locations.

Table 11-4. Bit Manipulation Instructions

Instruction	Mnemonic
Bit Clear	BCLR
Branch if Bit Clear	BRCLR
Branch if Bit Set	BRSET
Bit Set	BSET

11.4.5 Control Instructions

These instructions act on CPU registers and control CPU operation during program execution.

Table 11-5. Control Instructions

Instruction	Mnemonic
Clear Carry Bit	CLC
Clear Interrupt Mask	CLI
No Operation	NOP
Reset Stack Pointer	RSP
Return from Interrupt	RTI
Return from Subroutine	RTS
Set Carry Bit	SEC
Set Interrupt Mask	SEI
Stop Oscillator and Enable IRQ Pin	STOP
Software Interrupt	SWI
Transfer Accumulator to Index Register	TAX
Transfer Index Register to Accumulator	TXA
Stop CPU Clock and Enable Interrupts	WAIT

MC68HC05C9A — Rev. 5.0

11.5 Instruction Set Summary

Table 11-6. Instruction Set Summary (Sheet 1 of 7)

Source	Operation	Description			ffe	ct CR		Address Mode	Opcode	Operand	Cycles
Form		·	Н	ı	N	Z	С	Ado	Op	ed O	Š
ADC #opr ADC opr ADC opr ADC opr,X ADC opr,X ADC ,X	Add with Carry	$A \leftarrow (A) + (M) + (C)$	‡	_	‡	‡	\$	IMM DIR EXT IX2 IX1 IX	A9 B9 C9 D9 E9 F9	ii dd hh II ee ff ff	2 3 4 5 4 3
ADD #opr ADD opr ADD opr ADD opr,X ADD opr,X ADD ,X	Add without Carry	$A \leftarrow (A) + (M)$	‡	_	‡	‡	\$	IMM DIR EXT IX2 IX1 IX	AB BB CB DB EB FB	ii dd hh II ee ff ff	2 3 4 5 4 3
AND #opr AND opr AND opr AND opr,X AND opr,X AND ,X	Logical AND	$A \leftarrow (A) \land (M)$	_	_	‡	‡	_	IMM DIR EXT IX2 IX1 IX	A4 B4 C4 D4 E4 F4	ii dd hh II ee ff ff	2 3 4 5 4 3
ASL opr ASLA ASLX ASL opr,X ASL ,X	Arithmetic Shift Left (Same as LSL)	C - 0 b0		_	‡	‡	\$	DIR INH INH IX1 IX	38 48 58 68 78	dd ff	5 3 3 6 5
ASR opr ASRA ASRX ASR opr,X ASR ,X	Arithmetic Shift Right	b7 b0	_	_	‡	‡	‡	DIR INH INH IX1 IX	37 47 57 67 77	dd ff	5 3 3 6 5
BCC rel	Branch if Carry Bit Clear	$PC \leftarrow (PC) + 2 + rel? C = 0$	_	_	_	_	_	REL	24	rr	3
BCLR n opr	Clear Bit n	Mn ← 0	_				_	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	11 13 15 17 19 1B 1D 1F	dd dd dd dd dd dd dd	5 5 5 5 5 5 5 5
BCS rel	Branch if Carry Bit Set (Same as BLO)	PC ← (PC) + 2 + rel? C = 1	_	_	_	_	_	REL	25	rr	3
BEQ rel	Branch if Equal	PC ← (PC) + 2 + rel? Z = 1	_	_	_	_	_	REL	27	rr	3

Advance Information MC68HC05C9A — Rev. 5.0

Table 11-6. Instruction Set Summary (Sheet 2 of 7)

Source Form	Operation	Description		Effect on CCR			on CCR		Address Mode	Opcode	Operand	Cycles
Form	-	•	Н	I	N	Z	С	¥ĕ	Q	obe	ં	
BHCC rel	Branch if Half-Carry Bit Clear	PC ← (PC) + 2 + rel? H = 0	_	_	_	_	_	REL	28	rr	3	
BHCS rel	Branch if Half-Carry Bit Set	PC ← (PC) + 2 + rel? H = 1	_	_	_	_	_	REL	29	rr	3	
BHI rel	Branch if Higher	$PC \leftarrow (PC) + 2 + rel? C \lor Z = 0$	_	_	_	_	_	REL	22	rr	3	
BHS rel	Branch if Higher or Same	PC ← (PC) + 2 + rel? C = 0	_	_	_	_	_	REL	24	rr	3	
BIH rel	Branch if IRQ Pin High	$PC \leftarrow (PC) + 2 + rel ? \overline{IRQ} = 1$	_	_	_	_	_	REL	2F	rr	3	
BIL rel Branch if IRQ Pin Low		$PC \leftarrow (PC) + 2 + rel ? \overline{IRQ} = 0$	_	_	_	_	_	REL	2E	rr	3	
BIT #opr BIT opr BIT opr BIT opr,X BIT opr,X BIT ,X	Bit Test Accumulator with Memory Byte	(A) ∧ (M)	_	_	‡	‡	_	IMM DIR EXT IX2 IX1 IX	A5 B5 C5 D5 E5 F5	ii dd hh II ee ff ff p	2 3 4 5 4 3	
BLO rel	Branch if Lower (Same as BCS)	PC ← (PC) + 2 + rel? C = 1	_	_	_	_	_	REL	25	rr	3	
BLS rel	Branch if Lower or Same	PC ← (PC) + 2 + rel ? C ∨ Z = 1	_	_	_	_	_	REL	23	rr	3	
BMC rel	Branch if Interrupt Mask Clear	PC ← (PC) + 2 + rel ? I = 0	_	_	_	_	_	REL	2C	rr	3	
BMI rel	Branch if Minus	PC ← (PC) + 2 + rel? N = 1	_	_	_	_	_	REL	2B	rr	3	
BMS rel	Branch if Interrupt Mask Set	PC ← (PC) + 2 + rel ? I = 1	_	_	_	_	_	REL	2D	rr	3	
BNE rel	Branch if Not Equal	$PC \leftarrow (PC) + 2 + rel? Z = 0$	_	_	_	_	_	REL	26	rr	3	
BPL rel	Branch if Plus	$PC \leftarrow (PC) + 2 + rel? N = 0$	_	_	_	_	_	REL	2A	rr	3	
BRA rel	Branch Always	PC ← (PC) + 2 + rel? 1 = 1	_	_	_	_	_	REL	20	rr	3	
BRCLR n opr rel	Branch if bit n clear	PC ← (PC) + 2 + <i>rel</i> ? Mn = 0					‡	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b6)	01 03 05 07 09 0B 0D	dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5 5	

MC68HC05C9A — Rev. 5.0

Table 11-6. Instruction Set Summary (Sheet 3 of 7)

Source Form	Operation	Description			ffe			Address Mode	Opcode	Operand	Cycles
1 01111			Н	I	N	Z	С	Adç	do)do	Ç
BRSET n opr rel	Branch if Bit n Set	PC ← (PC) + 2 + <i>rel</i> ? Mn = 1	_				‡	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	00 02 04 06 08 0A 0C 0E	dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5
BRN rel	Branch Never	$PC \leftarrow (PC) + 2 + rel? 1 = 0$	_	_	_	_	_	REL	21	rr	3
BSET n opr	Set Bit n	Mn ← 1	_		_			DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	10 12 14 16 18 1A 1C 1E	dd dd dd dd dd dd dd	5555555
BSR rel	Branch to Subroutine	$PC \leftarrow (PC) + 2; push (PCL)$ $SP \leftarrow (SP) - 1; push (PCH)$ $SP \leftarrow (SP) - 1$ $PC \leftarrow (PC) + rel$	_			_	_	REL	AD	rr	6
CLC	Clear Carry Bit	C ← 0	_	_	_	_	0	INH	98		2
CLI	Clear Interrupt Mask	I ← 0	_	0	_	_	_	INH	9A		2
CLR opr CLRA CLRX CLR opr,X CLR ,X	Clear Byte	$\begin{array}{l} M \leftarrow \$00 \\ A \leftarrow \$00 \\ X \leftarrow \$00 \\ M \leftarrow \$00 \\ M \leftarrow \$00 \end{array}$	_	_	0	1	_	DIR INH INH IX1 IX	3F 4F 5F 6F 7F	dd ff	5 3 6 5
CMP #opr CMP opr CMP opr CMP opr,X CMP opr,X CMP ,X	Compare Accumulator with Memory Byte	(A) – (M)	_	_	‡	‡	\$	IMM DIR EXT IX2 IX1 IX	A1 B1 C1 D1 E1	ii dd hh II ee ff ff	2 3 4 5 4 3
COM opr COMA COMX COM opr,X COM ,X	Complement Byte (One's Complement)	$\begin{array}{l} M \leftarrow (\overline{M}) = \$FF - (M) \\ A \leftarrow (\overline{A}) = \$FF - (M) \\ X \leftarrow (\overline{X}) = \$FF - (M) \\ M \leftarrow (\overline{M}) = \$FF - (M) \\ M \leftarrow (\overline{M}) = \$FF - (M) \end{array}$	_	_	‡	‡	1	DIR INH INH IX1 IX	33 43 53 63 73	dd ff	5 3 6 5

Advance Information MC68HC05C9A — Rev. 5.0

Table 11-6. Instruction Set Summary (Sheet 4 of 7)

Source Form	Operation	Description			ffe C			Address Mode	Opcode	Operand	Cycles
Form		·	Н	I	N	Z	С	Add	o	ope	ં
CPX #opr CPX opr CPX opr CPX opr,X CPX opr,X CPX ,X	Compare Index Register with Memory Byte	(X) – (M)	_	_	‡	‡	1	IMM DIR EXT IX2 IX1 IX	A3 B3 C3 D3 E3 F3	ii dd hh II ee ff ff	2 3 4 5 4 3
DEC opr DECA DECX DEC opr,X DEC ,X	Decrement Byte	$\begin{aligned} \mathbf{M} &\leftarrow (\mathbf{M}) - 1 \\ \mathbf{A} &\leftarrow (\mathbf{A}) - 1 \\ \mathbf{X} &\leftarrow (\mathbf{X}) - 1 \\ \mathbf{M} &\leftarrow (\mathbf{M}) - 1 \\ \mathbf{M} &\leftarrow (\mathbf{M}) - 1 \end{aligned}$		_	‡	‡		DIR INH INH IX1 IX	3A 4A 5A 6A 7A	dd ff	5 3 3 6 5
EOR #opr EOR opr EOR opr,X EOR opr,X EOR ,X	EXCLUSIVE OR Accumulator with Memory Byte	$A \leftarrow (A) \oplus (M)$	_	_	‡	‡		IMM DIR EXT IX2 IX1 IX	A8 B8 C8 D8 E8 F8	ii dd hh II ee ff ff	2 3 4 5 4 3
INC opr INCA INCX INC opr,X INC ,X	Increment Byte	$\begin{aligned} \mathbf{M} &\leftarrow (\mathbf{M}) + 1 \\ \mathbf{A} &\leftarrow (\mathbf{A}) + 1 \\ \mathbf{X} &\leftarrow (\mathbf{X}) + 1 \\ \mathbf{M} &\leftarrow (\mathbf{M}) + 1 \\ \mathbf{M} &\leftarrow (\mathbf{M}) + 1 \end{aligned}$	_	_	‡	‡		DIR INH INH IX1 IX	3C 4C 5C 6C 7C	dd ff	5 3 3 6 5
JMP opr JMP opr, JMP opr,X JMP opr,X JMP ,X	Unconditional Jump	PC ← Jump Address	_	_	_	_		DIR EXT IX2 IX1 IX	BC CC DC EC FC	dd hh II ee ff ff	2 3 4 3 2
JSR opr JSR opr JSR opr,X JSR opr,X JSR ,X	Jump to Subroutine	PC \leftarrow (PC) + n (n = 1, 2, or 3) Push (PCL); SP \leftarrow (SP) – 1 Push (PCH); SP \leftarrow (SP) – 1 PC \leftarrow Conditional Address	_	_	_	_		DIR EXT IX2 IX1 IX	BD CD DD ED FD	dd hh II ee ff ff	5 6 7 6 5
LDA #opr LDA opr LDA opr LDA opr,X LDA opr,X LDA ,X	Load Accumulator with Memory Byte	A ← (M)		_	‡	‡		IMM DIR EXT IX2 IX1 IX	A6 B6 C6 D6 E6 F6	ii dd hh II ee ff ff	2 3 4 5 4 3

