

TOSHIBA Bi-CMOS INTEGRATED CIRCUIT SILICON MONOLITHIC

# TB6526AF

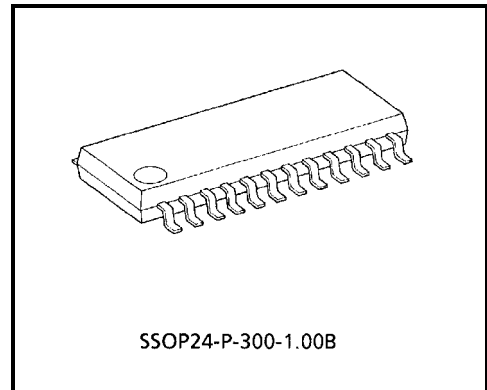
## CHOPPER-TYPE BIPOLAR STEPPING MOTOR CONTROL DRIVER IC

The TB6526AF is a PWM chopper-type sinusoidal micro-step bipolar stepping motor driver IC.

It is capable of 1-2 and 2W1-2 phase excitation modes and forward and reverse rotation modes, low-vibration, low-torque ripple, and high-efficiency driving.

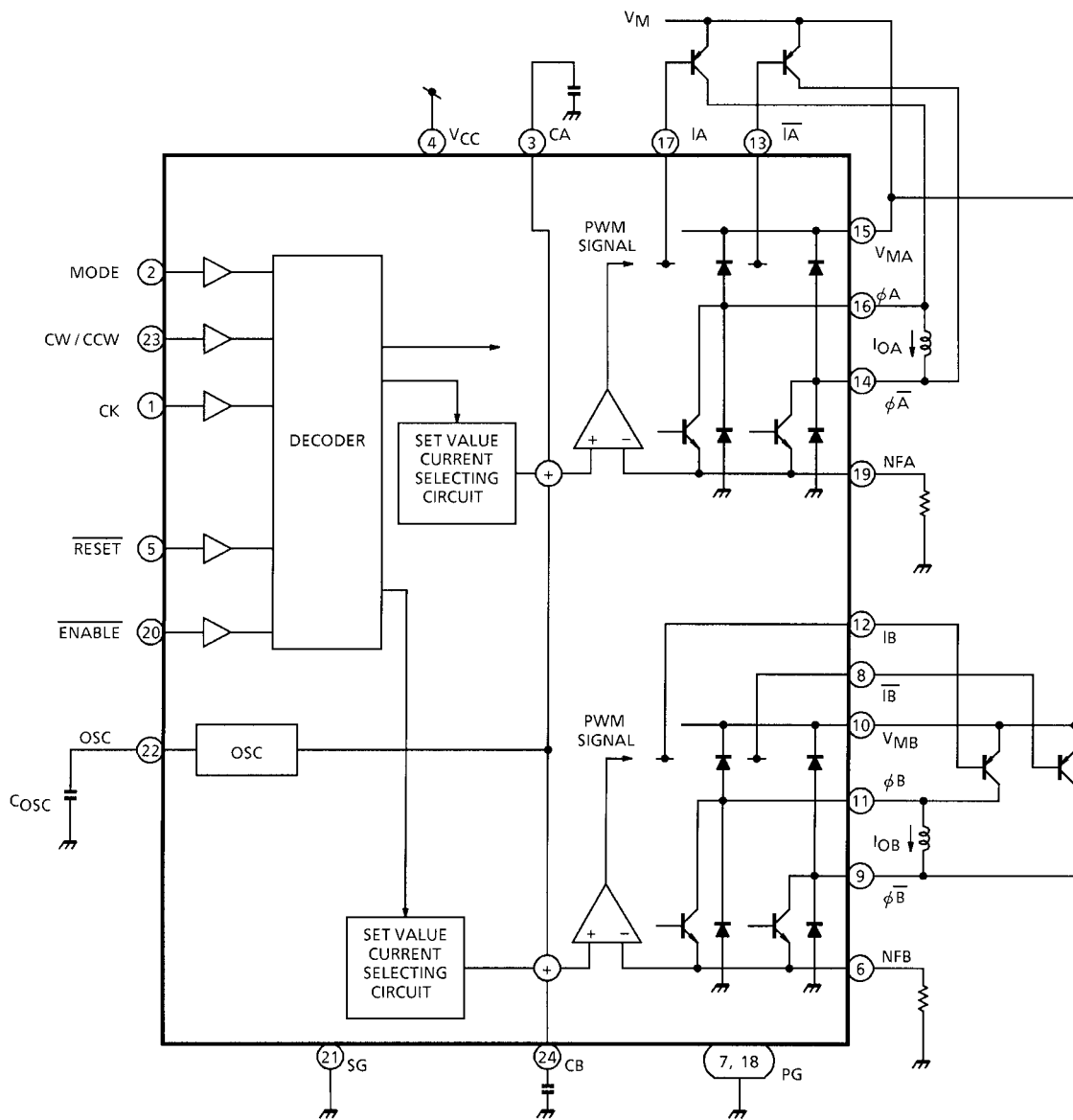
### FEATURES

- Forward and reverse rotations are available.
- 1-2, 2W1-2 phase driving is available.
- Structured by Bi-CMOS process.
- Package: SSOP24-P-300-1.00B
- Externally equipped with PNP output transistor.
- Reset and enable pins are attached.



Weight: 0.27 g (Typ.)

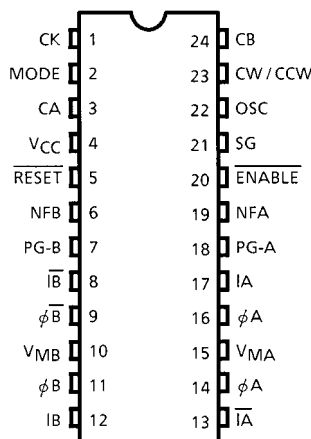
## BLOCK DIAGRAM





## PIN FUNCTION

PIN No.	SYMBOL	FUNCTIONAL DESCRIPTION	
1	CK	CLOCK Signal Input	Truth table A
2	MODE	Excitation Mode Setting terminal	Truth table B
3	CA	Noise reduction condenser outer terminal	
4	V <sub>CC</sub>	Power voltage supply terminal for Logic	
5	$\overline{\text{RESET}}$	$\overline{\text{RESET}}$ Signal Input terminal	Truth table A
6	NFB	B Channel current detective terminal	
7	PG-B	Power GND B terminal	
8	$\overline{\text{IB}}$	Upper PNP Transistor Base terminal ( $\overline{\text{B}}$ phase)	
9	$\phi\overline{\text{B}}$	$\overline{\text{B}}$ output	
10	V <sub>MB</sub>	Power voltage supply terminal for Motor B	
11	$\phi\text{B}$	Output B terminal	
12	IB	Upper PNP Transistor Base terminal (B phase)	
13	$\overline{\text{IA}}$	Upper PNP Transistor Base terminal ( $\overline{\text{A}}$ phase)	
14	$\phi\overline{\text{A}}$	Output $\overline{\text{A}}$ terminal	
15	V <sub>MA</sub>	Power voltage supply terminal for Motor A	
16	$\phi\text{A}$	Output A terminal	
17	IA	Upper side PNP transistor Base terminal (A phase)	
18	PG-A	Power GND A terminal	
19	NFA	A Channel current detection terminal	
20	$\overline{\text{ENABLE}}$	$\overline{\text{ENABLE}}$ Signal input terminal	Truth table A
21	SG	Signal GND terminal	
22	OSC	Internal Oscillation frequency detective terminal with external condenser	
23	CW / CCW	Forward rotation / Reverse rotation signal input	Truth table A
24	CB	Noise reduction condenser outside terminal	

## PIN CONNECTION



## TRUTH TABLE A

INPUT				MODE
CK1	CW / CCW	RESET	ENABLE	
	L	H	L	CW
	H	H	L	CCW
X	X	L	L	INITIAL MODE
X	X	X	H	Z

Z : High impedance  
X : Don't Care

Note: Do not use INHIBIT MODE.

