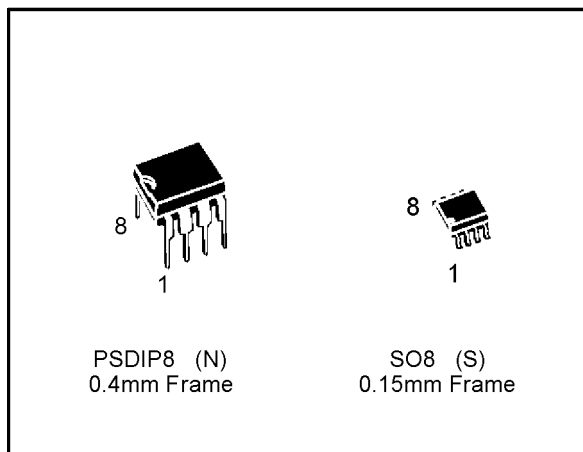


CMOS 64 x 8 SERIAL ACCESS TIMEKEEPER SRAM

- COUNTERS for SECONDS, MINUTES, HOURS, DAY, DATE, MONTH and YEARS
- SOFTWARE CLOCK CALIBRATION
- AUTOMATIC POWER FAIL DETECT and SWITCH CIRCUITRY
- I²C BUS COMPATIBLE
- 56 BYTES of GENERAL PURPOSE RAM
- ULTRA-LOW BATTERY SUPPLY CURRENT of 500nA
- AVAILABLE with an OPERATING TEMPERATURE of -40 to 85°C
- AUTOMATIC LEAP YEAR COMPENSATION



DESCRIPTION

The MK41T56 TIMEKEEPER™ RAM is a low power 512 bit static CMOS RAM organized as 64 words by 8 bits. A built-in 32.768 kHz oscillator (external crystal controlled) and the first 8 bytes of the RAM are used for the clock/calendar function and are configured in BCD format. Addresses and data are transferred serially via a two-line bi-directional bus. The built-in address register is incremented automatically after each write or read data byte. The MK41T56 clock has a built-in power sense circuit which detects power failures and automatically switches to the battery supply during power failures. The energy needed to sustain the RAM and clock operations can be supplied from a small lithium button cell.

Table 1. Signal Names

OSCI	Oscillator Input
OCSO	Oscillator Output
FT/OUT	Frequency Test / Output Driver
SDA	Serial Data Address Input / Output
SCL	Serial Clock
V _{BAT}	Battery Supply Voltage
V _{CC}	Supply Voltage
V _{SS}	Ground

Figure 1. Logic Diagram

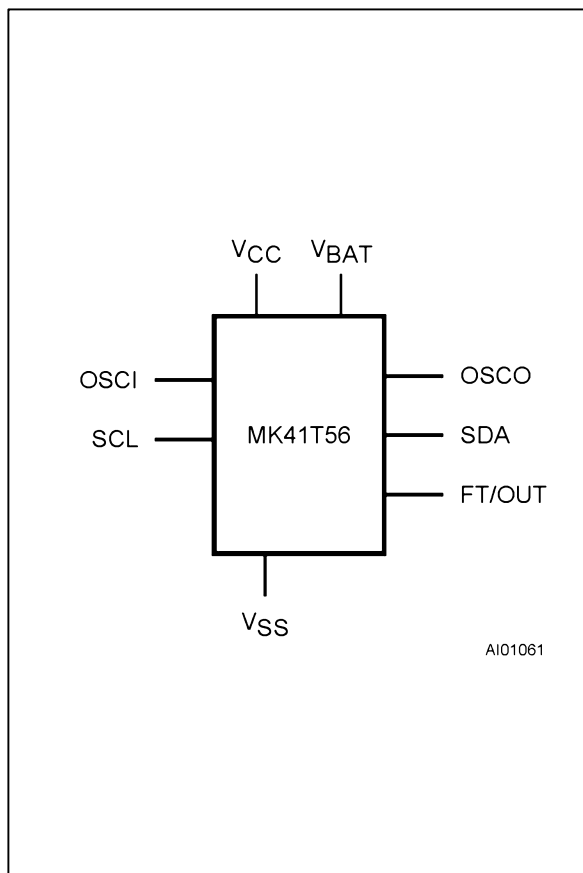


Figure 2A. DIP Pin Connections

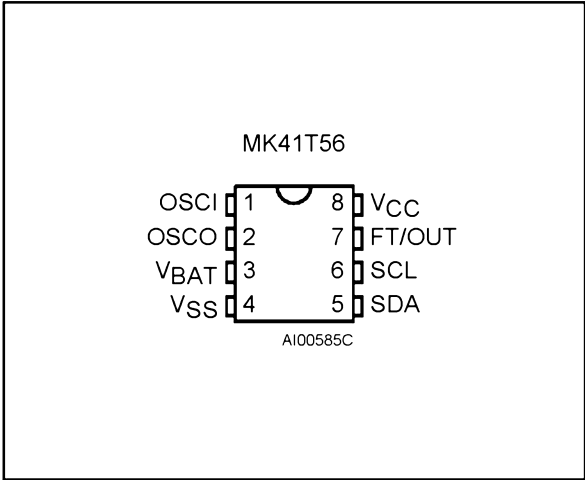


Figure 2B. SO Pin Connections

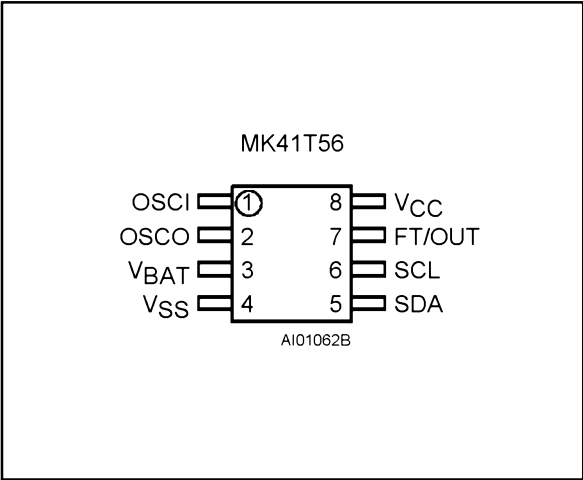


Table 2. Absolute Maximum Ratings

Symbol	Parameter	Value	Unit
T_A	Ambient Operating Temperature	0 to 70 -40 to 85	°C
T_{STG}	Storage Temperature (V_{CC} Off, Oscillator Off)	-55 to 125	°C
V_{IO}	Input or Output Voltages	-0.3 to 7	V
V_{CC}	Supply Voltage	-0.3 to 7	V
I_O	Output Current	20	mA
P_D	Power Dissipation	0.25	W

Note: Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to the absolute maximum ratings conditions for extended periods of time may affect reliability.