MC68HC05C9A — Rev. 5.0

Table 11-6. Instruction Set Summary (Sheet 5 of 7)

LDX #opr LOX opr Load Index Register with Memory Byte X ← (M) — ↓ ↓ ↓ ☐ IIMM AE EXT CE he II 3 4 4 4 4 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1	Source Form	Operation	Description	Effect on CCR				Address Mode	Opcode	Operand	Cycles	
LDX opr LDX opr LDX opr LDX opr with Memory Byte X ← (M) X ← (M) DIR SE dd 3 EXT CE hh II 4 EXT CE hh II 4 LDX opr	Form		•	Н	I	N	Z	С	Ado	o	эdО	Ç
LSLA LSL (Same as ASL) LOgical Shift Left (Same as ASL) LSR opr LSRA LSRX LSRX Logical Shift Right LSRX LSR opr,X	LDX #opr LDX opr LDX opr, LDX opr,X LDX opr,X LDX ,X		X ← (M)	_		‡	‡	_	DIR EXT IX2 IX1	BE CE DE EE	dd hh II ee ff	3 4 5 4
LSRA LSR X Logical Shift Right $X : A \leftarrow (X) \times (A)$ $X : A \leftarrow (X)$	LSLA LSLX			_	1	‡	‡	‡	INH INH IX1	48 58 68		3 3 6
NEG opr Negate Byte (Two's Complement) Negate Byte Byte (Two's Soo (N) Negate Byte (Two's Complement) Negate Byte Byte (Two's Complement) Negate Byte Byte (Two's Complement) Negate Byte (LSRA	Logical Shift Right		_	_	0	‡	\$	INH INH IX1	44 54 64		3 3 6
NEGA NEGX (Two's Complement) Negate Byte (Two's Complement) Neg Operation Neg	MUL	Unsigned Multiply	$X : A \leftarrow (X) \times (A)$	0	_	_	_	0	INH	42		11
ORA #opr ORA opr ORA opr ORA opr ORA opr ORA opr, X ORA opr, X ORA \downarrow Logical OR Accumulator with Memory	NEG opr NEGA NEGX NEG opr,X NEG ,X		$A \leftarrow -(A) = \$00 - (A)$ $X \leftarrow -(X) = \$00 - (X)$ $M \leftarrow -(M) = \$00 - (M)$	_	-	‡	‡	‡	INH INH IX1	40 50 60		3 3 6
ORA opr ORA opr ORA opr ORA opr ORA opr , X ROL opr , X ROR opr RORA RORA RORX RORA RORX ROR opr , X ROR op	NOP	No Operation		_	_	_	_	_	INH	9D		2
ROLA ROLX ROL opr,X ROL opr,X ROL opr,X ROR opr RORA RORX ROR opr,X ROR, X	ORA opr ORA opr, ORA opr,X	Accumulator with	$A \leftarrow (A) \lor (M)$	_		‡	‡	_	DIR EXT IX2 IX1	BA CA DA EA	dd hh II ee ff	3 4 5 4
RORA RORX ROR opr,X ROR, X ROR, X ROR, X ROR opr,X ROR, \times ROR,	ROLX ROL <i>opr</i> ,X			_		‡	‡	‡	INH INH IX1	49 59 69		3 3 6
RSP Reset Stack Pointer SP ← \$00FF — — — — INH 9C 2	RORA RORX ROR <i>opr</i> ,X			_	_	1	‡	‡	INH INH IX1	46 56 66		3 3 6
	RSP	Reset Stack Pointer	SP ← \$00FF	_	_	_			INH	9C		2

Advance Information

MC68HC05C9A — Rev. 5.0

Table 11-6. Instruction Set Summary (Sheet 6 of 7)

Source	Operation	Description			ffe	ct CR		Address Mode	Opcode	Operand	Cycles
Form	·	·	Н	I	N	Z	С	Adc	o	ed O	ં
RTI	Return from Interrupt	$\begin{array}{c} SP \leftarrow (SP) + 1; Pull (CCR) \\ SP \leftarrow (SP) + 1; Pull (A) \\ SP \leftarrow (SP) + 1; Pull (X) \\ SP \leftarrow (SP) + 1; Pull (PCH) \\ SP \leftarrow (SP) + 1; Pull (PCL) \end{array}$	‡	‡	‡	‡	‡	INH	80		6
RTS	Return from Subroutine	$SP \leftarrow (SP) + 1$; Pull (PCH) $SP \leftarrow (SP) + 1$; Pull (PCL)	-	_	_	_	_	INH			
SBC #opr SBC opr SBC opr SBC opr,X SBC opr,X SBC ,X	Subtract Memory Byte and Carry Bit from Accumulator	$A \leftarrow (A) - (M) - (C)$	_	_	‡	‡		IMM DIR EXT IX2 IX1 IX	A2 B2 C2 D2 E2 F2	ii dd hh II ee ff ff	2 3 4 5 4 3
SEC	Set Carry Bit	C ← 1	_	_	_	_	1	INH	99		2
SEI	Set Interrupt Mask	I ← 1	_	1	_	_	_	INH	9B		2
STA opr STA opr STA opr,X STA opr,X STA ,X	Store Accumulator in Memory	$M \leftarrow (A)$	_	_	‡	‡	_	DIR EXT IX2 IX1 IX	B7 C7 D7 E7 F7	dd hh II ee ff ff	4 5 6 5 4
STOP	Stop Oscillator and Enable IRQ Pin		-	0	_	_	_	INH	8E		2
STX opr STX opr STX opr,X STX opr,X STX ,X	Store Index Register In Memory	$M \leftarrow (X)$		_	‡	‡	_	DIR EXT IX2 IX1 IX	BF CF DF EF FF	dd hh II ee ff ff	4 5 6 5 4
SUB #opr SUB opr SUB opr SUB opr,X SUB opr,X SUB ,X	Subtract Memory Byte from Accumulator	$A \leftarrow (A) - (M)$	_	_	‡	‡	‡	IMM DIR EXT IX2 IX1 IX	A0 B0 C0 D0 E0 F0	ii dd hh II ee ff ff	2 3 4 5 4 3
SWI	Software Interrupt	$\begin{array}{c} PC \leftarrow (PC) + 1; Push (PCL) \\ SP \leftarrow (SP) - 1; Push (PCH) \\ SP \leftarrow (SP) - 1; Push (X) \\ SP \leftarrow (SP) - 1; Push (A) \\ SP \leftarrow (SP) - 1; Push (CCR) \\ SP \leftarrow (SP) - 1; I \leftarrow 1 \\ PCH \leftarrow Interrupt Vector High Byte \\ PCL \leftarrow Interrupt Vector Low Byte \\ \end{array}$	_	1	_			INH	83		10

MC68HC05C9A — Rev. 5.0

Table 11-6. Instruction Set Summary (Sheet 7 of 7)

Source Form	Operation	Description			ffe C	ct CR		Address Mode	Opcode	Operand	Cycles
1 01111			Н	I	N	Z	С	Ade	o	obe	ડે
TAX	Transfer Accumulator to Index Register	X ← (A)		_	_	_	_	INH	97		2
TST opr TSTA TSTX TST opr,X TST ,X	Test Memory Byte for Negative or Zero	(M) – \$00	_	_	_	_	_	DIR INH INH IX1 IX	3D 4D 5D 6D 7D	dd ff	4 3 3 5 4
TXA	Transfer Index Register to Accumulator	$A \leftarrow (X)$	-	_	-	-	_	INH	9F		2
WAIT	Stop CPU Clock and Enable Interrupts		_	‡	_	_	_	INH	8F		2

	·							
Α	Accumulator		opr	Operand	(one or two	bytes	s)	
С	Carry/borrow flag		PC	Program (counter			
CCR	Condition code register		PCH	Program	counter high	n byte)	
dd	Direct address of operand		PCL	Program (counter low	byte		
dd rr	Direct address of operand and re	lative offset of branch instruction	REL	Relative a	addressing i	mode		
DIR	Direct addressing mode		rel	Relative p	rogram cou	ınter	offset	byte
ee ff	High and low bytes of offset in in-	dexed, 16-bit offset addressing	rr	Relative p	orogram cou	ınter	offset	byte
EXT	Extended addressing mode		SP	Stack poin	nter			
ff	Offset byte in indexed, 8-bit offse	t addressing	Χ	Index reg	ister			
Н	Half-carry flag		Z	Zero flag				
hh II	High and low bytes of operand ac	ddress in extended addressing	#	Immediate	e value			
I	Interrupt mask		^	Logical Al	ND			
ii	Immediate operand byte		V	Logical O	R			
IMM	Immediate addressing mode		\oplus	Logical E	XCLUSIVE	OR		
INH	Inherent addressing mode		()	Contents	of			
IX	Indexed, no offset addressing mo	ode	-()	Negation	(two's comp	oleme	ent)	
IX1	Indexed, 8-bit offset addressing r	node	\leftarrow	Loaded w	rith			
IX2	Indexed, 16-bit offset addressing	mode	?	lf				
M	Memory location		:	Concaten	ated with			
N	Negative flag		‡	Set or cle	ared			
n	Any bit		_	Not affect	ed			

11.6 Opcode Map

See **Table 11-7**.