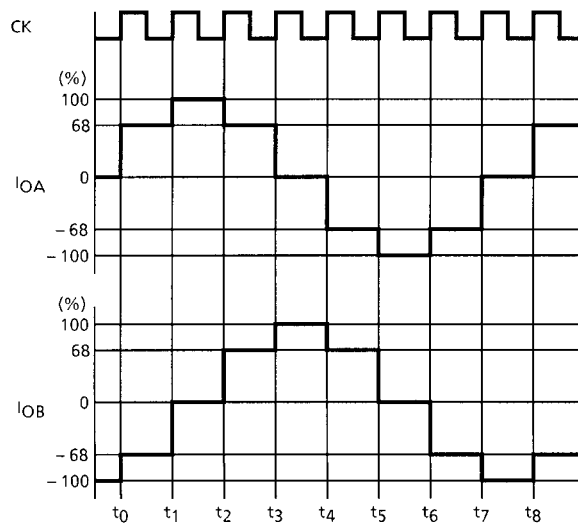
## TRUTH TABLE B

INPUT MODE	MODE (EXCITATION)
L	1-2 phase
H	2W1-2 phase

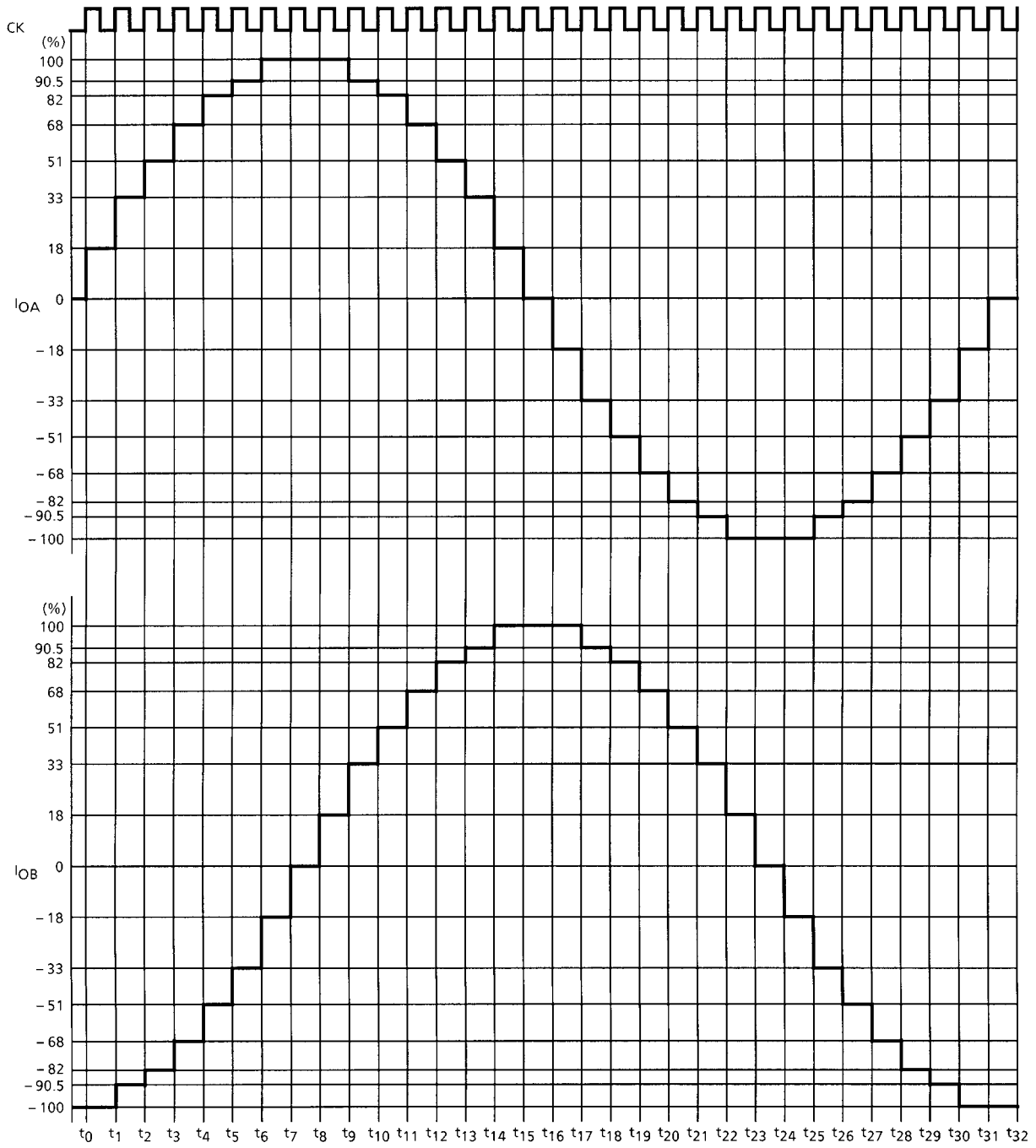
## INITIAL MODE

MODE EXCITATION	A-PHASE CURRENT	B-PHASE CURRENT
1-2 phase	100%	0%
2W1-2 phase	100%	0%

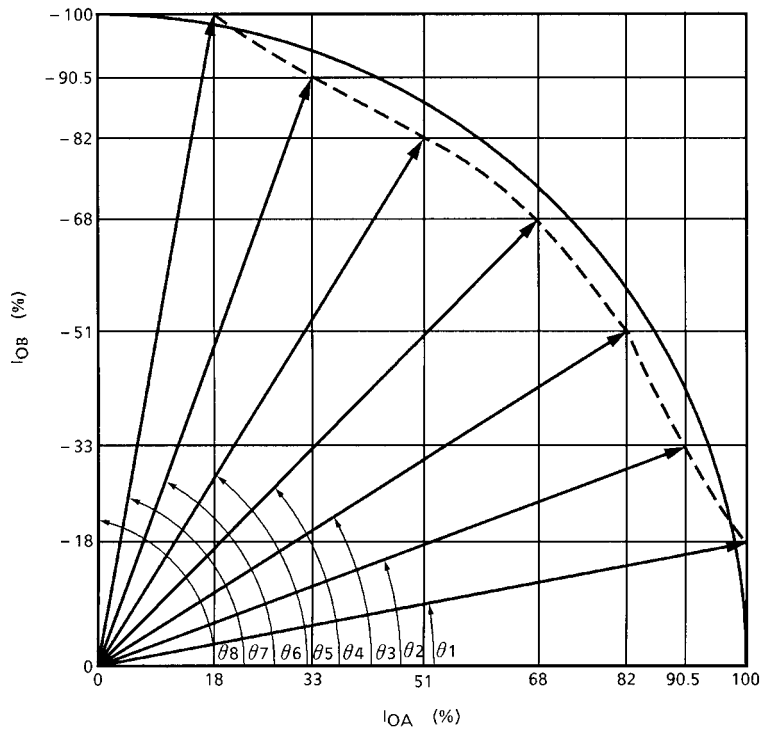
## 1-2 PHASE EXCITATION (MODE : L, CW mode)



2W1-2 EXCITATION (MODE : H, CW mode)