CAUTION: Negative undershoots below -0.3 volts are not allowed on any pin while in the Battery Back-up mode.

Table 3. Register Map

Address	Data								Function/Range BCD Format	
	D7	D6	D5	D4	D3	D2	D1	D0		
0	ST	10 Seconds			Seconds				Seconds	00-59
1	X	10 Minutes			Minutes				Minutes	00-59
2	X	X	10 Hours		Hours				Hour	00-23
3	X	X	X	X	X	Day			Day	01-07
4	X	X	10 Date		Date				Date	01-31
5	X	X	X	10 M.	Month				Month	01-12
6	10 Years				Years				Year	00-99
7	OUT	FT	S	Calibration					Control	

Keys: S = SIGN Bit; FT = FREQUENCY TEST Bit; ST = STOP Bit; OUT = Output level; X = Don't care.

Table 4. Capacitance⁽¹⁾ ($T_A = 25^\circ\text{C}$, $f = 1\text{ MHz}$)

Symbol	Parameter	Min	Max	Unit
C_{IN}	Input Capacitance (SCL)		7	pF
$C_{OUT}^{(2)}$	Output Capacitance (SDA, FT/OUT)		10	pF

Notes: 1. Effective capacitance calculated from the equation $C = I\Delta t/\Delta V$ with $\Delta V = 3V$ and power supply at 5V.

2. Outputs deselected.

Table 5. DC Characteristics ($T_A = 0$ to 70°C or -40 to 85°C ; $V_{CC} = 4.5V$ to $5.5V$)

Symbol	Parameter	Test Condition	Min	Typ	Max	Unit
I_{LI}	Input Leakage Current	$0V \leq V_{IN} \leq V_{CC}$			± 10	μA
I_{LO}	Output Leakage Current	$0V \leq V_{OUT} \leq V_{CC}$			± 10	μA
I_{CC1}	Supply Current	SCL/SDA = $V_{CC}-0.3V$			1	mA
I_{CC2}	Supply Current (Standby)				1	mA
V_{IL}	Input Low Voltage		-0.3		1.5	V
V_{IH}	Input High Voltage		3		$V_{CC} + 0.8$	V
V_{OL}	Output Low Voltage	$I_{OL} = 5\text{mA}$, $V_{CC} = 4.5V$			0.4	V
$V_{BAT}^{(1)}$	Battery Supply Voltage		2.6	3	3.5	V
I_{BAT}	Battery Supply Current	$T_A = 25^\circ\text{C}$, $V_{CC} = 0V$, Oscillator ON, $V_{BAT} = 3V$		450	500	nA

Note: SGS-THOMSON recommends the RAYOVAC BR1225 or equivalent as the battery supply.

Table 6. Power Down/Up Trip Points DC Characteristics⁽¹⁾ ($T_A = 0$ to 70°C or -40 to 85°C)

Symbol	Parameter	Min	Typ	Max	Unit
V_{PFD}	Power-fail Deselect Voltage	$1.2 V_{BAT}$	$1.25 V_{BAT}$	$1.285 V_{BAT}$	V
V_{SO}	Battery Back-up Switchover Voltage		V_{BAT}		V

Note: 1. All voltages referenced to V_{SS} .

Table 7. Crystal Electrical Characteristics (Externally Supplied)

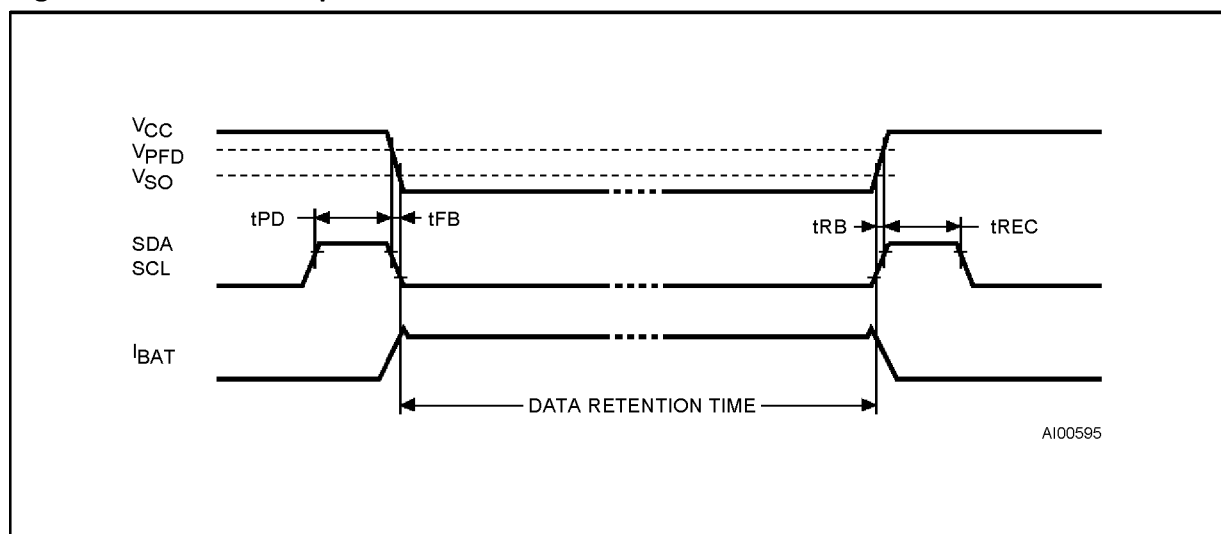
Symbol	Parameter	Min	Typ	Max	Unit
f_0	Resonant Frequency		32.768		kHz
R_S	Series Resistance			35	$k\Omega$
C_L	Load Capacitance		12.5		pF

Notes: Load capacitors are internally supplied with the MK41T56. Circuit board layout considerations for the 32.768 kHz crystal of minimum trace lengths and isolation from RF generating signals should be taken into account.