Table 11-7. Opcode Map

1C05C9A		Bit Man	ipulation	Branch		d-Modify-V	Vrite		Con	trol			Register	/Memory				
წ ე		DIR	DIR	REL	DIR	INH	INH	IX1	IX	INH	INH	IMM	DIR	EXT	IX2	IX1	IX	
9A _	MSB LSB	0	1	2	3	4	5	6	7	8	9	Α	В	С	D	E	F	MSB LSB
-Rev	0	BRSET0 3 DIR	BSET0 2 DIR	BRA 2 REL	5 NEG 2 DIR	3 NEGA 1 INH	3 NEGX 1 INH	6 NEG 2 IX1	NEG 1 IX	RTI		SUB 2 IMM	3 SUB 2 DIR	SUB 3 EXT	5 SUB 3 IX2	SUB 2 IX1	SUB 1	0
5.0	1	BRCLR0 3 DIR	5 BCLR0 2 DIR	3 BRN 2 REL						6 RTS 1 INH		2 CMP 2 IMM	3 CMP 2 DIR	4 CMP 3 EXT	5 CMP 3 IX2	4 CMP 2 IX1	CMP 1 IX	1
	2	5 BRSET1 3 DIR	5 BSET1 2 DIR	3 BHI 2 REL		11 MUL 1 INH						SBC 2 IMM	3 SBC 2 DIR	SBC 3 EXT	SBC 3 IX2	4 SBC 2 IX1	SBC 1 IX	2
	3	BRCLR1 3 DIR	5 BCLR1 2 DIR	3 BLS 2 REL	5 COM 2 DIR	COMA 1 INH	COMX 1 INH	6 COM 2 IX1	COM IX	10 SWI 1 INH		CPX 2 IMM	3 CPX 2 DIR	CPX 3 EXT	5 CPX 3 IX2	CPX 2 IX1	CPX 1 IX	3
	4	BRSET2 3 DIR	BSET2 2 DIR	BCC 2 REL	LSR 2 DIR	LSRA 1 INH	LSRX 1 INH	LSR 2 IX1	LSR 1 IX			AND 2 IMM	3 AND 2 DIR	AND 3 EXT	5 AND 3 IX2	AND 2 IX1	AND IX	4
	5	5 BRCLR2 3 DIR	5 BCLR2 2 DIR	3 BCS/BLO 2 REL								BIT 2 IMM	3 BIT 2 DIR	BIT 3 EXT	5 BIT 3 IX2	BIT 2 IX1	BIT IX	5
	6	BRSET3 3 DIR	BSET3 2 DIR	3 BNE 2 REL	5 ROR 2 DIR	RORA 1 INH	RORX 1 INH	6 ROR 2 IX1	ROR 1 IX			2 LDA 2 IMM	3 LDA 2 DIR	LDA 3 EXT	5 LDA 3 IX2	4 LDA 2 IX1	LDA 1 IX	6
	7	5 BRCLR3 3 DIR	5 BCLR3 2 DIR	3 BEQ 2 REL	5 ASR 2 DIR	3 ASRA 1 INH	3 ASRX 1 INH	6 ASR 2 IX1	ASR 1 IX		TAX 1 INH		4 STA 2 DIR	5 STA 3 EXT	6 STA 3 IX2	5 STA 2 IX1	STA IX	7
	8	5 BRSET4 3 DIR	5 BSET4 2 DIR	3 BHCC 2 REL	5 ASL/LSL 2 DIR	3 ASLA/LSLA 1 INH	3 ASLX/LSLX 1 INH	6 ASL/LSL 2 IX1	ASL/LSL 1 IX		CLC 1 INH	EOR 2 IMM	3 EOR 2 DIR	EOR 3 EXT	5 EOR 3 IX2	EOR 2 IX1	EOR 1 IX	8
	9	5 BRCLR4 3 DIR	5 BCLR4 2 DIR	3 BHCS 2 REL	5 ROL 2 DIR	3 ROLA 1 INH	3 ROLX 1 INH	6 ROL 2 IX1	FOL IX		SEC 1 INH	ADC 2 IMM	3 ADC 2 DIR	ADC 3 EXT	5 ADC 3 IX2	ADC 2 IX1	ADC 1 IX	9
	Α	BRSET5 3 DIR	5 BSET5 2 DIR	3 BPL 2 REL	5 DEC 2 DIR	3 DECA 1 INH	3 DECX 1 INH	6 DEC 2 IX1	DEC 1 IX		CLI 1 INH	ORA 2 IMM	ORA	ORA 3 EXT	5 ORA 3 IX2	ORA 2 IX1	ORA 1 IX	A
	В	BRCLR5 3 DIR	5 BCLR5 2 DIR	3 BMI 2 REL							SEI 1 INH	ADD	ADD	4 ADD 3 EXT	5 ADD 3 IX2	4 ADD 2 IX1	ADD IX	В
	С	BRSET6 3 DIR	BSET6 2 DIR	BMC 2 REL	5 INC 2 DIR	INCA 1 INH	INCX 1 INH	INC 2 IX1	INC 1 IX		RSP 1 INH		JMP 2 DIR	JMP 3 EXT	JMP 3 IX2	3 JMP 2 IX1	JMP 1 IX	С
	D	5 BRCLR6 3 DIR	5 BCLR6 2 DIR	3 BMS 2 REL	4 TST 2 DIR	3 TSTA 1 INH	3 TSTX 1 INH	5 TST 2 IX1	TST IX		2 NOP 1 INH	6 BSR 2 REL	JSR	6 JSR 3 EXT	JSR 3 IX2	6 JSR 2 IX1	JSR 1 IX	D
A	E	BRSET7 3 DIR	BSET7 2 DIR	3 BIL 2 REL						2 STOP 1 INH		LDX 2 IMM	3 LDX 2 DIR	LDX 3 EXT	5 LDX 3 IX2	4 LDX 2 IX1	LDX 1 IX	E
Advance	F	BRCLR7 3 DIR		3 BIH 2 REL	5 CLR 2 DIR	3 CLRA 1 INH	3 CLRX 1 INH	6 CLR 2 IX1	5 CLR 1 IX	2 WAIT 1 INH	2 TXA 1 INH		STX 2 DIR	5 STX 3 EXT	6 STX 3 IX2	5 STX 2 IX1	STX 1 IX	F
ë Info		INH = Inh IMM = Im	erent mediate		Relative	Offset						MSB LSB	0	MSB of O	pcode in H	exadecima	al	

IMM = Immediate DIR = Direct

IX = Indexed, No Offset

IX2 = Indexed, 16-Bit Offset

LSB of Opcode in Hexadecimal

LSB 5 BRSET0 3 DIR Number of Cycles Opcode Mnemonic Number of Bytes/Addressing Mode 0

Instruction Set

Advance Information

MOTOROLA

Section 12. Electrical Specifications

12.1 Contents

12.2	Maximum Ratings122
12.3	Operating Temperature123
12.4	Thermal Characteristics
12.5	Power Considerations
12.6	5.0-Volt DC Electrical Characteristics125
12.7	3.3-Volt DC Electrical Characteristics126
12.8	5.0-Volt Control Timing128
12.9	3.3-Volt Control Timing129
12.10	5.0-Volt Serial Peripheral Interface Timing132
12.11	3.3-Volt Serial Peripheral Interface Timing

MC68HC05C9A — Rev. 5.0 Advance Information

Electrical Specifications

12.2 Maximum Ratings

Maximum ratings are the extreme limits to which the microcontroller unit (MCU) can be exposed without permanently damaging it.

The MCU contains circuitry to protect the inputs against damage from high static voltages; however, do not apply voltages higher than those shown in the table here. Keep V_{ln} and V_{Out} within the range $V_{SS} \leq (V_{ln} \text{ or } V_{Out}) \leq V_{DD}$. Connect unused inputs to the appropriate voltage level, either V_{SS} or V_{DD} .

Rating	Symbol	Value	Unit
Supply voltage	V _{DD}	-0.3 to +7.0	V
Input voltage Normal operation Self-check mode (IRQ pin only)	V _{In} V _{TST}	V_{SS} =0.3 to V_{DD} + 0.3 V_{SS} =0.3 to 2 x V_{DD} + 0.3	V
Current drain per pin (Excluding V _{DD} and V _{SS})	I	25	mA
Storage temperature range	T _{STG}	-65 to +150	°C

NOTE:

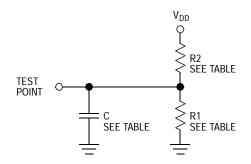
This device is not guaranteed to operate properly at the maximum ratings. Refer to 12.6 5.0-Volt DC Electrical Characteristics for guaranteed operating conditions.

12.3 Operating Temperature

Characteristic	Symbol	Value	Unit
Operating temperature range MC68HC05C9AP, FN, B, FB MC68HC05C9AMP, MFN, MB, MFB	ТА	T _L to T _H 0 to +70 –40 to +125	°C

12.4 Thermal Characteristics

Characteristic	Symbol	Value	Unit
Thermal resistance plastic dual in-line (PDIP)	θ_{JA}	60	°C/W
Thermal resistance plastic-leaded chip carrier (PLCC)	θ_{JA}	70	°C/W
Thermal resistance quad flat pack (QFP)	θ_{JA}	95	°C/W
Thermal resistance plastic shrink DIP (SDIP)	θ_{JA}	60	°C/W



 $V_{DD} = 4.5 V$

Pins	R1	R2	С
PA7-PA0	3.26 Ω	2.38 Ω	50 pF
PB7-PB0			
PC7-PC0			
PD5-PD0, PD7			

 $V_{DD} = 3.0 V$

Pins	R1	R2	С
PA7-PA0	10.91 Ω	6.32 Ω	50 pF
PB7–PB0			
PC7-PC0			
PD5-PD0, PD7			

Figure 12-1. Test Load

 $\mathsf{MC68HC05C9A} - \mathsf{Rev.}\ 5.0$

12.5 Power Considerations

The average chip-junction temperature, T_J, in °C, can be obtained from:

$$T_{J} = T_{A} + (P_{D} \times \theta_{JA}) \tag{1}$$

where:

T_A = Ambient temperature, °C

 θ_{JA} = Package thermal resistance, junction to ambient, °C/W

 $P_D = P_{INT} + P_{I/O}$

 $P_{INT} = I_{DD} \times V_{DD}$ watts (chip internal power)

 $P_{I/O}$ = Power dissipation on input and output pins (user determined)

For most applications, P_{I/O} « P_{INT} and can be neglected.

The following is an approximate relationship between P_D and T_J (neglecting P_J):

$$P_D = K \div (T_J + 273^{\circ}C)$$
 (2)

Solving equations (1) and (2) for K gives:

$$K = P_D \times (T_A + 273^{\circ}C) + \theta_{JA} \times (P_D)^2$$
 (3)

where K is a constant pertaining to the particular part. K can be determined from equation (3) by measuring P_D (at equilibrium) for a known T_A . Using this value of K, the values of P_D and P_D and P_D can be obtained by solving equations (1) and (2) iteratively for any value of P_D .