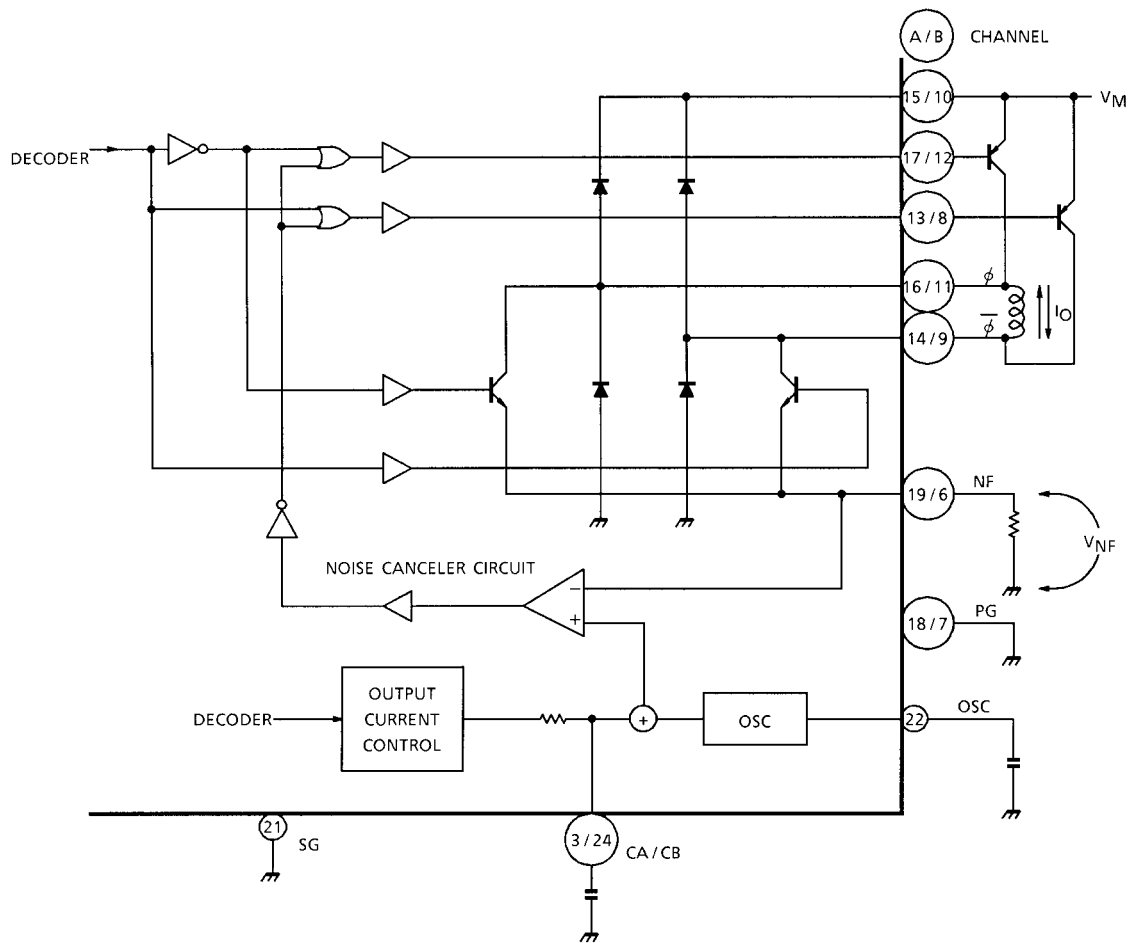
**OUTPUT CURRENT VECTOR OR BIT (Normalize to 90 deg for each one step)**



$\theta$	ROTATION ANGLE		VECTOR LENGTH	
	IDEAL	TB6526AF	IDEAL	TB6526AF
$\theta_0$	0°	0°	100	100.00
$\theta_1$	11.25°	10.20°	100	101.65
$\theta_2$	22.5°	20.03°	100	96.35
$\theta_3$	33.75°	31.88°	100	96.56
$\theta_4$	45°	45°	100	96.17
$\theta_5$	56.25°	58.12°	100	96.57
$\theta_6$	67.5°	69.97°	100	96.33
$\theta_7$	78.75°	79.80°	100	101.61
$\theta_8$	90°	90°	100	100.00

1-2 / 2W1-2, Phase

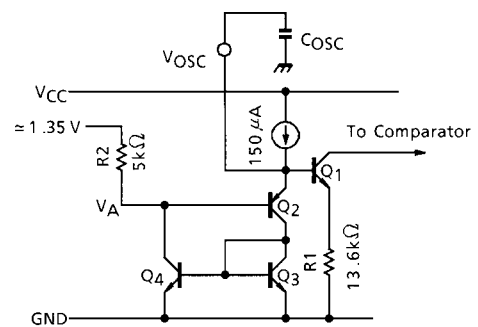
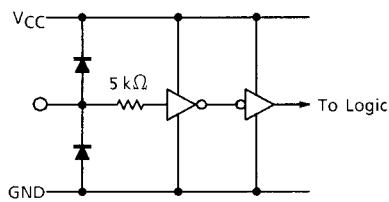
**OUTPUT CIRCUIT**



**INPUT CIRCUIT**

CK, CW / CCW,  $\overline{\text{RESET}}$ ,  
 $\overline{\text{ENABLE}}$ , MODE Terminals

OSC : Terminals



- OSC frequency calculation

$V_{OSC}$  is increased by  $C_{OSC}$  charging through the constant current source (150  $\mu$ A).

$V_{OSC}$  is calculated by following equation.

$$V_{OSC} = \frac{150 \times 10^{-6} \times t}{C_{OSC}}$$

Q2 is turned "off" when  $V_{OSC}$  is less than the voltage of 1.35 V +  $V_{BE}$  (Q2) approximately equal to 2.05 V.

Q3 and Q4 are turned "on" when  $V_{OSC}$  becomes 2.05 V.

$$V_{OSC} (H) = V_{BE} (Q2) + 1.35$$

$$\approx 2.05 \text{ V}$$

Lower level of V (22) pin is equal to  $V_{BE}$  (Q2) +  $V_{CE}$  (SAT) (Q4) approximately equal to 1.0 V.

$$V_{OSC} (L) = V_{BE} (Q2) + V_{CE} (SAT) (Q4)$$

$$\approx 1.0 \text{ V}$$

Assuming that  $V_{OSC} = 1.0 \text{ V}$  ( $t = t_1$ ) and  $= 2.05 \text{ V}$  ( $t = t_2$ ), OSC frequency is calculated as follows.

$$t_1 = \frac{1.0 \cdot C_{OSC}}{150 \times 10^{-6}}$$

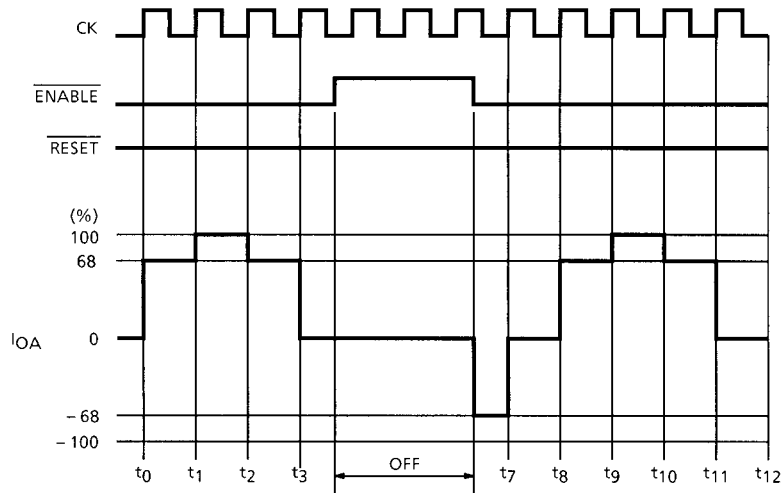
$$t_2 = \frac{2.05 \cdot C_{OSC}}{150 \times 10^{-6}}$$

$$f_{OSC} = \frac{1}{t_2 - t_1} = \frac{150 \times 10^{-6}}{C_{OSC} (2.05 - 1.0)}$$

$$\approx \frac{0.143}{C_{OSC}} \text{ (kHz) (} C_{OSC} \text{ unit} = \mu\text{F)}$$