SGS-THOMSON recommends the ECS-.327-12.5-8SP-2 quartz crystal for industrial temperature operations. ESC Inc. can be contacted at 800-237-1041 or 913-782-7787 for further information on this crystal type.

Table 8. Power Down/Up Mode AC Characteristics ($T_A = 0$ to 70°C or -40 to 85°C)

Symbol	Parameter	Min	Max	Unit
t_{PD}	SCL and SDA at V_{IH} before Power Down	0		ns
t_{FB}	V_{PFD} (min) to V_{SO} V_{CC} Fall Time	300		μs
t_{RB}	V_{SO} to V_{PFD} (min) V_{CC} Rise Time	100		μs
t_{REC}	SCL and SDA at V_{IH} after Power Up	200		μs

Figure 5. Power Down/Up Mode AC Waveforms**OPERATION (cont'd)**

The MK41T56 clock continually monitors V_{CC} for an out of tolerance condition. Should V_{CC} fall below V_{PFD} the device terminates an access in progress and resets the device address counter. Inputs to the device will not be recognized at this time to prevent erroneous data from being written to the device from an out of tolerance system. When V_{CC} falls below V_{BAT} the device automatically switches over to the battery and powers down into an ultra low current mode of operation to conserve battery life. Upon power up the device switches from battery to V_{CC} at V_{BAT} and recognizes inputs when V_{CC} goes above V_{PFD} volts.

2-WIRE BUS CHARACTERISTICS

This bus is intended for communication between different ICs. It consists of two lines: one bi-directional for data signals (SDA) and one for clock signals (SCL). Both the SDA and the SCL lines must be connected to a positive supply voltage via a pull-up resistor.

The following protocol has been defined:

- Data transfer may be initiated only when the bus is not busy.
- During data transfer, the data line must remain stable whenever the clock line is High. Changes in the data line while the clock line is High will be interpreted as control signals.

Table 9. AC Characteristics ($T_A = 0$ to 70°C or -40 to 85°C ; $V_{CC} = 4.5\text{V}$ to 5.5V)

Symbol	Parameter	Min	Max	Unit
f_{SCL}	SCL Clock Frequency	0	100	kHz
t_{LOW}	Clock Low Period	4.7		μs
t_{HIGH}	Clock High Period	4		μs
t_{R}	SDA and SCL Rise Time		1	μs
t_{F}	SDA and SCL Fall Time		300	ns
$t_{\text{HD STA}}$	START Condition Hold Time (after this period the first clock pulse is generated)	4		μs
$t_{\text{SU STA}}$	START Condition Setup Time (only relevant for a repeated start condition)	4.7		μs
$t_{\text{SU DAT}}^{(1)}$	Data Setup Time	250		ns
$t_{\text{HD DAT}}$	Data Hold Time	0		μs
$t_{\text{SU STO}}$	STOP Condition Setup Time	4.7		μs
t_{BUF}	Time the bus must be free before a new transmission can start	4.7		μs
t_i	Noise suppression time constant at SCL and SDA input	0.25	1	μs

Note: 1. Transmitter must internally provide a hold time to bridge the undefined region (300ns max.) of the falling edge of SCL.

2-WIRE BUS CHARACTERISTICS (cont'd)

Accordingly, the following bus conditions have been defined:

Bus not busy. Both data and clock lines remain High.

Start data transfer. A change in the state of the data line, from High to Low, while the clock is High, defines the START condition.

Stop data transfer. A change in the state of the data line, from Low to High, while the clock is High, defines the STOP condition.

Data valid. The state of the data line represents valid data when after a start condition, the data line is stable for the duration of the High period of the clock signal. The data on the line may be changed during the Low period of the clock signal. There is one clock pulse per bit of data.

Each data transfer is initiated with a start condition and terminated with a stop condition. The number of data bytes transferred between the start and stop conditions is not limited. The information is transmitted byte-wide and each receiver acknowledges with a ninth bit.

Within the bus specifications a low speed mode (2kHz clock rate) and a high speed mode (100kHz clock rate) are defined. The MK41T56 clock works in both modes. By definition a device that gives out a message is called "transmitter", the receiving device that gets the message is called "receiver". The device that controls the message is called "master". The devices that are controlled by the master are called "slaves".

Acknowledge. Each byte of eight bits is followed by one acknowledge bit. This acknowledge bit is a low level put on the bus by the receiver whereas the master generates an extra acknowledge related clock pulse.

A slave receiver which is addressed is obliged to generate an acknowledge after the reception of each byte. Also a master receiver must generate an acknowledge after the reception of each byte that has been clocked out of the slave transmitter.