12.6 5.0-Volt DC Electrical Characteristics

Characteristic ^{(1) (2)}	Symbol	Min	Тур	Max	Unit
Output voltage $I_{Load} = 10.0 \ \mu A$ $I_{Load} = -10.0 \ \mu A$	V _{OL} V _{OH}	 V _{DD} -0.1	_	0.1 —	V
Output high voltage (I _{Load} = -0.8 mA) PA7-PA0, PB7-PB0, PC6-PC0, TCMP, PD7, PD0 (I _{Load} = -1.6 mA) PD5-PD1 (I _{Load} = -5.0 mA) PC7	V _{OH}	V _{DD} -0.8 V _{DD} -0.8 V _{DD} -0.8	_ _ _	_ _ _	V
Output low voltage (I _{Load} = 1.6 mA) PA7–PA0, PB7–PB0, PC6–PC0, PD7, PD5–PD0, TCMP (I _{Load} = 10 mA) PC7	V _{OL}			0.4 0.4	V
Input high voltage PA7-PA0, PB7-PB0, PC7-PC0, PD7, PD5-PD0, TCAP, IRQ, RESET, OSC1	V _{IH}	$0.7 \times V_{DD}$	_	V _{DD}	V
Input low voltage PA7-PA0, PB7-PB0, PC7-PC0, PD7, PD5-PD0, TCAP, IRQ, RESET, OSC1	V _{IL}	V _{SS}	_	$0.2 \times V_{DD}$	V
Supply current (4.5–5.5 Vdc @ f _{OP} = 2.1 MHz) Run ⁽³⁾ Wait ⁽⁴⁾ Stop ⁽⁵⁾ 25°C 0 to 70°C -40 to +125°C	I _{DD}	_ _ _ _	3.5 1.0 1.0 2.0 7.0	5.25 3.25 20.0 40.0 50.0	mA mA μA μA μA
I/O ports hi-z leakage current PA7-PA0, PB7-PB0 (without pullup) PC7-PC0, PD7, PD5-PD0	I _{OZ}	_	1.0	10	μΑ
Input current RESET, IRQ, OSC1, TCAP, PD7, PD5–PD0	I _{In}	_	0.5	1	μΑ
Input pullup current ⁽⁶⁾ PB7–PB0 (with pullup)	I _{In}	5	_	60	μΑ
Capacitance Ports (as input or output) RESET, IRQ, OSC1, TCAP, PD7, PD5, PD0	C _{Out} C _{In}			12 8	pF

^{1.} V_{DD} = 5.0 Vdc \pm 10%, V_{SS} = 0 Vdc, T_A = -40 to +125°C, unless otherwise noted

MC68HC05C9A - Rev. 5.0

^{2.} Typical values reflect measurements taken on average processed devices at the midpoint of voltage range, 25°C only.

^{3.} Run (operating) I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} –0.2 V; no dc loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2

^{4.} Wait I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} -0.2 V; no dc loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2. Wait I_{DD} is affected linearly by the OSC2 capacitance.

^{5.} Stop I_{DD} measured with OSC1 = 0.2 V; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, $V_{IH} = V_{DD} - 0.2 \text{ V}$

^{6.} Input pullup current measured with $V_{IL} = 0.2 \text{ V}$

12.7 3.3-Volt DC Electrical Characteristics

Characteristic ^{(1) (2)}	Symbol	Min	Тур	Max	Unit
Output voltage $I_{Load} = 10.0 \ \mu A$ $I_{Load} = -10.0 \ \mu A$	V _{OL} V _{OH}	 V _{DD} -0.1	_	0.1 —	V
Output high voltage (I _{Load} = -0.2 mA) PA7-PA0, PB7-PB0, PC6-PC0, TCMP, PD7, PD0 (I _{Load} = -0.4 mA) PD5-PD1 (I _{Load} = -1.5 mA) PC7	V _{OH}	V _{DD} -0.3 V _{DD} -0.3 V _{DD} -0.3	_ _ _	_ _ _	V
Output low voltage (I _{Load} = 0.4 mA) PA7–PA0, PB7–PB0, PC6–PC0, PD7, PD5–PD0, TCMP (I _{Load} = 6 mA) PC7	V _{OL}			0.3 0.3	V
Input high voltage PA7-PA0, PB7-PB0, PC7-PC0, PD7, PD5-PD0, TCAP, IRQ, RESET, OSC1	V _{IH}	$0.7 \times V_{DD}$	_	V _{DD}	V
Input low voltage PA7–PA0, PB7–PB0, PC7–PC0, PD7, PD5–PD0, TCAP, IRQ, RESET, OSC1	V _{IL}	V _{SS}	_	$0.2 \times V_{DD}$	V
Supply current (3.0–3.6 Vdc @ f _{OP} = 1.0 MHz) Run ⁽³⁾ Wait ⁽⁴⁾ Stop ⁽⁵⁾ 25°C 0 to 70°C -40 to +125°C	I _{DD}	_ _ _ _	1.0 500 1.0 1.0 2.5	1.6 900 8 16 20	mA μA μA μA μA
I/O ports hi-z leakage current PA7-PA0, PB7-PB0 (without pullup) PC7-PC0, PD7, PD5-PD0	l _{oz}	_	1.0	10	μА
Input current RESET, IRQ, OSC1, TCAP, PD7, PD5–PD0	I _{In}	_	0.5	1	μΑ
Input pullup current ⁽⁶⁾ PB7–PB0 (with pullup)	I _{In}	0.5	_	20	μΑ
Capacitance Ports (as input or output) RESET, IRQ, OSC1, TCAP, PD7, PD5, PD0	C _{Out} C _{In}			12 8	pF

- 1. V_{DD} = 3.3 Vdc \pm 0.3 Vdc, V_{SS} = 0 Vdc, T_A = -40 to +125°C, unless otherwise noted
- 2. Typical values reflect measurements taken on average processed devices at the midpoint of voltage range, 25°C only.
- 3. Run (operating) I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} -0.2 V; no dc loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2
- 4. Wait I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD}, all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} -0.2 V; no dc loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2. Wait I_{DD} is affected linearly by the OSC2 capacitance.
- 5. Stop I_{DD} measured with OSC1 = 0.2 V; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, $V_{IH} = V_{DD} 0.2$ V
- 6. Input pullup current measured with V_{IL} = 0.2 V

Advance Information

MC68HC05C9A — Rev. 5.0

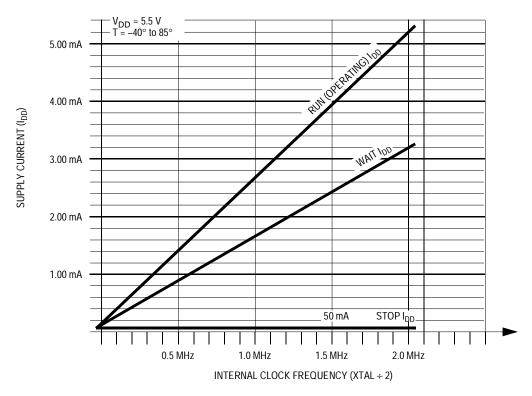


Figure 12-2. Maximum Supply Current versus Internal Clock Frequency, $V_{DD} = 5.5 \text{ V}$

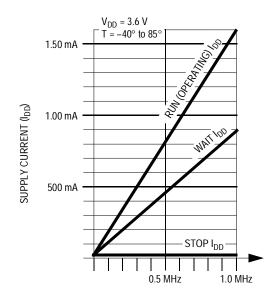


Figure 12-3. Maximum Supply Current versus Internal Clock Frequency, V_{DD} = 3.6 V

MC68HC05C9A — Rev. 5.0

Electrical Specifications

12.8 5.0-Volt Control Timing

Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
Frequency of operation Crystal External clock	fosc	— dc	4.2 4.2	MHz
Internal operating frequency (f _{OSC} ÷ 2) Crystal External clock	f _{OP}	— dc	2.1 2.1	MHz
Cycle time	t _{cyc}	480	_	ns
Crystal oscillator startup time	t _{OXOV}	_	100	ms
Stop recovery startup time (crystal oscillator)	t _{ILCH}	_	100	ms
RESET pulse width	t _{RL}	1.5	_	t _{cyc}
Timer Resolution ⁽²⁾ Input capture pulse width Input capture pulse period	t _{RESL} t _{TH} , t _{TL} t _{TLTL}	4.0 125 (3)	_ _ _	t _{cyc} ns t _{cyc}
Interrupt pulse width low (edge-triggered)	t _{ILIH}	125	_	ns
Interrupt pulse period	t _{ILIL}	(4)	_	t _{cyc}
OSC1 pulse width	t _{OH} , t _{OL}	90	_	ns

^{1.} V_{DD} = 5.0 Vdc \pm 10%, V_{SS} = 0 Vdc, T_A = -40 to +125°C, unless otherwise noted

^{2.} Because a 2-bit prescaler in the timer must count four internal cycles (t_{CYC}), this is the limiting minimum factor in determining the timer resolution.

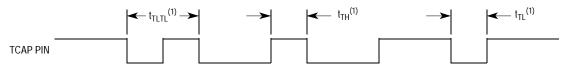
^{3.} The minimum period t_{TLTL} should not be less than the number of cycle times it takes to execute the capture interrupt service routine plus 24 t_{CYC} .

^{4.} The minimum t_{ILIL} should not be less than the number of cycle times it takes to execute the interrupt service routine plus 19 t_{CYC} .

12.9 3.3-Volt Control Timing

Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
Frequency of operation Crystal External clock	fosc	— dc	2.0 2.0	MHz
Internal operating frequency (f _{OSC} ÷ 2) Crystal External clock	f _{OP}	— dc	1.0 1.0	MHz
Cycle time	t _{cyc}	1000	_	ns
Crystal oscillator start-up time	t _{OXOV}	_	100	ms
Stop recovery start-up time (crystal oscillator)	t _{ILCH}	_	100	ms
RESET pulse width	t _{RL}	1.5	_	t _{cyc}
Timer Resolution ⁽²⁾ Input capture pulse width Input capture pulse period	tRESL t _{TH} , t _{TL} t _{TLTL}	4.0 125 (3)	_ _ _	t _{cyc} ns t _{cyc}
Interrupt pulse width low (edge-triggered)	t _{ILIH}	250	_	ns
Interrupt pulse period	t _{ILIL}	(4)	_	t _{cyc}
OSC1 pulse width	t _{OH} , t _{OL}	200	_	ns

- 1. V_{DD} = 3.3 Vdc \pm 0.3 Vdc, V_{SS} = 0 Vdc, T_A = -40 to +125°C, unless otherwise noted
- 2. Because a 2-bit prescaler in the timer must count four internal cycles (t_{CYC}), this is the limiting minimum factor in determining the timer resolution.
- 3. The minimum period t_{TLTL} should not be less than the number of cycle times it takes to execute the capture interrupt service routine plus 24 t_{CYC} .
- 4. The minimum t_{ILIL} should not be less than the number of cycle times it takes to execute the interrupt service routine plus 19 t_{CYC} .