**ENABLE AND RESET FUNCTION AND  $\overline{MO}$  SIGNAL**

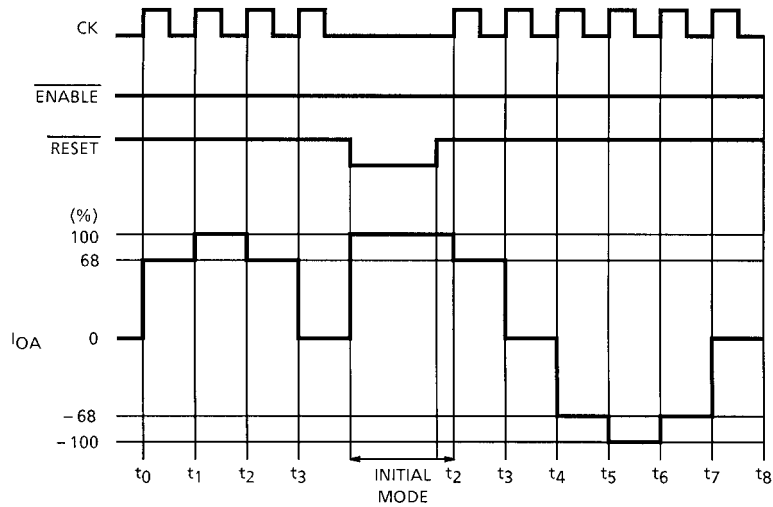


**Fig.1. 1-2 phase drive mode (MODE : L)**

$\overline{ENABLE}$  signal disables only Output signal. Internal logic functions are proceeded by CK signal without regard to  $\overline{ENABLE}$  signal.

Therefore, Output Current is initiated from the proceeded timing point of internal logic circuit, after release of disable mode.

Fig.1 shows the  $\overline{ENABLE}$  functions, when the system is selected in 1-2 phase drive mode.



**Fig.2. 1-2 phase drive mode (MODE : L)**

As  $\overline{RESET}$  is low, the decoder is initialized. (Output Current : A-Phase 100%, B-Phase 0%)

After  $\overline{RESET}$  is high, the motion is resumed from next clock as show in Fig.2.

$\overline{MO}$  (Monitor Output) signals is used as rotation and initial signal for stable rotation checking.

## MAXIMUM RATING (Ta = 25°C)

CHARACTERISTIC	SYMBOL	RATING	UNIT
Supply Voltage	V <sub>CC</sub>	5.5	V
Output Voltage	V <sub>M (opr.)</sub>	3.5~8.0	V
	V <sub>M (MAX.)</sub>	10.0	
Output Current	I <sub>O (MAX.)</sub>	120	mA
Input Voltage	V <sub>IN</sub>	~V <sub>CC</sub> + 0.5	V
Power Dissipation	P <sub>D</sub>	0.83 (Note 1)	W
		1.04 (Note 2)	
Operating Temperature	T <sub>opr</sub>	-30~85	°C
Storage Temperature	T <sub>stg</sub>	-55~150	°C
Feed Back Voltage	V <sub>I</sub>	1.0	V

Note 1: No heat sink

Note 2: When mounted on substrate (50 × 50 × 1.6 mm Cu 10%)

## RECOMMENDED OPERATING CONDITIONS (Ta = -30~85°C)

CHARACTERISTIC	SYMBOL	TEST CONDITION	MIN	TYP.	MAX	UNIT
Control Power Supply Voltage	V <sub>CC (opr.)</sub>		2.7	3.0	5.5	V
Motor Power Supply Voltage	V <sub>M (opr.)</sub>		3.5	—	8.0	V
Output Current	I <sub>OUT</sub>		—	—	100	mA
Input Voltage	V <sub>IN</sub>		-0.4	—	V <sub>CC</sub> + 0.4	V
Clock Frequency	f <sub>CLOCK</sub>		—	—	5	kHz
OSC Frequency	f <sub>OSC</sub>		15	—	80	kHz

## ELECTRICAL CHARACTERISTICS

Unless otherwise specified ( $T_a = 25^\circ\text{C}$ ,  $V_{CC} = 3\text{ V}$ ,  $V_M = 5\text{ V}$ , load inductance :  
 $L = 8\text{ mH}$  /  $R = 50\ \Omega$ , with outer PNP)

CHARACTERISTIC		SYMBOL	TEST CIR-CUIT	TEST CONDITION	MIN	TYP.	MAX	UNIT
Input Voltage	High	$V_{IN(H)}$	1	MODE, CW / CCW, $\overline{\text{ENABLE}}$ CK, RESET	$V_{CC} \times 0.7$	—	$V_{CC} + 0.4$	V
	Low	$V_{IN(L)}$			GND - 0.4	—	$V_{CC} \times 0.3$	
Input Current		$I_{IN(H)}$	2	$V_{IN} = 3.0\text{ V}$	—	—	100	nA
		$I_{IN(L)}$		$V_{IN} = 0\text{ V}$	—	—	100	
Current Consumption $V_{CC}$ Pin		$I_{CC1}$	3	Output open, $\overline{\text{RESET}} : \text{H}$ , $\overline{\text{ENABLE}} : \text{L}$ , (1-2 phase excitation)	—	7	9	mA
		$I_{CC2}$		Output open, $\overline{\text{RESET}} : \text{H}$ , $\overline{\text{ENABLE}} : \text{L}$ , (2W1-2 phase excitation)	—	7	9	
		$I_{CC3}$		$\overline{\text{RESET}} : \text{L}$ , $\overline{\text{ENABLE}} : \text{H}$	—	1.3	—	
		$I_{CC4}$		$\overline{\text{RESET}} : \text{H}$ , $\overline{\text{ENABLE}} : \text{H}$	—	1.3	—	
Comparator Reference Voltage Level		$V_{NF1}$	9	$C_A, C_B$	0.245	0.275	0.305	V
		$V_{NF2}$	4	$R_{NF} = 3.3\ \Omega$ , $C_{OSC} = 3300\text{ pF}$	175	195	220	mV
		$V_{NF3}$	4	$R_{NF} = 2.2\ \Omega$ , $C_{OSC} = 3300\text{ pF}$	150	172	190	mV
Output Inter-channel Differential		$\Delta V_O$	4	$(V_{NFA} - V_{NFB}) / V_{NFA}$ , $C_{OSC} = 3300\text{ pF}$ , $R_{NF} = 3.3\ \Omega$	-10	—	10	%
Maximum OSC Frequency		$f_{OSC(\text{MAX.})}$	—		100	—	—	kHz
Minimum OSC Frequency		$f_{OSC(\text{MIN.})}$	—		—	—	10	kHz
OSC Frequency		$f_{OSC}$	5	$C_{OSC} = 3300\text{ pF}$	31	44	70	kHz

## ELECTRICAL CHARACTERISTICS

Unless otherwise specified ( $T_a = 25^\circ\text{C}$ ,  $V_{CC} = 3\text{ V}$ ,  $V_M = 5\text{ V}$ , load inductance :  $L = 8\text{ mH} / R = 50\ \Omega$ , with outer PNP)

### OUTPUT SECTION

CHARACTERISTIC		SYMBOL	TEST CIRCUIT	TEST CONDITION	MIN	TYP.	MAX	UNIT	
Upper Side Driving Current		$I_U$	6	$V_C = 3\text{ V}$	—	1.5	1.6	mA	
Lower Side Saturation Voltage		$V_{SAT L1}$	7	$I_{OUT} = 0.06\text{ A}$	—	0.10	—	V	
		$V_{SAT L2}$		$I_{OUT} = 0.12\text{ A}$	—	0.16	0.43		
Diode Forward Voltage	Upper Side	$V_{FU}$	8	$I_{OUT} = 0.12\text{ A}$	—	1.24	1.8	V	
	Lower Side	$V_{FL}$			—	0.95	1.6		
Output Dark Current (A + B channel)		$I_{M1}$	3	ENABLE : "H" level RESET : "L" level Output open	—	—	50	$\mu\text{A}$	
		$I_{M2}$		ENABLE : "L" level RESET : "H" level Output open	—	17	28		
NF Dark Current (1 channel)		$I_{NF}$		ENABLE : "L" level RESET : "H" level Output open	1	2.5	7	mA	
A-B Chop- per Current (Note)	2W1-2 phase excitation	1-2 phase excitation	4	Vector	$\theta = 0$	$R_{NF} = 3.3\ \Omega$ $C_{OSC} = 3300\text{ pF}$ $V_{NF}$	—	100	—
	2W1-2 phase excitation	—			$\theta = 1/8$		—	100	—
	2W1-2 phase excitation	—			$\theta = 2/8$		85.5	90.5	95.5
	2W1-2 phase excitation	—			$\theta = 3/8$		77	82	87
	2W1-2 phase excitation	1-2 phase excitation			$\theta = 4/8$		64	69	74
	2W1-2 phase excitation	—			$\theta = 5/8$		48	53	58
	2W1-2 phase excitation	—			$\theta = 6/8$		31	36	41
	2W1-2 phase excitation	—			$\theta = 7/8$		16	21	26