The device that acknowledges has to pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is a stable Low during the High period of the acknowledge related clock pulse. Of course, setup and hold times must be taken into account. A master receiver must signal

Figure 6. Serial Bus Data Transfer Sequence

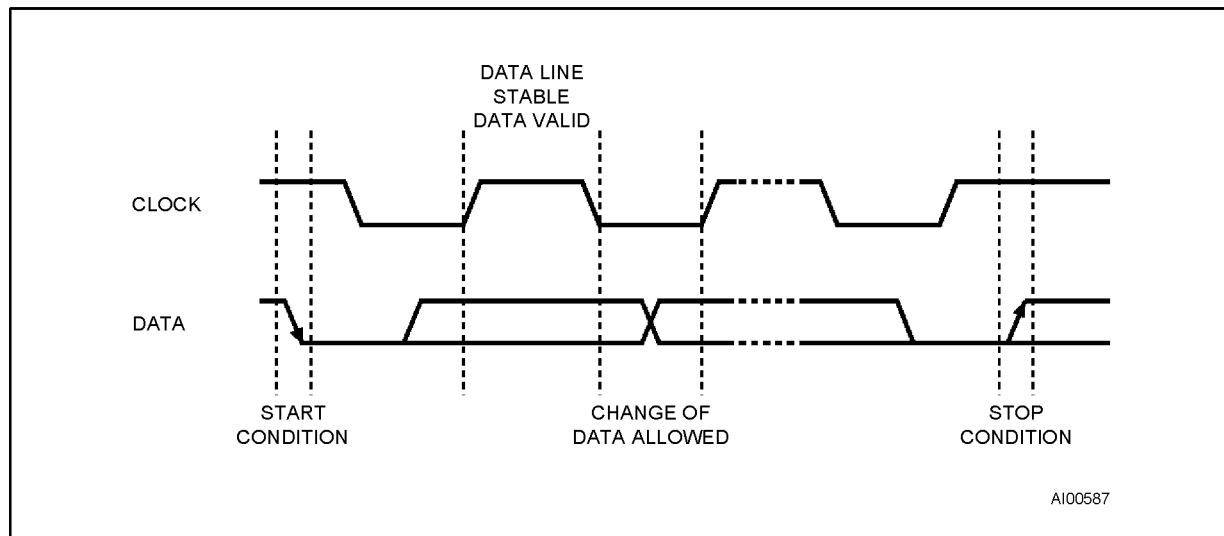


Figure 7. Acknowledgement Sequence

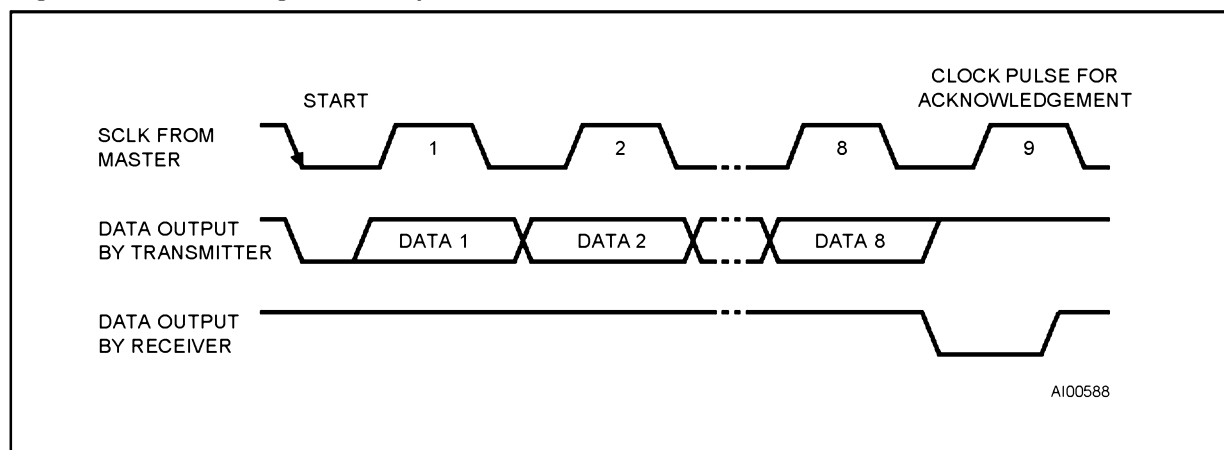
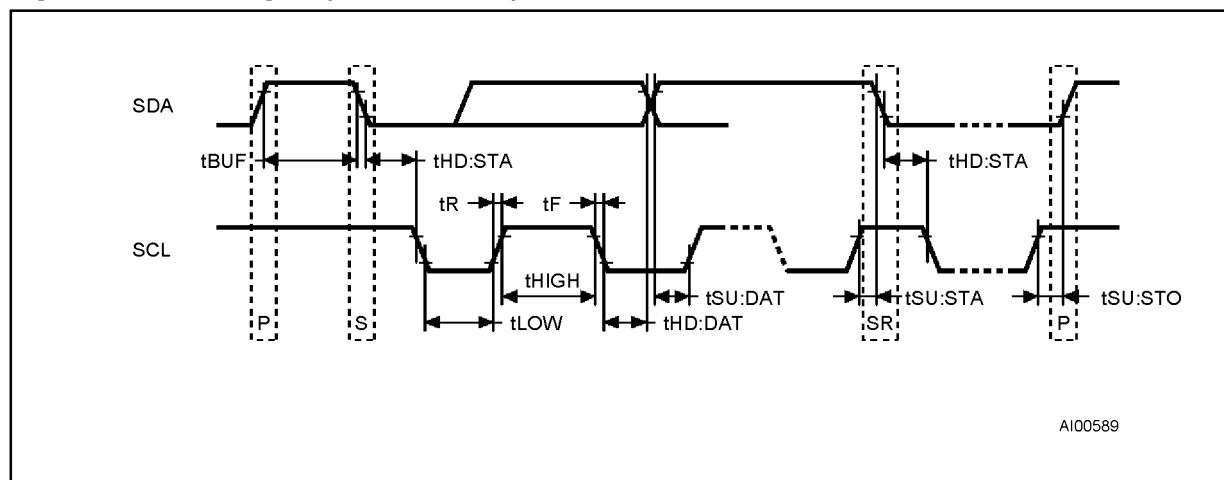


Figure 8. Bus Timing Requirements Sequence



2-WIRE BUS CHARACTERISTICS (cont'd)

an end of data to the slave transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this case the transmitter must leave the data line High to enable the master to generate the STOP condition.