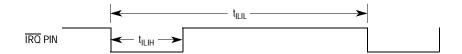


1. Refer to timer resolution data in 12.8 5.0-Volt Control Timing and 12.9 3.3-Volt Control Timing.

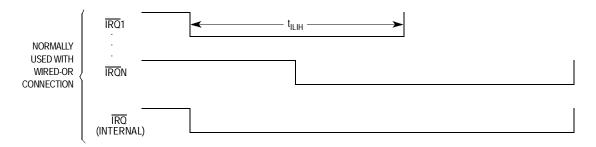
Figure 12-4. TCAP Timing Relationships

MC68HC05C9A — Rev. 5.0

Electrical Specifications



a. **Edge-Sensitive Trigger Condition**. The minimum pulse width (t_{ILIH}) is either 125 ns $(f_{OP} = 2.1 \text{ MHz})$ or 250 ns $(f_{OP} = 1 \text{ MHz})$. The period t_{ILIL} should not be less than the number of t_{CYC} cycles it takes to execute the interrupt service routine plus 19 t_{CYC} cycles.



b. Level-Sensitive Trigger Condition. If after servicing an interrupt the $\overline{\mbox{IRQ}}$ remains low, the next interrupt is recognized.

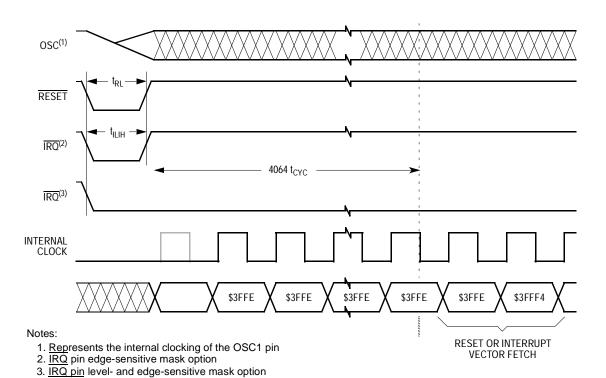
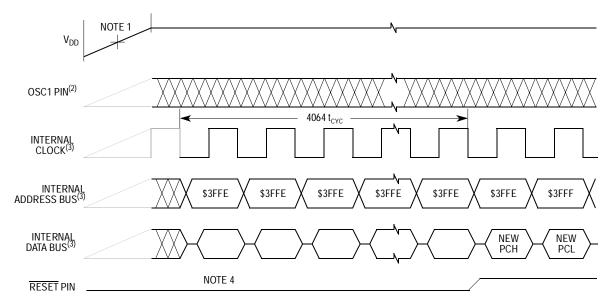


Figure 12-5. External Interrupt Timing

Figure 12-6. Stop Recovery Timing Diagram

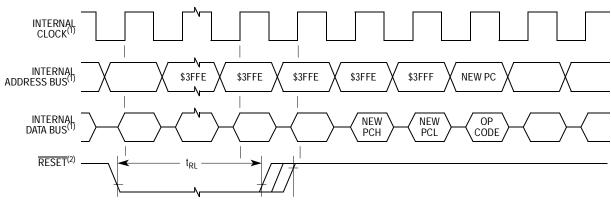
4. RESET vector address shown for timing example



Notes:

- 1. Power-on reset threshold is typically between 1 V and 2 V.
- OSC1 line is meant to represent time only, not frequency.
 Internal clock, internal address bus, and internal data bus are not available externally.
- 4. RESET outputs V_{OL} during 4064 POR cycles.

Figure 12-7. Power-On Reset Timing Diagram



Notes:

- 1. Internal clock, internal address bus, and internal data bus are not available externally.
- 2. The next rising edge of the internal clock after the rising edge of RESET initiates the reset sequence.

Figure 12-8. External Reset Timing

MC68HC05C9A — Rev. 5.0

12.10 5.0-Volt Serial Peripheral Interface Timing

No.	Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
	Operating frequency Master Slave	f _{OP(M)} f _{OP(S)}	dc dc	0.5 2.1	f _{OP} MHz
1	Cycle time Master Slave	t _{CYC(M)}	2.0 480		t _{CYC}
2	Enable lead time Master Slave	t _{Lead(M)} t _{Lead(S)}	Note ⁽²⁾ 240		ns
3	Enable lag time Master Slave	t _{Lag(M)} t _{Lag(S)}	Note ⁽²⁾ 720		ns
4	Clock (SCK) high time Master Slave	tw(sckh)m tw(sckh)s	340 190		ns
5	Clock (SCK) low time Master Slave	tw(SCKL)M	340 190		ns
6	Data setup time (inputs) Master Slave	t _{SU(M)} t _{SU(S)}	100 100	_	ns
7	Data hold time (inputs) Master Slave	t _{H(M)}	100 100	_	ns
8	Slave access time (time to data active from high-impedance state)	t _A	0	120	ns
9	Slave disable time (hold time to high-impedance state)	t _{DIS}	_	240	ns
10	Data Valid Master (before capture edge) Slave (after enable edge) ⁽³⁾	t _{V(M)} t _{V(S)}	0.25 —	_ 240	t _{CYC(M)} ns
11	Data hold time (outputs) Master (after capture edge) Slave (after enable edge)	t _{HO(M)}	0.25 0	_	t _{CYC(M)} ns
12	Rise time (20% V_{DD} to 70% V_{DD} , C_L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{RM} t _{RS}		100 2.0	ns µs
13	Fall time (70% V_{DD} to 20% V_{DD} , C_L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{FM}	_ _	100 2.0	ns μs

^{1.} V_{DD} = 5.0 Vdc ± 10%; V_{SS} = 0 Vdc, T_A = -40 to +125°C, unless otherwise noted. Refer to **Figure 12-9** and **Figure 12-10** for timing diagrams.

^{2.} Signal production depends on software.

^{3.} Assumes 200 pF load on all SPI pins

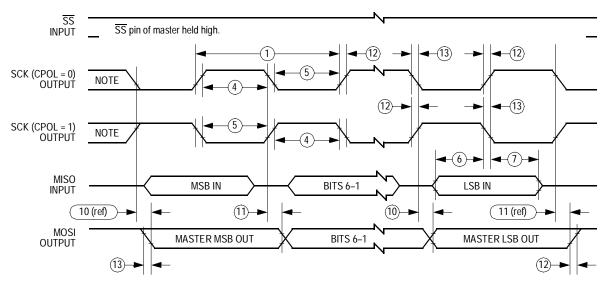
12.11 3.3-Volt Serial Peripheral Interface Timing

No.	Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
	Operating frequency Master Slave	f _{OP(M)} f _{OP(S)}	dc dc	0.5 1.0	f _{OP} MHz
1	Cycle time Master Slave	t _{CYC(M)} t _{CYC(S)}	2.0 1.0		t _{CYC}
2	Enable lead time Master Slave	t _{Lead(M)} t _{Lead(S)}	Note 2 500		ns
3	Enable lag time Master Slave	t _{Lag(M)} t _{Lag(S)}	Note 2 1.5		ns
4	Clock (SCK) high time Master Slave	tw(sckh)m	720 400		ns
5	Clock (SCK) low time Master Slave	tw(SCKL)M	720 400	_	ns
6	Data setup time (inputs) Master Slave	t _{SU(M)}	200 200	_	ns
7	Data hold time (inputs) Master Slave	t _{H(M)}	200 200	_	ns
8	Slave access time (time to data active from high-impedance state)	t _A	0	250	ns
9	Slave disable time (hold time to high-impedance state)	t _{DIS}	_	500	ns
10	Data valid Master (before capture edge) Slave (after enable edge) ⁽³⁾	t _{V(M)} t _{V(S)}	0.25 —	 500	t _{CYC(M)}
11	Data hold time (outputs) Master (after capture edge) Slave (after enable edge)	t _{HO(M)}	0.25 0		t _{CYC(M)}
12	Rise time (20% V _{DD} to 70% V _{DD} , C _L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{RM} t _{RS}		200 2.0	ns μs
13	Fall time (70% V _{DD} to 20% V _{DD} , C _L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{FM} t _{FS}		200 2.0	ns µs

^{1.} V_{DD} = 3.3 Vdc ± 0.3 Vdc; V_{SS} = 0 Vdc, T_A = -40 to +125°C, unless otherwise noted. Refer to **Figure 12-9** and **Figure 12-10** for timing diagrams.

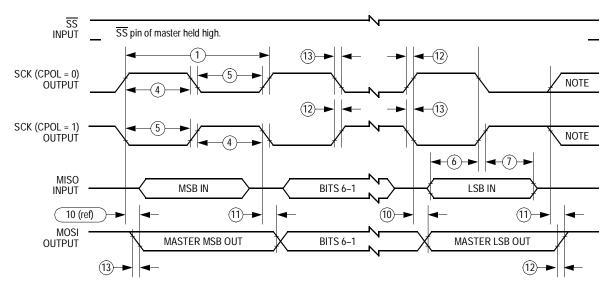
MC68HC05C9A — Rev. 5.0

Signal production depends on software.
 Assumes 200 pF load on all SPI pins



Note: This first clock edge is generated internally, but is not seen at the SCK pin.

a) SPI Master Timing (CPHA = 0)



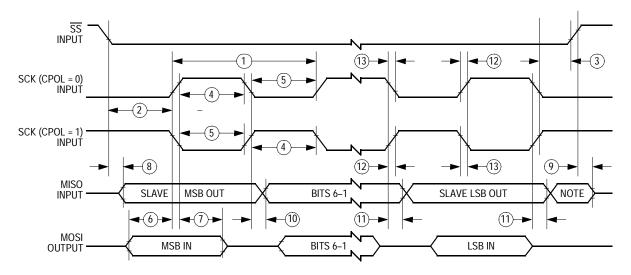
Note: This last clock edge is generated internally, but is not seen at the SCK pin.

b) SPI Master Timing (CPHA = 1)

Figure 12-9. SPI Master Timing Diagram

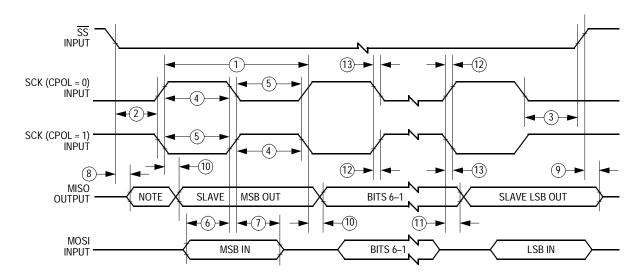
Advance Information

MC68HC05C9A — Rev. 5.0



Note: Not defined but normally MSB of character just received

a) SPI Slave Timing (CPHA = 0)



Note: Not defined but normally LSB of character previously transmitted

a) SPI Slave Timing (CPHA = 1)

Figure 12-10. SPI Slave Timing Diagram

Electrical Specifications

Section 13. Mechanical Specifications

13.1 Contents

13.2	Introduction
13.3	40-Pin Plastic Dual In-Line (DIP) Package (Case 711-03)
13.4	42-Pin Plastic Shrink Dual In-Line (SDIP) Package (Case 858-01)
13.5	44-Lead Plastic-Leaded Chip Carrier (PLCC) (Case 777-02)
13.6	44-Lead Quad Flat Pack (QFP) (Case 824A-01)

13.2 Introduction

This section describes the dimensions of the plastic dual in-line package (DIP), plastic shrink dual in-line package (SDIP), plastic-leaded chip carrier (PLCC), and quad flat pack (QFP) MCU packages.

Package dimensions available at the time of this publication are provided in this section.