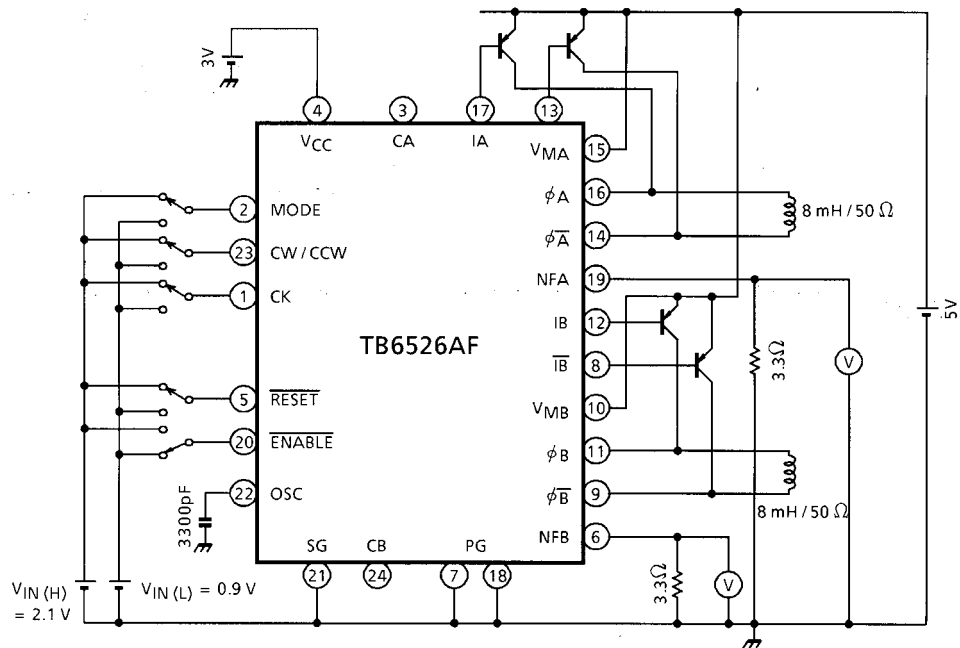
Note: Maximum current  $\theta = 0$  is set at 100.

## ELECTRICAL CHARACTERISTICS

Unless otherwise specified ( $T_a = 25^\circ\text{C}$ ,  $V_{CC} = 3\text{ V}$ ,  $V_M = 5\text{ V}$ , load inductance :  
 $L = 8\text{ mH} / R = 50\ \Omega$ , with outer PNP)

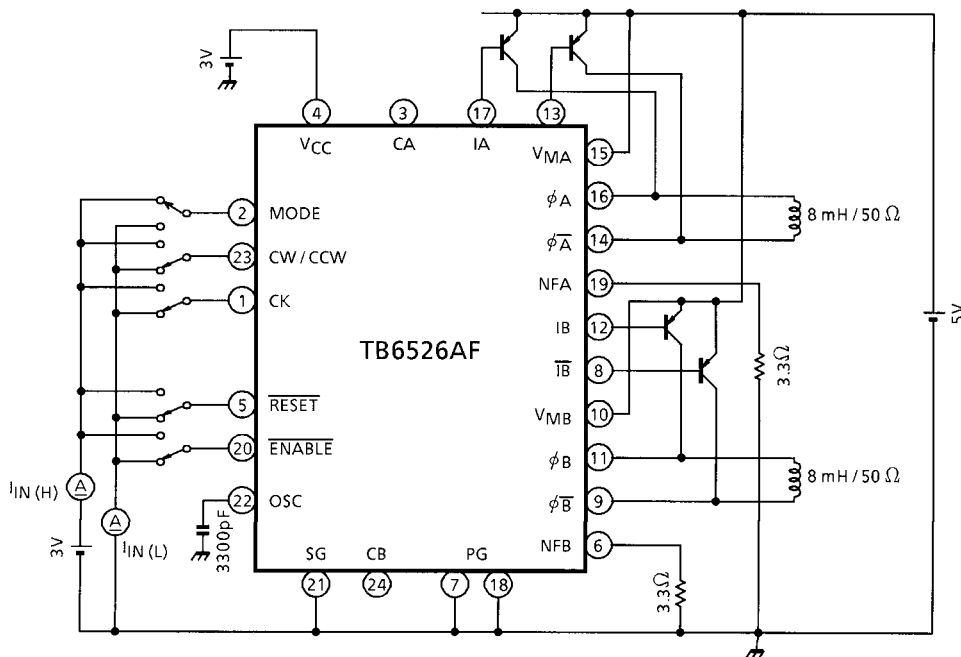
CHARACTERISTIC	SYMBOL	TEST CIR-CUIT	TEST CONDITION	MIN	TYP.	MAX	UNIT	
Reference Voltage	$\Delta V_{NF}$	9	$\Delta\theta = 0/8-1/8$	Measured by CA and CB	—	0	—	mV
			$\Delta\theta = 1/8-2/8$		10	17	35	
			$\Delta\theta = 2/8-3/8$		5	16	30	
			$\Delta\theta = 3/8-4/8$		16.25	21	41.25	
			$\Delta\theta = 4/8-5/8$		25	32	50	
			$\Delta\theta = 5/8-6/8$		26.25	31	51.25	
			$\Delta\theta = 6/8-7/8$		15	28	45	
Output Tr Switching	$t_r$	12	$R_L = 2\ \Omega$ , $V_{NF} = 0\text{ V}$ , $C_L = 15\text{ pF}$	—	0.3	—	$\mu\text{s}$	
	$t_f$			—	2.2	—		
	$t_{pLH}$		CK~output	—	1.5	—		
	$t_{pHL}$			—	2.7	—		
	$t_{pLH}$		OSC~output	—	5.4	—		
	$t_{pHL}$			—	6.3	—		
	$t_{pLH}$		$\overline{\text{RESET}} \sim \text{output}$	—	2.0	—		
	$t_{pHL}$			—	2.5	—		
	$t_{pLH}$		$\overline{\text{ENABLE}} \sim \text{output}$	—	5.0	—		
	$t_{pHL}$			—	6.0	—		
Output Leakage Current	$I_{OL}$	10	$V_M = 10\text{ V}$	—	—	50	$\mu\text{A}$	
$V_{MA} / V_{MB}$ Off Current	$I_{off}$	11	$V_{CC} = 0$ , $V_M = 5\text{ V}$	—	—	1	$\mu\text{A}$	