WRITE MODE

In this mode the master transmitter transmits to the MK41T56 slave receiver. Bus protocol is shown in Figure 10. Following the START condition and slave address, a logic '0' ($R/\overline{W} = 0$) is placed on the bus and indicates to the addressed device that word address A_n will follow and is to be written to the on-chip address pointer. The data word to be written to the memory is strobed in next and the internal address pointer is incremented to the next memory location within the RAM on the reception of an acknowledge clock. The MK41T56 slave receiver will send an acknowledge clock to the master transmitter after it has received the slave address and again after it has received the word address and each data byte, see Figure 9.

READ MODE

In this mode the master reads the MK41T56 slave after setting the slave address, see Figure 11. Following the write mode control bit ($R/\overline{W} = 0$) and the acknowledge bit, the word address A_n is written to the on-chip address pointer. Next the START condition and slave address are repeated followed by the READ mode control bit ($R/\overline{W} = 1$). At this point the master transmitter becomes the master receiver. The data byte which was addressed will be

transmitted and the master receiver will send an acknowledge bit to the slave transmitter. The address pointer is only incremented on reception of an acknowledge bit. The MK41T56 slave transmitter will now place the data byte at address $A_n + 1$ on the bus, the master receiver reads and acknowledges the new byte and the address pointer is incremented to $A_n + 2$.

This cycle of reading consecutive addresses will continue until the master receiver sends a STOP condition to the slave transmitter.

An alternate READ mode may also be implemented whereby the master reads the MK41T56 slave without first writing to the (volatile) address pointer. The first address that is read is the last one stored in the pointer, see Figure 12.

CLOCK CALIBRATION

The MK41T56 is driven by a quartz controlled oscillator with a nominal frequency of 32,768 Hz. A typical MK41T56 is accurate within ± 1 minute per month at 25°C without calibration. The devices are tested not to exceed 35 PPM (parts per million) oscillator frequency error at 25°C, which equates to about ± 1.53 minutes per month. Of course the oscillation rate of any crystal changes with temperature.

Most clock chips compensate for crystal frequency and temperature shift error with cumbersome trim capacitors. The MK41T56 design, however, employs periodic counter correction. The calibration circuit adds or subtracts counts from the oscillator divider circuit at the divide by 128 stage, as shown in Figure 13. The number of times pulses are blanked (subtracted, negative calibration) or split (added, positive calibration) depends upon the value loaded into the five bit Calibration byte found in the Control Register. Adding counts speeds the clock up, subtracting counts slows the clock down.

The Calibration byte occupies the five lower order bits in the Control register. This byte can be set to represent any value between 0 and 31 in binary form. The sixth bit is a sign bit; '1' indicates positive calibration, '0' indicates negative calibration. Calibration occurs within a 64 minutes cycle. The first 62 minutes in the cycle may, once per minute, have one second either shortened by 128 or lengthened by 256 oscillator cycles. If a binary '1' is loaded into the register, only the first 2 minutes in the 64 minutes cycle will be modified; if a binary 6 is loaded, the first 12 will be affected, and so on.

Therefore, each calibration step has the effect of adding 512 or subtracting 256 oscillator cycles for every 125,829,120 actual oscillator cycles, that is $+ 4.068$ or -2.034 PPM of adjustment per calibration step in the calibration register. Assuming that the oscillator is in fact running at exactly 32,768 Hz,