To make sure that you have the latest case outline specifications, contact one of the following:

- Local Motorola Sales Office
- World Wide Web at http://www.mcu.motsps.com

Follow World Wide Web on-line instructions to retrieve the current mechanical specifications.

MC68HC05C9A — Rev. 5.0

13.3 40-Pin Plastic Dual In-Line (DIP) Package (Case 711-03)

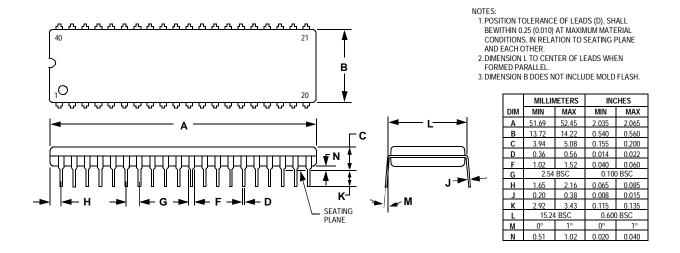


Figure 13-1. 40-Pin Plastic DIP Package (Case 711-03)

13.4 42-Pin Plastic Shrink Dual In-Line (SDIP) Package (Case 858-01)

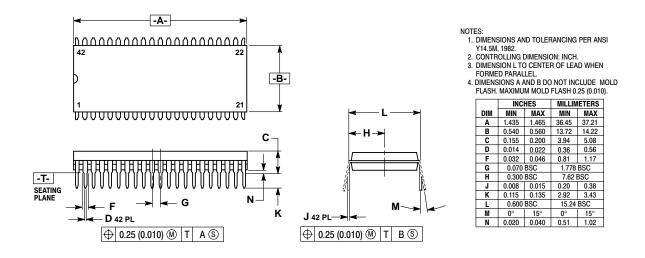


Figure 13-2. 42-Pin Plastic SDIP Package (Case 858-01)

13.5 44-Lead Plastic-Leaded Chip Carrier (PLCC) (Case 777-02)

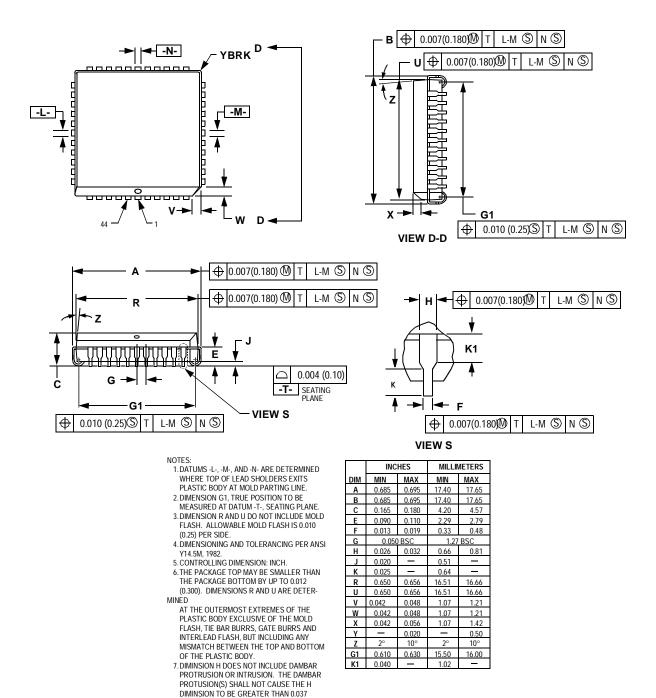


Figure 13-3. 44-Lead PLCC (Case 777-02)

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MC68HC05C9A — Rev. 5.0

13.6 44-Lead Quad Flat Pack (QFP) (Case 824A-01)

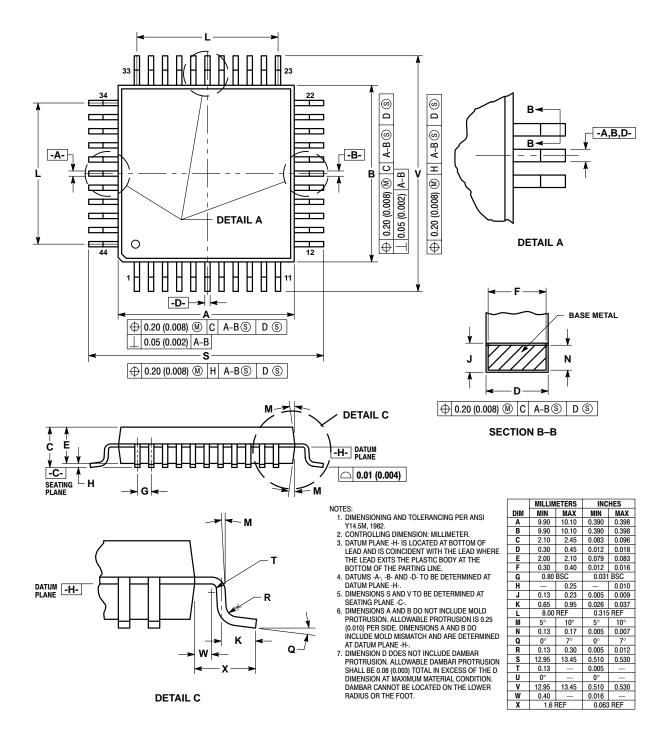


Figure 13-4. 44-Lead QFP (Case 824A-01)

Advance Information MC68HC05C9A — Rev. 5.0

Section 14. Ordering Information

14.1 Contents

14.2	Introduction	41
14.3	MC Order Numbers	41

14.2 Introduction

This section contains ordering information for the available package types.

14.3 MC Order Numbers

Table 14-1 shows the MC order numbers for the available package types.

Table 14-1. MC Order Numbers

Package Type	Temperature Range	Order Number
40-pin plastic dual in-line	0°C to 70°C	MC68HC05C9AP
package (DIP)	–40°C to 125°C	MC68HC05C9AMP
42-pin shrink dual in-line	0°C to 70°C	MC68HC05C9AB
package (SDIP)	-40°C to 125°C	MC68HC05C9AMB
44-lead plastic-leaded	0°C to 70°C	MC68HC05C9AFN
chip carrier (PLCC)	-40°C to 125°C	MC68HC05C9AMFN
44-pin quad flat pack	0°C to 70°C	MC68HC05C9AFB
(QFP)	-40°C to 125°C	MC68HC05C9AMFB

- 1. P = Plastic dual in-line package (PDIP)
- 2. B = Shrink dual in-line package (SDIP)
- 3. FN = Plastic-leaded chip carrier (PLCC)
- 4. FB = Quad flat pack (QFP)

MC68HC05C9A — Rev. 5.0

Ordering Information

Advance Information MC68HC05C9A — Rev. 5.0

Appendix A. MC68HCL05C9A

A.1 Contents

A.2	Introduction
A.3	Operating Temperature143
A.4	DC Electrical Characeristics
A.4.1	1.8–2.4-Volt Low-Power Output Voltage
A.4.2	1.8–2.4-Volt Input Pullup Current144
A.4.3	2.5–3.6-Volt Low-Power Output Voltage
A.4.4	2.6–3.6-Volt Input Pullup Current145
A.4.5	Low-Power Supply Current146

A.2 Introduction

Appendix A describes the MC68HCL05C9A, a low-power version of the MC68HC05C9A. The technical data applying to the MC68HC05C9A applies to the MC68HCL05C9 with the exceptions given in this appendix.

A.3 Operating Temperature

The data shown here replaces the corresponding data in **12.3 Operating Temperature**.

Rating	Symbol	Value	Unit
Operating temperature range MC68HCL05C9AP, FN, B, FB ⁽¹⁾	T _A	T _L to T _H 0 to +70	°C

1. P = Plastic dual in-line package (PDIP) FN = Plastic-leaded chip carrier (PLCC)

B = Shrink dual in-line package (SDIP)

FB = Quad flat pack (QFP)

MC68HC05C9A — Rev. 5.0

A.4 DC Electrical Characeristics

The data in 12.6 5.0-Volt DC Electrical Characteristics and 12.7 3.3-Volt DC Electrical Characteristics applies to the MC68HCL05C9A with the exceptions given here.

A.4.1 1.8-2.4-Volt Low-Power Output Voltage

Characteristic ⁽¹⁾	Symbol	Min	Тур	Max	Unit
Output high voltage $ (I_{Load} = -0.1 \text{ mA}) \text{ PA7-PA0}, \text{ PB7-PB0}, \text{ PC6-PC0}, \\ \text{TCMP, PD7, PD0} \\ (I_{Load} = -0.2 \text{ mA}) \text{ PD5-PD1} \\ (I_{Load} = -0.75 \text{ mA}) \text{ PC7} $	V _{OH}	V _{DD} -0.3 V _{DD} -0.3 V _{DD} -0.3	111		V
Output low voltage (I _{Load} = 0.2 mA) PA7–PA0, PB7–PB0, PC6–PC0, PD7, PD5–PD0, TCMP (I _{Load} = 2.0 mA) PC7	V _{OL}		_ _	0.3 0.3	V

^{1.} $V_{DD} = 1.8-2.4 \text{ Vdc}$

A.4.2 1.8–2.4-Volt Input Pullup Current

Characteristic ⁽¹⁾	Symbol	Min	Тур	Max	Unit
Input pullup current PB7-PB0 (with pullup)	I _{In}	0.45	1.5	6.5	μА

^{1.} $V_{DD} = 1.8-2.4 \text{ Vdc}$

A.4.3 2.5–3.6-Volt Low-Power Output Voltage

Characteristic	Symbol	Min	Тур	Max	Unit
Output high voltage ⁽¹⁾ $(I_{Load} = -0.2 \text{ mA}) \text{ PA7-PA0, PB7-PB0, PC6-PC0,}$ TCMP, PD7, PD0 $(I_{Load} = -0.4 \text{ mA}) \text{ PD5-PD1}$ $(I_{Load} = -1.5 \text{ mA}) \text{ PC7}$	V _{ОН}	V _{DD} - 0.3 V _{DD} - 0.3 V _{DD} - 0.3	111	111	V
Output low voltage (I _{LOAD} = 0.4 mA) PA7–PA0, PB7–PB0, PC6–PC0, PD7, PD5–PD0, TCMP (I _{LOAD} = 5.0 mA) PC7	V _{OL}		_ _	0.3 0.3	V

^{1.} $V_{DD} = 2.5-3.6 \text{ Vdc}$

A.4.4 2.6-3.6-Volt Input Pullup Current

Characteristic ⁽¹⁾	Symbol	Min	Тур	Max	Unit
Input pullup current PB7-PB0 (with pullup)	I _{In}	1	5	16	μА