**TEST CIRCUIT 1 :  $V_{IN(H)}$ ,  $V_{IN(L)}$**

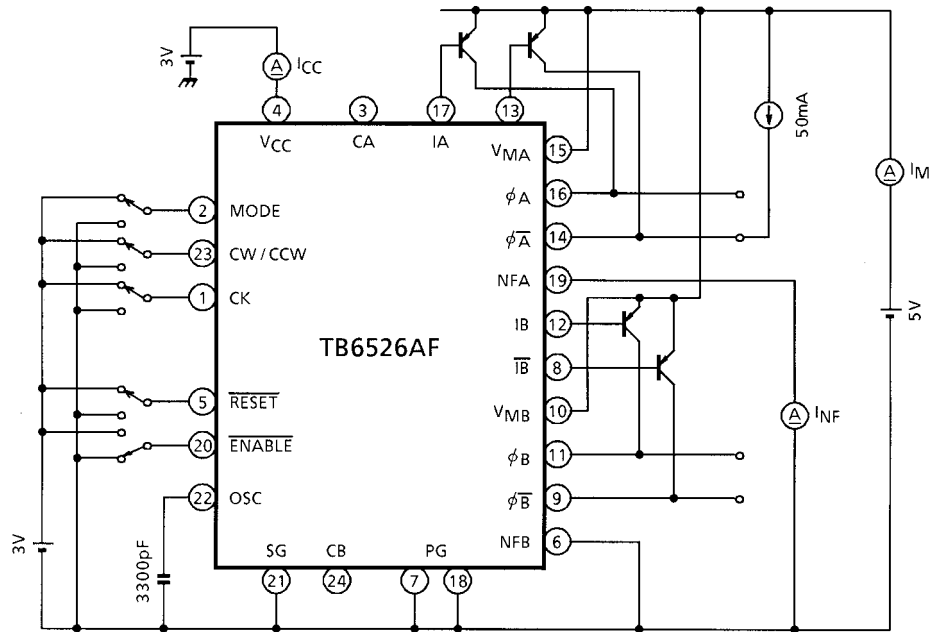


Note: When input voltage  $V_{IN(H)}$ ,  $V_{IN(L)}$  is applied, verify the output function (NF voltage measurement).

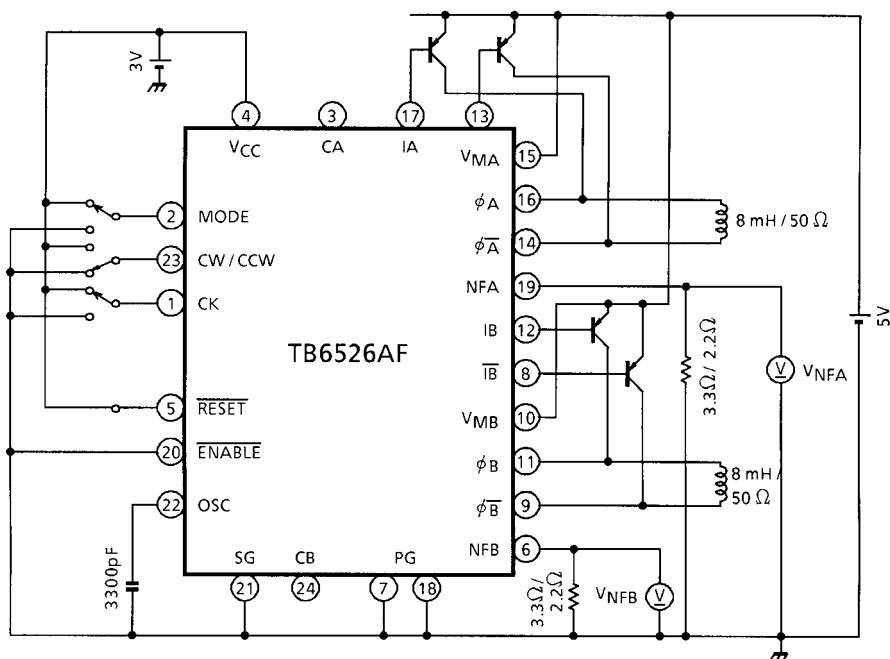
**TEST CIRCUIT 2 :  $I_{IN(H)}$ ,  $I_{IN(L)}$**



## TEST CIRCUIT 3 : $I_{CC}$ , $I_M$ , $I_{NF}$

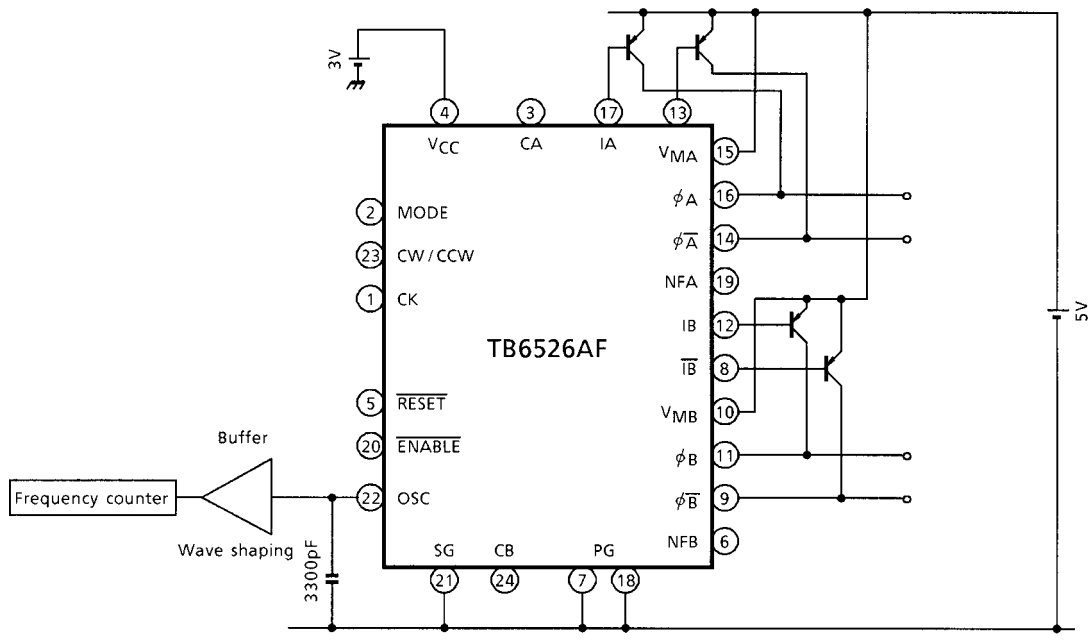


## TEST CIRCUIT 4 : $V_{NF2}$ , $V_{NF3}$ , $\Delta V_O$

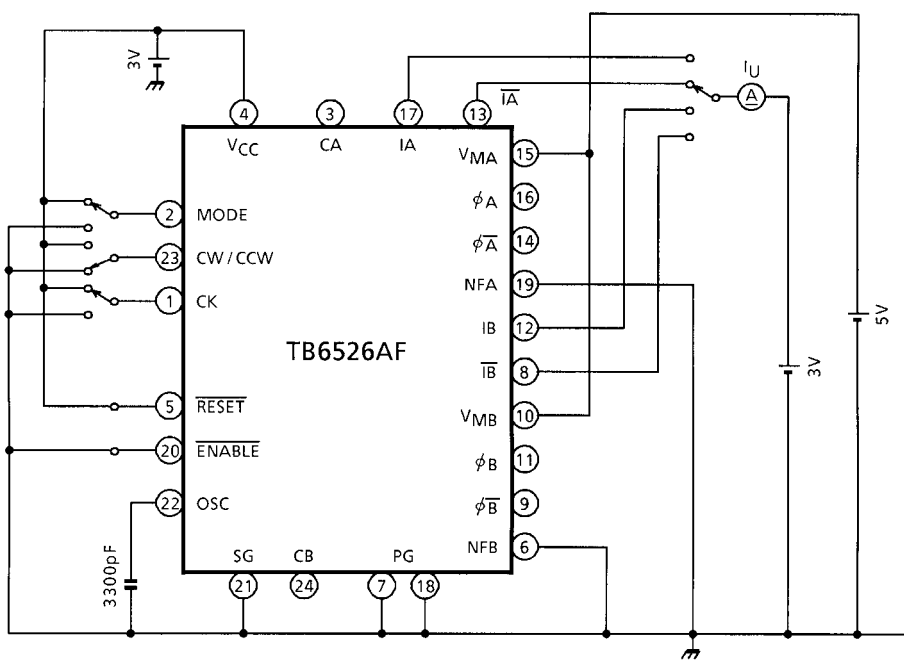


Note:  $V_{NF2}$ :  $V_{NFA}$  (100%),  $V_{NFB}$  (100%) when  $R_{NF} = 3.3 \Omega$   
 $V_{NF3}$ :  $V_{NFA}$  (100%),  $V_{NFB}$  (100%) when  $R_{NF} = 2.2 \Omega$