Figure 9. Slave Address Location

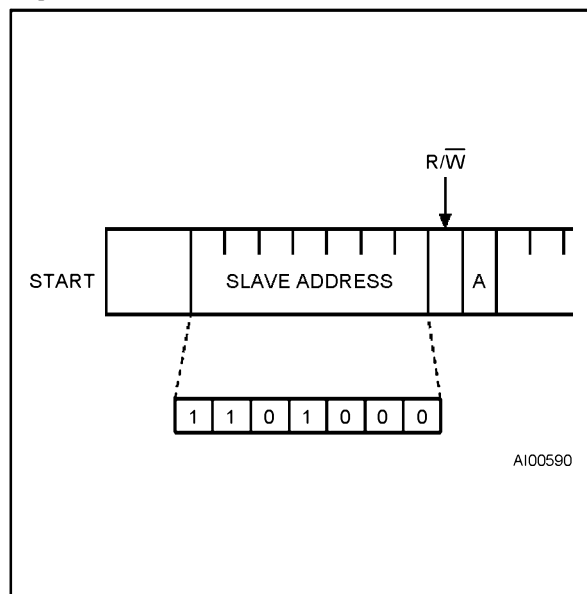


Figure 10. Write Mode Sequence

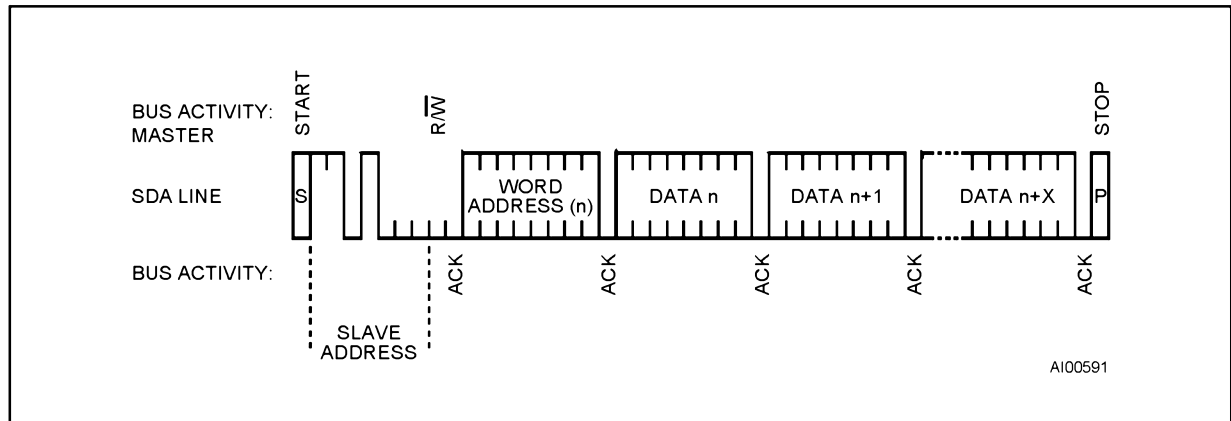


Figure 11. Read Mode Sequence

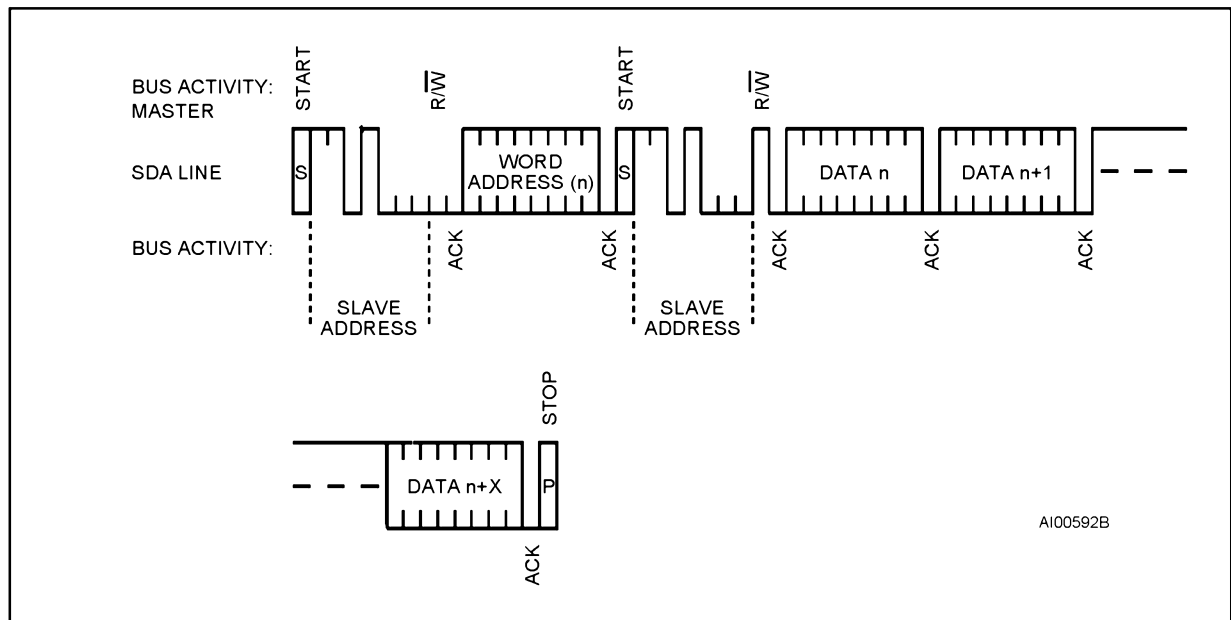
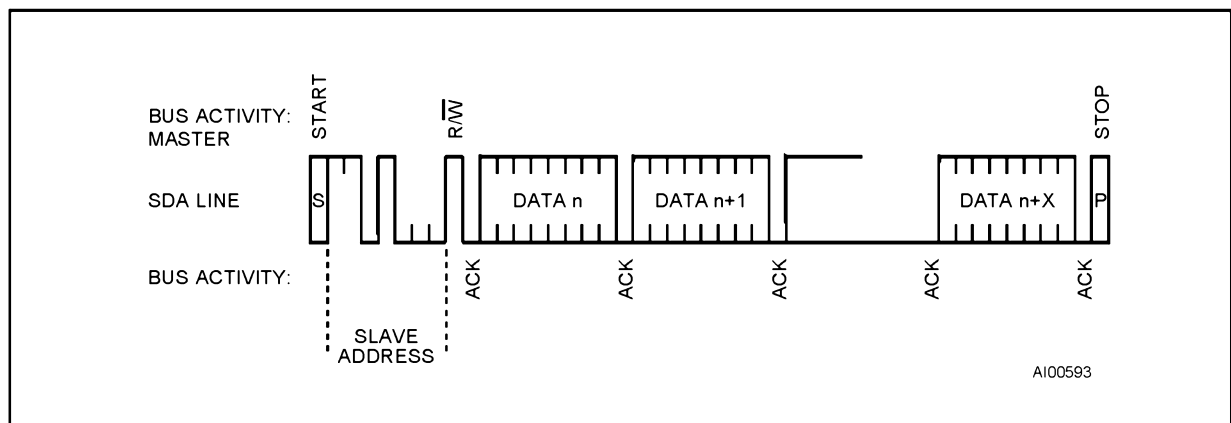


Figure 12. Alternate Read Mode Sequence



CLOCK CALIBRATION (cont'd)

each of the 31 increments in the Calibration byte would represent 10.7 seconds per month.

Two methods are available for ascertaining how much calibration a given MK41T56 may require. The first involves simply setting the clock, letting it run for a month and comparing it to a known accurate reference (like WWV broadcasts). While that may seem crude, it allows the designer to give the end user the ability to calibrate his clock as his environment may require, even after the final product is packaged in a non-user serviceable enclosure. All the designer has to do is provide a simple utility that accessed the Calibration byte. The utility could even be menu driven and made foolproof.

The second approach is better suited to a manufacturing environment, and involves the use of some test equipment. When the Frequency Test (FT) bit, the seventh-most significant bit in the Con-

trol Register, is set to a '1', and the oscillator is running at 32,768 Hz, the FT/OUT pin of the device will toggle at 512 Hz. Any deviation from 512 Hz indicates the degree and direction of oscillator frequency shift at the test temperature.