^{1.} $V_{DD} = 2.5-3.6 \text{ Vdc}$

A.4.5 Low-Power Supply Current

Characteristic ⁽¹⁾	Symbol	Min	Тур	Max	Unit
Supply current (4.5–5.5 Vdc @ f _{Bus} = 2.1 MHz) Run ⁽²⁾ Wait ⁽³⁾ Stop ⁽⁴⁾ 25°C 0°C to +70°C (standard)	I _{DD}	_ _ _ _	3.5 1.6 1 2	4.25 2.25 15 25	mA mA μA μA
Supply current (2.4–3.6 Vdc @ f _{Bus} = 1.0 MHz) Run ⁽²⁾ Wait ⁽³⁾ Stop ⁽⁴⁾ 25 °C 0 °C to +70 °C (standard)	I _{DD}	_ _ _ _	1.00 0.7 1 1	1.4 1.0 5 10	mA mA μΑ μΑ
Supply current (2.5–3.6 Vdc @ f _{Bus} = 500 kHz) Run ⁽²⁾ Wait ⁽³⁾ Stop ⁽⁴⁾ 25 °C 0 °C to +70 °C (standard)	I _{DD}	_ _ _ _	500 300 1 1	750 500 5 10	μΑ μΑ μΑ μΑ
Supply current (1.8–2.4 Vdc @ f _{Bus} = 500 kHz) Run ⁽²⁾ Wait ⁽³⁾ Stop ⁽⁴⁾ 25 °C 0 °C to +70 °C (standard)	I _{DD}		300 250 1 1	600 400 2 5	μΑ μΑ μΑ μΑ

- 1. Typical values reflect measurements taken on average processed devices at the midpoint of voltage range, 25°C only.
- 2. Run (operating) I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} –0.2 V; no DC loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2
- 3. Wait I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} -0.2 V; no DC loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2. Wait I_{DD} is affected linearly by the OSC2 capacitance.
- 4. Stop I_{DD} measured with OSC1 = 0.2 V; all I/O pins configured as inputs, Port B = V_{DD} , all other inputs V_{IL} = 0.2 V, $V_{IH} = V_{DD}$ -0.2 V

Appendix B. MC68HSC05C9A

B.1 Contents

B.2	Introduction
B.3	Operating Temperature148
B.4 B.4.1 B.4.2	DC Electrical Characeristics
B.5 B.5.1 B.5.2 B.5.3	Control Timing
B.5.4	2.4–3.6-Volt High-Speed SPI Timing

B.2 Introduction

Appendix B describes the MC68HSC05C9A, a high-speed version of the MC68HC05C9A. The technical data applying to the MC68HC05C9A applies to the MC68HSC05C9A with the exceptions given in this appendix.

MC68HC05C9A — Rev. 5.0

B.3 Operating Temperature

The data shown here replaces the corresponding data in **12.3 Operating Temperature**.

Table B-1. High-Speed Operating Temperature Range

Rating	Symbol	Value	Unit
Operating temperature range ⁽¹⁾ MC68HSC05C9AP, FN, B, FB MC68HSC05C9ACP, CFN, CB, CFB	T _A	T _L to T _H 0 to +70 –40 to +85	°C

^{1.} P = Plastic dual in-line package (PDIP)

Advance Information MC68HC05C9A — Rev. 5.0

FN = Plastic-leaded chip carrier (PLCC)

C = Extended temperature range (-40° C to $+85^{\circ}$ C)

B = Shrink dual in-line package (SDIP)

FB = Quad flat pack (QFP)

B.4 DC Electrical Characeristics

The data in 12.6 5.0-Volt DC Electrical Characteristics and 12.7 3.3-Volt DC Electrical Characteristics applies to the MC68HSC05C9A with the exceptions given here.

B.4.1 High-Speed Supply Current

Characteristic ⁽¹⁾	Symbol	Min	Тур	Max	Unit
Supply current (4.5–5.5 Vdc @ f _{Bus} = 4.0 MHz)					
Run ⁽²⁾		_	7.00	11.0	mA
Wait ⁽³⁾	ı	_	2.00	6.50	mA
Stop ⁽⁴⁾ 25°C	I _{DD}	_	1	20	μА
0°C to 70°C (standard)		_	1.0	40	μΑ
-40°C to 85°C (standard)		_	7.0	50	μΑ
Supply current (2.4–3.6 Vdc @ f _{Bus} = 2.0 MHz)					
Run ⁽²⁾			2.50	4.00	mA
Wait ⁽³⁾			1.00	2.00	mA
Stop ⁽⁴⁾	I _{DD}				
25°C		_	1	8	μA
0°C to 70°C (standard)		_	1.0	16	μΑ
-40°C to 85°C (standard)		_	2.5	20	μΑ

- 1. Typical values reflect measurements taken on average processed devices at the midpoint of voltage range, 25°C only.
- 2. Run (operating) I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} -0.2 V; no dc loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2
- 3. Wait I_{DD} measured using external square wave clock source; all I/O pins configured as inputs, Port B = V_{DD} , all other inputs V_{IL} = 0.2 V, V_{IH} = V_{DD} -0.2 V; no dc loads; less than 50 pF on all outputs; C_L = 20 pF on OSC2. Wait I_{DD} is affected linearly by the OSC2 capacitance
- 4. Stop I_{DD} measured with OSC1 = 0.2 V; all I/O pins configured as inputs, port B = V_{DD} , all other inputs V_{IL} = 0.2 V, $V_{IH} = V_{DD}$ -0.2 V

B.4.2 Input Pullup Current

Characteristic	Symbol	Min	Тур	Max	Unit
Input pullup current (V _{DD} = 4.5–5.5 V) PB7–PB0 (with pullup)	I _{In}	5	15	60	μА
Input pullup current (V _{DD} = 2.4–3.6 V) PB7–PB0 (with pullup)	I _{In}	1	5	16	μА

MC68HC05C9A — Rev. 5.0

B.5 Control Timing

The data in 12.8 5.0-Volt Control Timing and 12.9 3.3-Volt Control Timing applies to the MC68HSC05C9A with the exceptions given here.

B.5.1 4.5-5.5-Volt High-Speed Control Timing

Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
Oscillator frequency Crystal External clock	fosc	— dc	8.2 8.2	MHz
Internal operating frequency (f _{OSC} ÷ 2) Crystal External clock	f _{OP}	— dc	4.1 4.1	MHz
Cycle time	t _{cyc}	244	_	ns
Crystal oscillator startup time	t _{OXOV}		100	ms
Stop recovery startup time	t _{ILCH}		100	ms
RESET pulse width	t _{RL}	1.5	_	t _{cyc}
Timer Resolution ⁽²⁾ Input capture pulse width Input capture pulse width	t _{RESL} t _{TH} or t _{TL} t _{THT} I	4.0 64 Note ⁽³⁾		t _{cyc} ns t _{cyc}
Interrupt pulse width low (edge-triggered)	t _{ILIH}	64	_	ns
Interrupt pulse period	t _{ILIL}	Note ⁽⁴⁾	_	t _{cyc}
OSC1 pulse width	t _{OH} or t _{OL}	50	_	ns

^{1.} $V_{DD} = 4.5 - 5.5 \text{ Vdc}$

^{2.} Because a 2-bit prescaler in the timer must count four internal cycles (t_{cyc}), this is the limiting minimum factor in determining the timer resolution.

^{3.} The minimum period t_{TLTL} should not be less than the number of cycle times it takes to execute the capture interrupt service routine plus 24 t_{cyc} .

^{4.} The minimum t_{ILIL} should not be less than the number of cycle times it takes to execute the interrupt service routine plus 19 t_{cvc} .

B.5.2 2.4–3.6-Volt High-Speed Control Timing

Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
Oscillator frequency Crystal External Clock	fosc	— dc	4.2 4.2	MHz
Internal operating frequency (f _{OSC} ÷ 2) Crystal External clock	f _{OP}	— dc	2.1 2.1	MHz
Cycle time	t _{cyc}	480	_	ns
Crystal oscillator startup time	t _{OXOV}	_	100	ms
Stop recovery startup time	t _{ILCH}	_	100	ms
RESET pulse width	t _{RL}	1.5	_	t _{cyc}
Timer Resolution ⁽²⁾ Input capture pulse width Input capture pulse width	t _{RESL} t _{TH} or t _{TL} t _{THTL}	4.0 125 Note ⁽³⁾	_ _ _	t _{cyc} ns t _{cyc}
Interrupt pulse width low (edge-triggered)	t _{ILIH}	125	_	ns
Interrupt pulse period	t _{ILIL}	Note ⁽⁴⁾	_	t _{cyc}
OSC1 pulse width	t _{OH} or t _{OL}	90	_	ns

^{1.} $V_{DD} = 2.4-3.6 \text{ Vdc}$

^{2.} Because a 2-bit prescaler in the timer must count four internal cycles (t_{cyc}) , this is the limiting minimum factor in determining the timer resolution.

 $^{3. \} The \ minimum \ period \ t_{TLTL} \ should \ not \ be \ less \ than \ the \ number \ of \ cycle \ times \ it \ takes \ to \ execute \ the \ capture \ interrupt \ service$

routine plus 24 t_{cyc}.

4. The minimum t_{ILIL} should not be less than the number of cycle times it takes to execute the interrupt service routine plus 19 t_{cyc}.

B.5.3 4.5–5.5-Volt High-Speed Control Timing

Num	Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
	Operating frequency Master Slave	f _{OP(M)} f _{OP(S)}	dc dc	0.5 4.1	f _{OP} MHz
1	Cycle time Master Slave	t _{cyc(M)}	2.0 244		t _{cyc} ns
2	Enable lead time Master Slave	t _{Lead(M)}	Note ⁽²⁾ 122	_	ns
3	Enable lag time Master Slave	t _{Lag(M)} t _{Lag(S)}	Note ⁽²⁾ 366		ns
4	Clock (SCK) high time Master Slave	tw(sckh)m tw(sckh)s	166 93		ns
5	Clock (SCK) low time Master Slave	tw(sckl)m tw(sckl)s	166 93		ns
6	Data setup time (inputs) Master Slave	t _{SU(M)}	49 49		ns
7	Data hold time (inputs) Master Slave	t _{H(M)}	49 49	_	ns
8	Slave access time (time to data active from high-impedance state)	t _A	0	61	ns
9	Slave disable time (hold time to high-impedance state)	t _{DIS}	_	122	ns
10	Data valid Master (before capture edge) Slave (after enable edge) ⁽³⁾	t _{V(M)}	0.25 —	_ 122	t _{cyc(M)}
11	Data hold time (outputs) Master (after capture edge) Slave (after enable edge)	t _{HO(M)}	0.25 0		t _{cyc(M)} ns
12	Rise time (20% V _{DD} to 70% V _{DD} , C _L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{RM}		50 1.0	ns μs
13	Fall time (70% V _{DD} to 20% V _{DD} , C _L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{FM}	_	50 1.0	ns μs

^{1.} $V_{DD} = 4.5 - 5.5 \text{ Vdc}$

^{2.} Signal production depends on software.