**TEST CIRCUIT 5 : f<sub>osc</sub>**

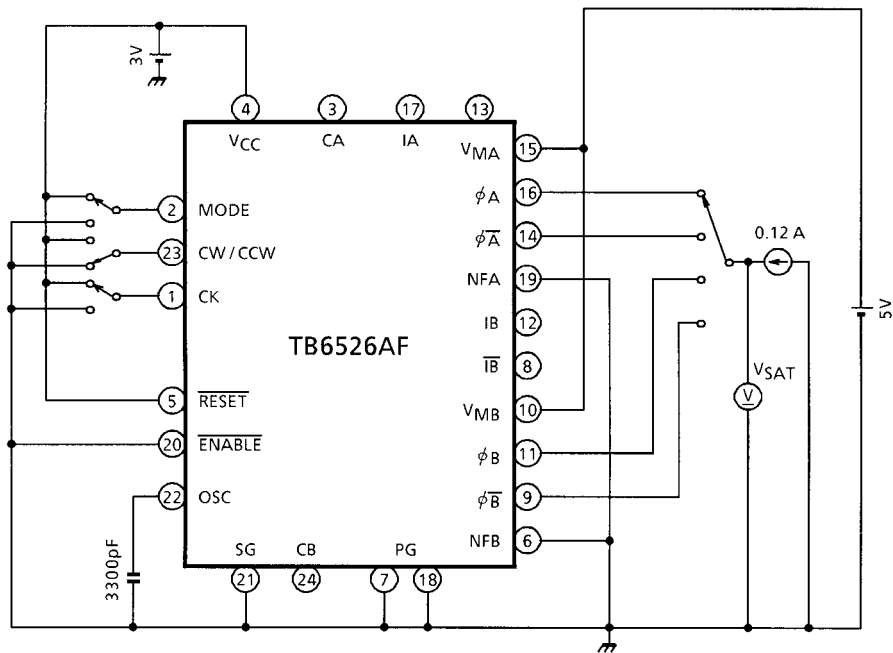


**TEST CIRCUIT 6 : I<sub>U</sub>**

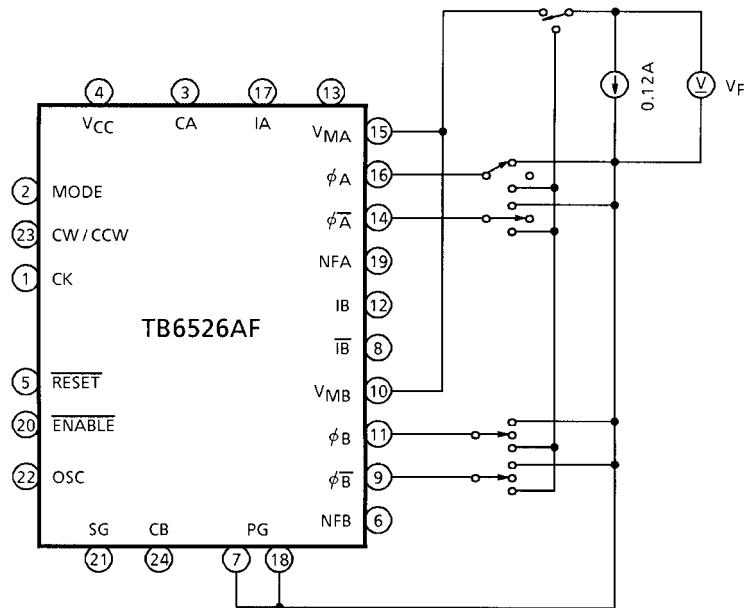




## TEST CIRCUIT 7 : $V_{SAT}$

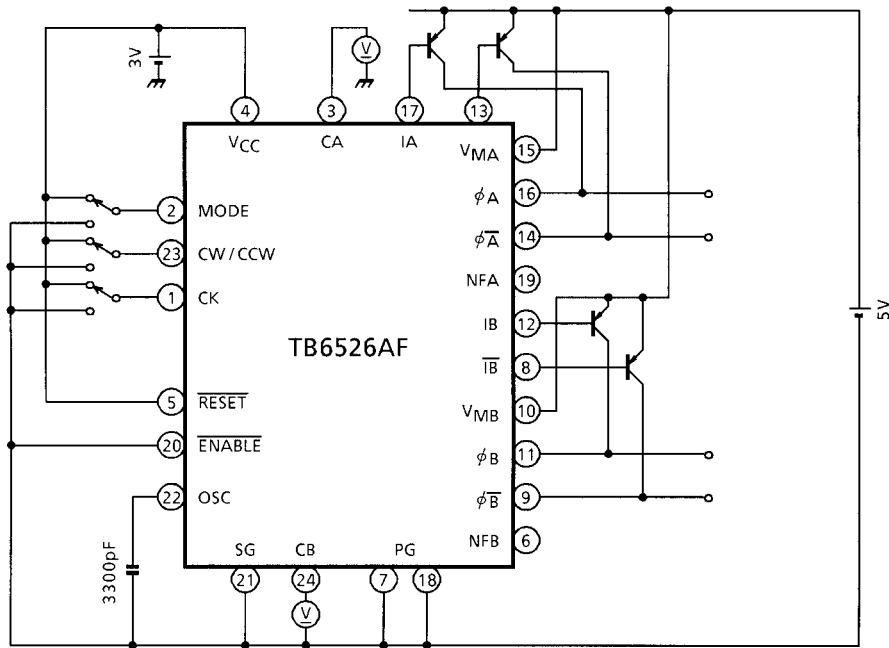


## TEST CIRCUIT 8 : $V_{F-U}$ , $V_{F-L}$

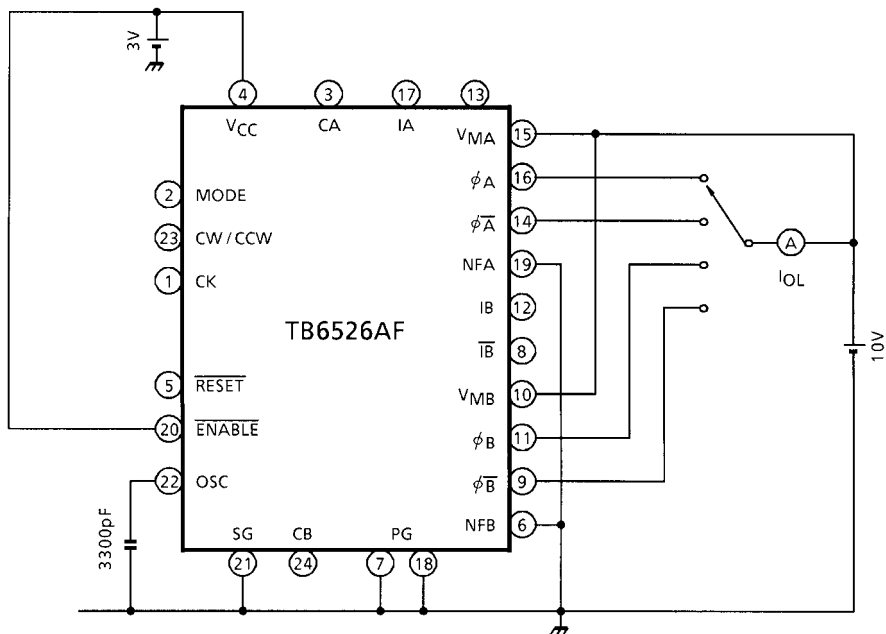


Note: Not to take GND with any non-connecting pins.