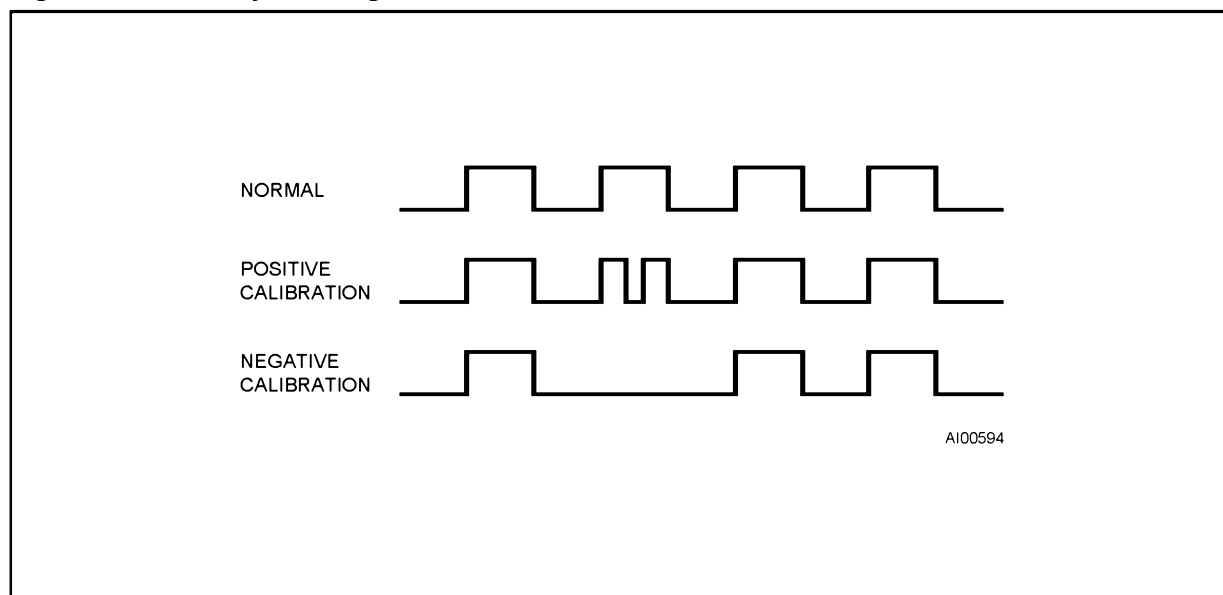
For example, a reading of 512.01024 Hz would indicate a +20 PPM oscillator frequency error, requiring a -10(001010) to be loaded into the Calibration Byte for correction. Note that setting or changing the Calibration Byte does not affect the Frequency test output frequency.

OUTPUT DRIVER PIN

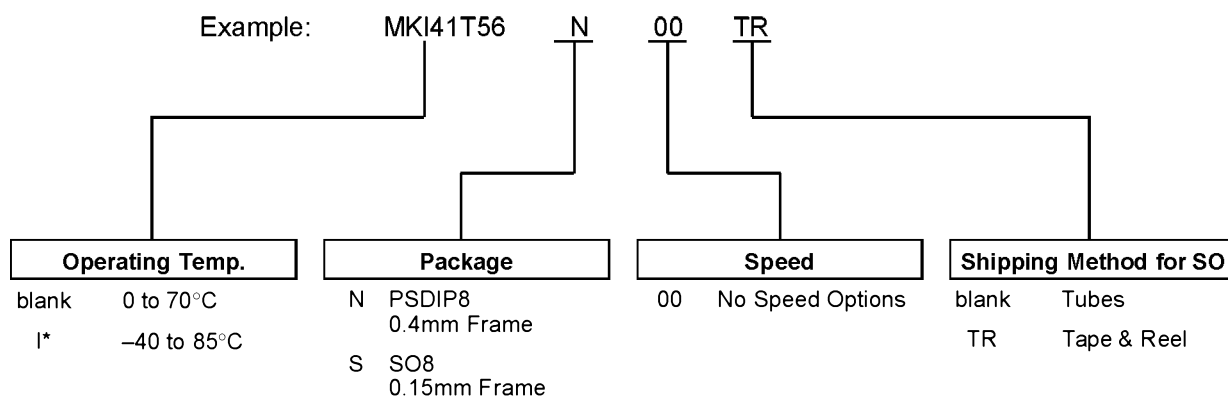
When the FT bit is not set the FT/OUT pin becomes an output driver that reflects the contents of D7 of the control register. In other words when D6 of location 7 is a zero and D7 of location 7 is a zero and then the FT/OUT pin will be driven low.

Note: The FT/OUT pin is open drain which requires an external pull-up resistor.

Figure 13. Divide by 128 Stage



ORDERING INFORMATION SCHEME



Note: I* Available in the SO package only.

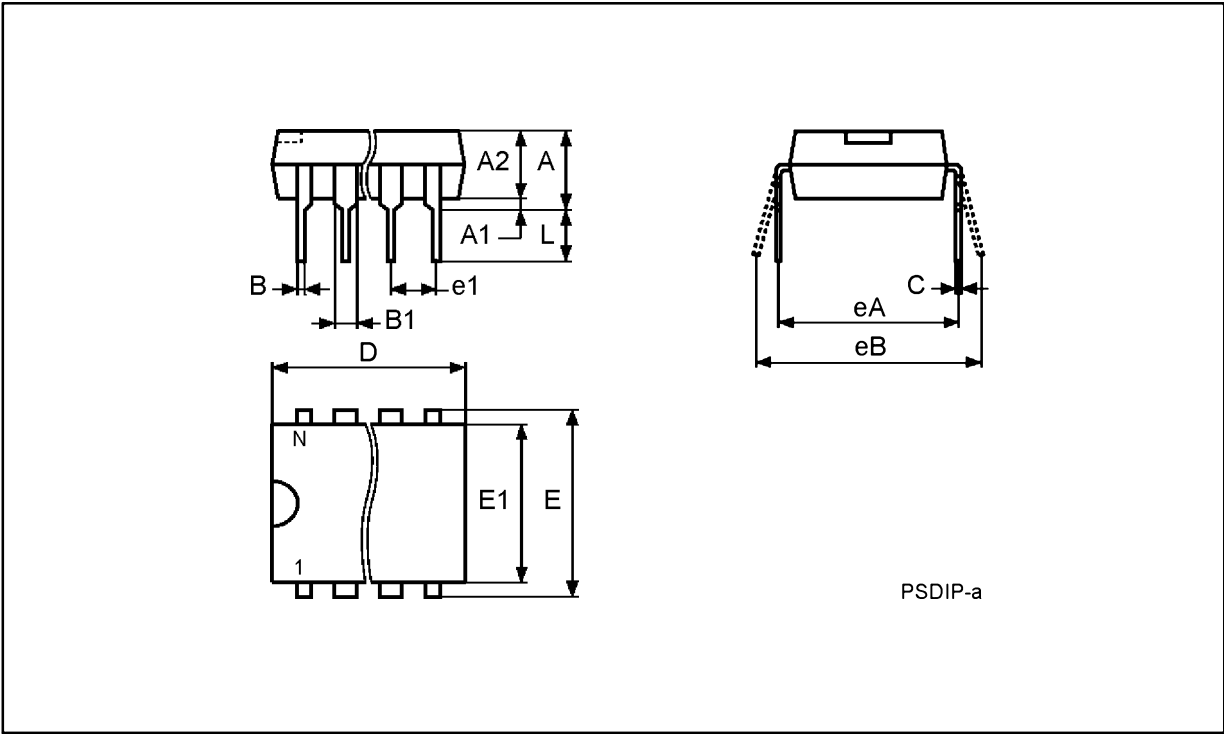
For a list of available options refer to the current Memory Shortform catalogue.