^{3.} Assumes 200 pF load on all SPI pins

B.5.4 2.4-3.6-Volt High-Speed SPI Timing

Num	Characteristic ⁽¹⁾	Symbol	Min	Max	Unit
	Operating frequency Master Slave	f _{OP(M)} f _{OP(S)}	dc dc	0.5 2.1	f _{OP} MHz
1	Cycle time Master Slave	t _{cyc(M)}	2.0 480	_ _	t _{cyc} ns
2	Enable lead time Master Slave	t _{Lead(M)}	Note ⁽²⁾ 240	_	ns
3	Enable lag time Master Slave	t _{Lag(M)} t _{Lag(S)}	Note ⁽²⁾ 720		ns
4	Clock (SCK) high time Master Slave	tw(sckh)m tw(sckh)s	340 190	_ _	ns
5	Clock (SCK) low time Master Slave	tw(sckl)m tw(sckl)s	340 190		ns
6	Data setup time (inputs) Master Slave	t _{SU(M)}	100 100		ns
7	Data hold time (inputs) Master Slave	t _{H(M)} t _{H(S)}	100 100		ns
8	Slave access time (time to data active from high-impedance state)	t _A	0	120	ns
9	Slave disable time (hold time to high-impedance state)	t _{DIS}	_	240	ns
10	Data valid Master (before capture edge) Slave (after enable edge) ⁽³⁾	t _{V(M)} t _{V(S)}	0.25 —	 240	t _{cyc(M)} ns
11	Data hold time (outputs) Master (after capture edge) Slave (after enable edge)	t _{HO(M)}	0.25 0		t _{cyc(M)}
12	Rise time (20% V _{DD} to 70% V _{DD} , C _L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{RM}		100 2.0	ns μs
13	Fall time (70% V _{DD} to 20% V _{DD} , C _L = 200 pF) SPI outputs (SCK, MOSI, and MISO) SPI inputs (SCK, MOSI, MISO, and SS)	t _{FM}	_	100 2.0	ns µs

^{1.} $V_{DD} = 2.4-3.6 \text{ Vdc}$

^{2.} Signal production depends on software.

^{3.} Assumes 200 pF load on all SPI pins

MC68HSC05C9A

Appendix C. Self-Check Mode

C.1 Contents

C.2	Introduction	.155
C.3	Self-Check Mode	.156
C.3.1	Self-Check Tests	.156
C.3.2	Self-Check Results	.158

C.2 Introduction

This appendix describes the self-check mode.

MC68HC05C9A — Rev. 5.0

C.3 Self-Check Mode

Self-check mode is entered upon the rising edge of \overline{RESET} if the \overline{IRQ} pin is at V_{tst} and the TCAP pin is at logic 1.

C.3.1 Self-Check Tests

The self-check read-only memory (ROM) at mask ROM location \$3F00–\$3FEF determines if the microcontroller unit (MCU) is functioning properly.

These tests are performed:

- Input/output (I/O) Functional test of ports A, B, and C
- Random-access memory (RAM) Counter test for each RAM byte
- 3. Timer Test of counter register and OCF bit
- 4. Serial communications interface (SCI) Transmission test; checks for RDRF, TDRE, TC, and FE flags
- 5. ROM Exclusive OR with odd ones parity result
- Serial peripheral interface (SPI) Transmission test; checks for SPIF and WCOL flags

The self-check circuit is shown in Figure C-1.

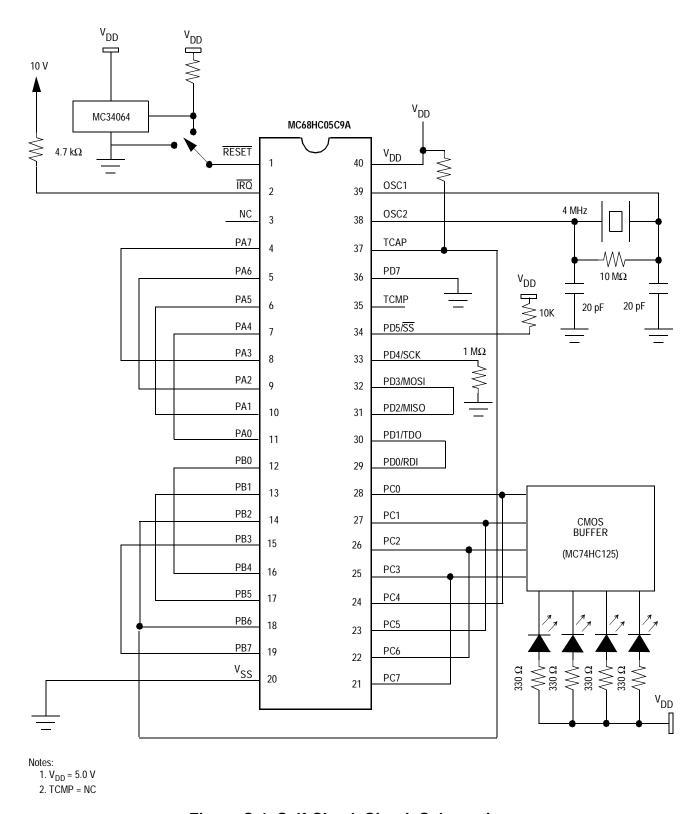


Figure C-1. Self-Check Circuit Schematic

MC68HC05C9A — Rev. 5.0 Advance Information

C.3.2 Self-Check Results

Table C-1 shows the LED codes that indicate self-check test results.

PC3 PC₂ PC1 PC0 Remarks Off On On Off I/O failure RAM failure Off Off On On Off On Off Off Timer failure Off Off On SCI failure On Off Off On Off ROM failure Off SPI failure Off Off On Flashing No failure All others Device failure

Table C-1. Self-Check Circuit LED Codes

Perform these steps to activate the self-check tests:

- 1. Apply 10 V (2 x V_{DD}) to the \overline{IRQ} pin.
- 2. Apply a logic 1 to the TCAP pin.
- 3. Apply a logic 0 to the $\overline{\mathsf{RESET}}$ pin.

The self-check tests begin on the rising edge of the RESET pin.

RESET must be held low for 4064 cycles after power-on reset (POR), or for a time, t_{RL} , for any other reset. For the value of t_{RL} , see 12.8 5.0-Volt Control Timing and 12.9 3.3-Volt Control Timing.

Appendix D. M68HC05Cx Family Feature Comparisons

Refer to **Table D-1** for a comparison of the features for all the M68HC05C Family members.

Table D-1. M68HC05Cx Feature Comparison

anc		C4	C4A	705C4A	C8	C8A	705C8	705C8A	C12	C12A	С9	C9A	705C9	705C9A
ë [USER ROM	4160	4160	_	7744	7744	-	_	12,096	12,096	15,760–15,936	15,760–15,936	_	-
of [USER EPROM	_	_	4160	_	_	7596–7740	7596–7740	_	_	_	_	15,760-15,936	12,096-15,936
ance Information	CODE SECURITY	NO	YES	YES	NO	YES	YES	YES	NO	YES	NO	YES	NO	YES
S	RAM	176	176	176	176	176	176–304	176–304	176	176	176–352	176–352	176-352	176-352
	OPTION REGISTER (IRQ/RAM/ SEC)	NO	NO	\$1FDF (IRQ/SEC)	NO	NO	\$1FDF (IRQ/RAM/ SEC)	\$1FDF (IRQ/RAM/SEC)	NO	NO	\$3FDF (IRQ/RAM)	\$3FDF (IRQ/RAM)	\$3FDF (IRQ/RAM)	\$3FDF (IRQ/RAM)
	MASK OPTION REGISTER(S)	NO	NO	\$1FF0-\$1FF1	NO	NO	NO	\$1FF0-\$1FF1	NO	NO	NO	NO	NO	\$3FF0-\$3FF1
	PORTB KEYSCAN (PULLUP/ INTERRUPT)	NO	YES MASK OPTION	YES MOR SELECT- ABLE	NO	YES MASK OPTION	NO	YES MOR SELECTABLE	YES MASK OPTION	YES MASK OPTION	NO	YES MASK OPTION	NO	YES MOR SELECTABLE
	PC7 DRIVE	STANDARD	HIGH CURRENT	HIGH CURRENT	STANDARD	HIGH CURRENT	STANDARD	HIGH CURRENT	HIGH CURRENT	HIGH CURRENT	STANDARD	HIGH CURRENT	STANDARD	HIGH CURRENT
	PORT D	PD7, 5–0 INPUT ONLY	PD7, 5-0 INPUT ONLY	PD7, 5–0 INPUT ONLY	PD7, 5–0 INPUT ONLY	PD7, 5–0 INPUT ONLY	PD7, 5-0 INPUT ONLY	PD7, 5-0 INPUT ONLY	PD7, 5–0 INPUT ONLY	PD7, 5-0 INPUT ONLY	PD7, 5-0 BIDIREC- TIONAL	PD7, 5-0 BIDIREC- TIONAL	PD7, 5-0 BIDIREC- TIONAL	PD7, 5-0 BIDIRECTIONAL
	COP	NO	YES	YES	NO	YES	YES	TWO TYPES	YES	YES	YES	YES	YES	TWO TYPES
	COP ENABLE	_	MASK OPTION	MOR	١	MASK OPTION	SOFTWARE	SOFTWARE+ MOR	MASK OPTION	MASK OPTION	SOFTWARE	SOFTWARE	SOFTWARE	SOFTWARE+ MOR
	COP TIMEOUT	_	64 ms (@4 MHz OSC)	64 ms (@4 MHz OSC)	-	64 ms (@4 MHz OSC)	SOFTWARE SELECTABLE	SOFTWARE+ MOR SELECTABLE	64 ms (@4 MHz OSC)	64 ms (@4MHz OSC)	SOFTWARE SELECTABLE	SOFTWARE SELECTABLE	SOFTWARE SELECTABLE	SOFTWARE+ MOR SELECTABLE
	COP CLEAR	_	CLR \$1FF0	CLR \$1FF0	_	CLR \$1FF0	WRITE \$55/\$AA TO \$001D	WRITE \$55/\$AA TO \$001D OR CLR \$1FF0	CLR \$3FF0	CLR \$3FF0	WRITE \$55/\$AA TO \$001D	WRITE \$55/\$AA TO \$001D	WRITE \$55/\$AA TO \$001D	WRITE \$55/\$AA TO \$001D OR CLR \$3FF0
M	CLOCK MONITOR	NO	NO	NO	NO	NO	YES	YES	NO	NO	YES	YES	YES	YES (C9A MODE)
МС68НС05С9А	ACTIVE RESET	NO	NO	NO	NO	NO	COP/CLOCK MONITOR	PROGRAM- MABLE COP/CLOCK MONITOR	NO	NO	POR/COP/ CLOCK MONITOR	POR/COP/ CLOCK MONITOR	POR/COP/ CLOCK MONITOR	POR/C9A COP/ CLOCK MONITOR
C9A -	STOP DISABLE	NO	MASK OPTION	NO	NO	MASK OPTION	NO	NO	MASK OPTION	MASK OPTION	NO	NO	NO	MOR SELECTABLE (C12A MODE)

^{1.} The expanded RAM map (from \$30-\$4F and \$100-\$15F) available on the OTP devices MC68HC705C8 and MC68HC705C8A is not available on the ROM devices MC68HC05C8A.

2. The programmable COP available on the MC68HC705C8 and MC68HC705C8A is not available on the MC68HC05C8A. For ROM compatibility, use the non-programmable COP.

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