For further information or any aspect of this device, please contact the SGS-THOMSON Sales Office nearest to you.

PSDIP8 - 8 pin Plastic Skinny DIP, 0.4mm lead frame

Symb	mm			inches		
	Typ	Min	Max	Typ	Min	Max
A			4.80			0.189
A1		0.70	–		0.028	–
A2		3.10	3.60		0.122	0.142
B		0.38	0.58		0.015	0.023
B1		1.15	1.65		0.045	0.065
C		0.38	0.52		0.015	0.020
D		9.20	9.90		0.362	0.390
E	7.62	–	–	0.300	–	–
E1		6.30	7.10		0.248	0.280
e1	2.54	–	–	0.100	–	–
eA		8.40	–		0.331	–
eB			9.20			0.362
L		3.00	3.80		0.118	0.150
N	8			8		

PSDIP8



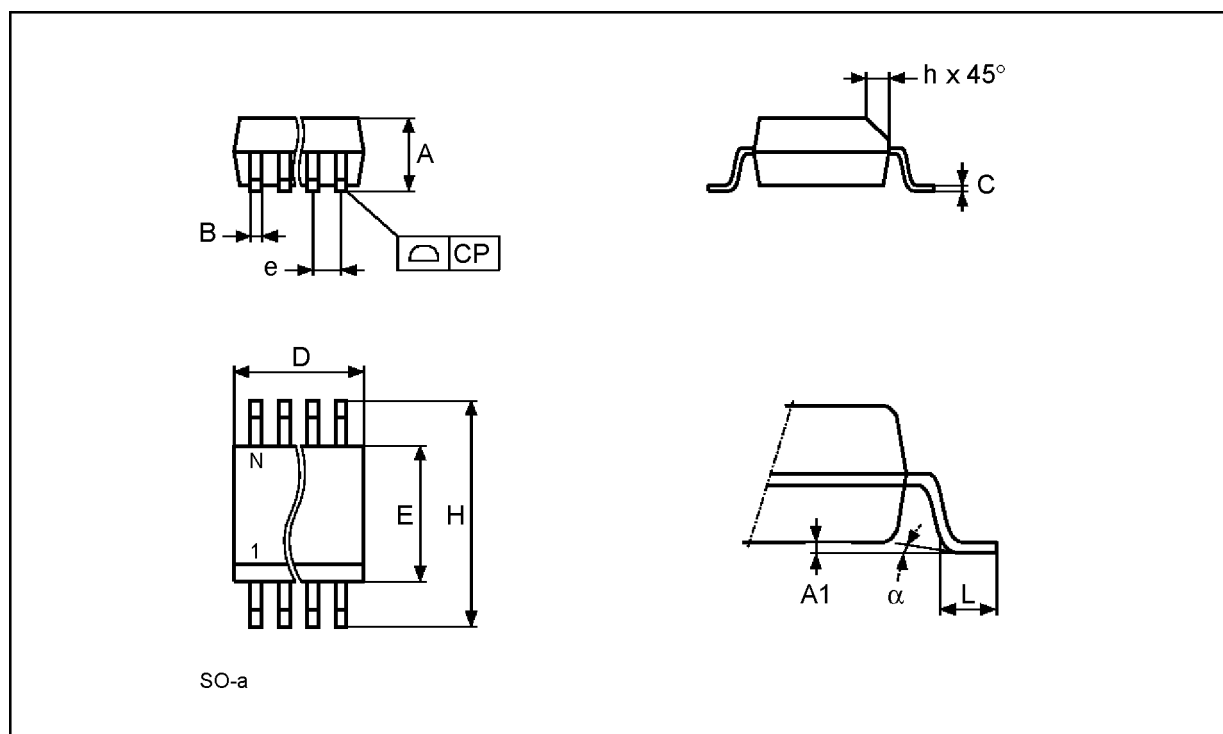
PSDIP-a

Drawing is not to scale

SO8 - 8 lead Plastic Small Outline, 150 mils body width

Symb	mm			inches		
	Typ	Min	Max	Typ	Min	Max
A		1.35	1.75		0.053	0.069
A1		0.10	0.25		0.004	0.010
B		0.33	0.51		0.013	0.020
C		0.19	0.25		0.007	0.010
D		4.80	5.00		0.189	0.197
E		3.80	4.00		0.150	0.157
e	1.27	–	–	0.050	–	–
H		5.80	6.20		0.228	0.244
h		0.25	0.50		0.010	0.020
L		0.40	0.90		0.016	0.035
α		0°	8°		0°	8°
N	8			8		
CP			0.10			0.004

SO8



Drawing is not to scale

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