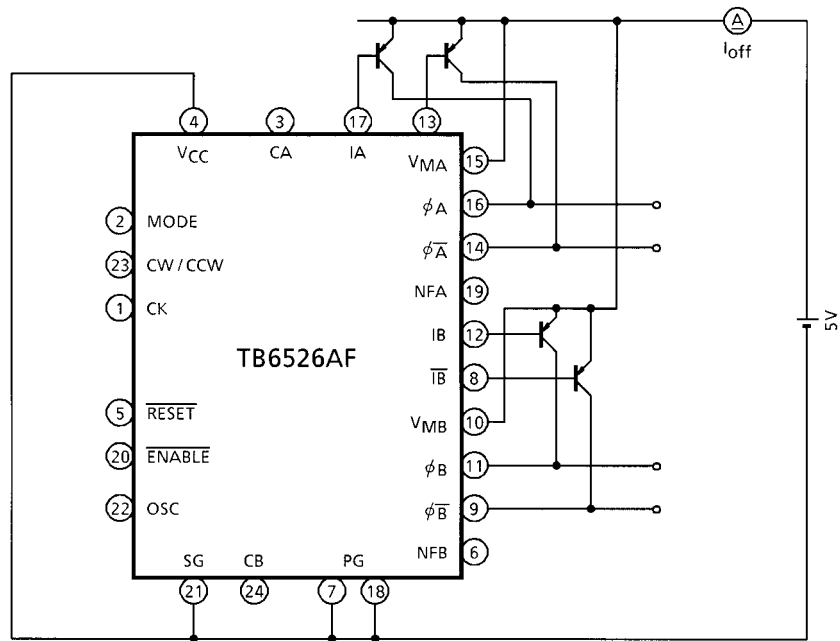
## TEST CIRCUIT 9 : $V_{NF1}$ , $\Delta V_{NF}$



## TEST CIRCUIT 10 : $I_{OL}$

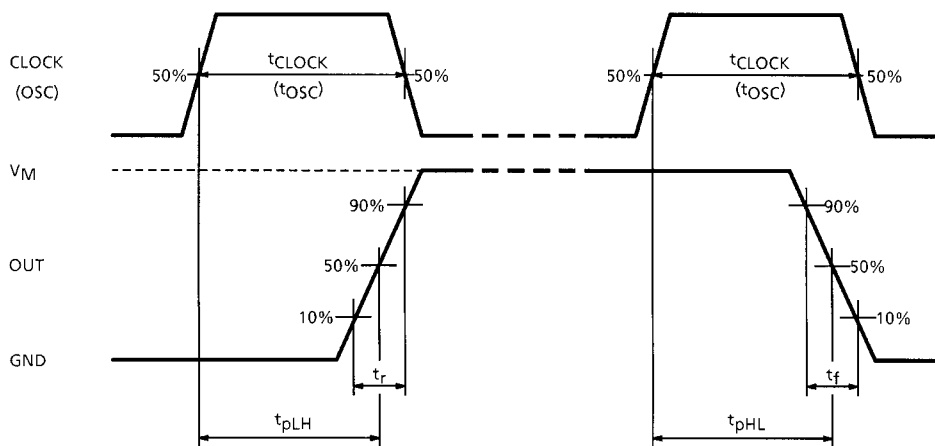


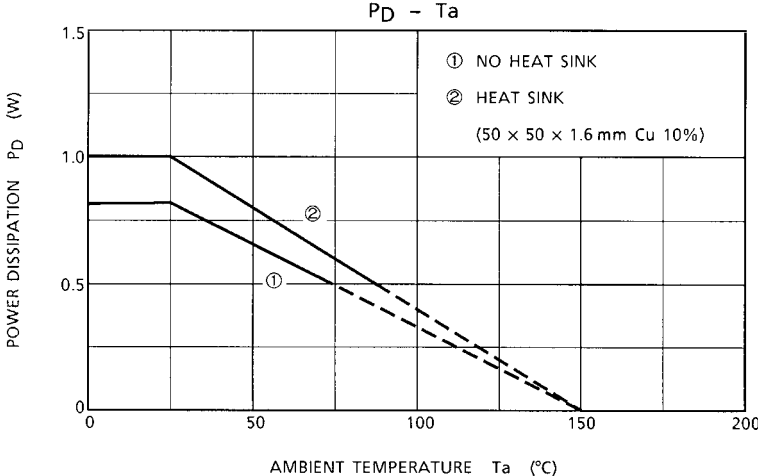
**TEST CIRCUIT 11**



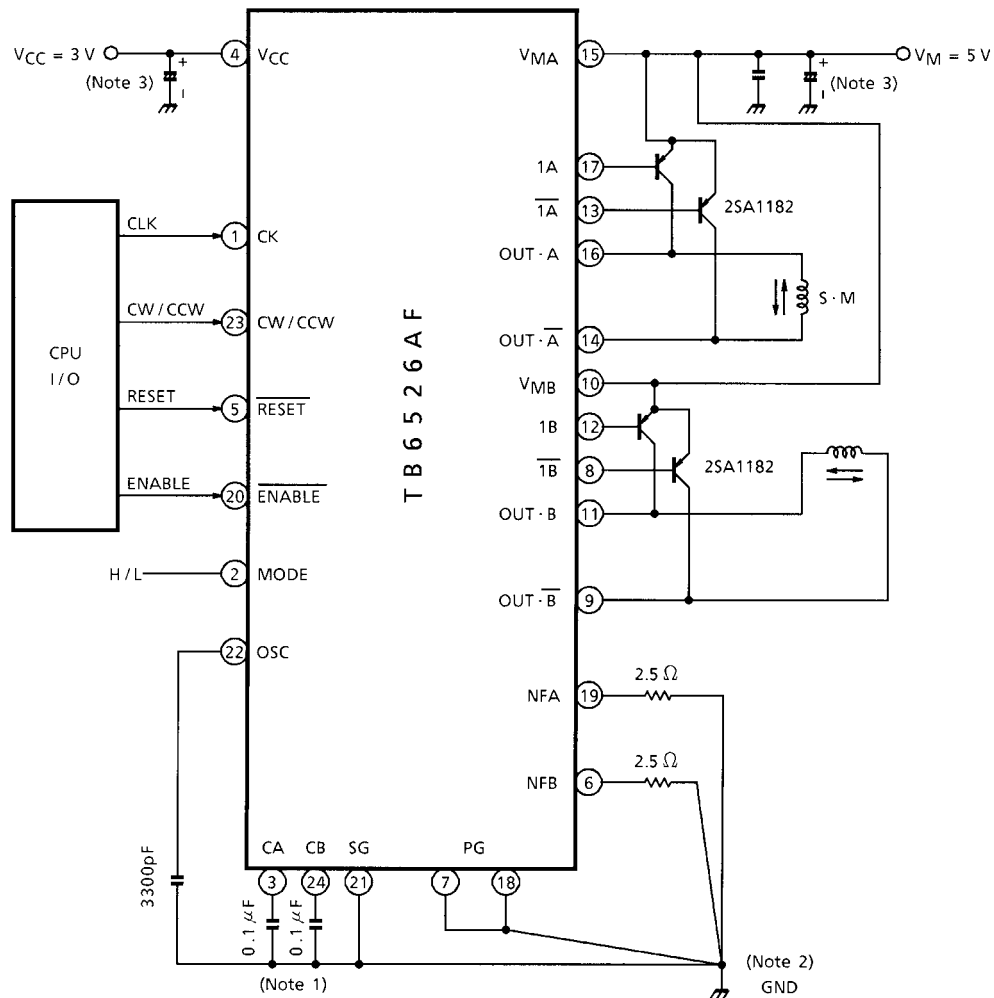
**AC ELECTRICAL CHARACTERISTICS, TEST CIRCUIT 12 CK (OSC) – OUT**

CK (OSC) – OUT





**APPLICATION CIRCUIT**



Note 1: A change in a step at the time of the micro-step can be improved smoothly with the capacitor of CA, CB.

Note 2: GND pattern to be laid out at one point in order to prevent common impedance.

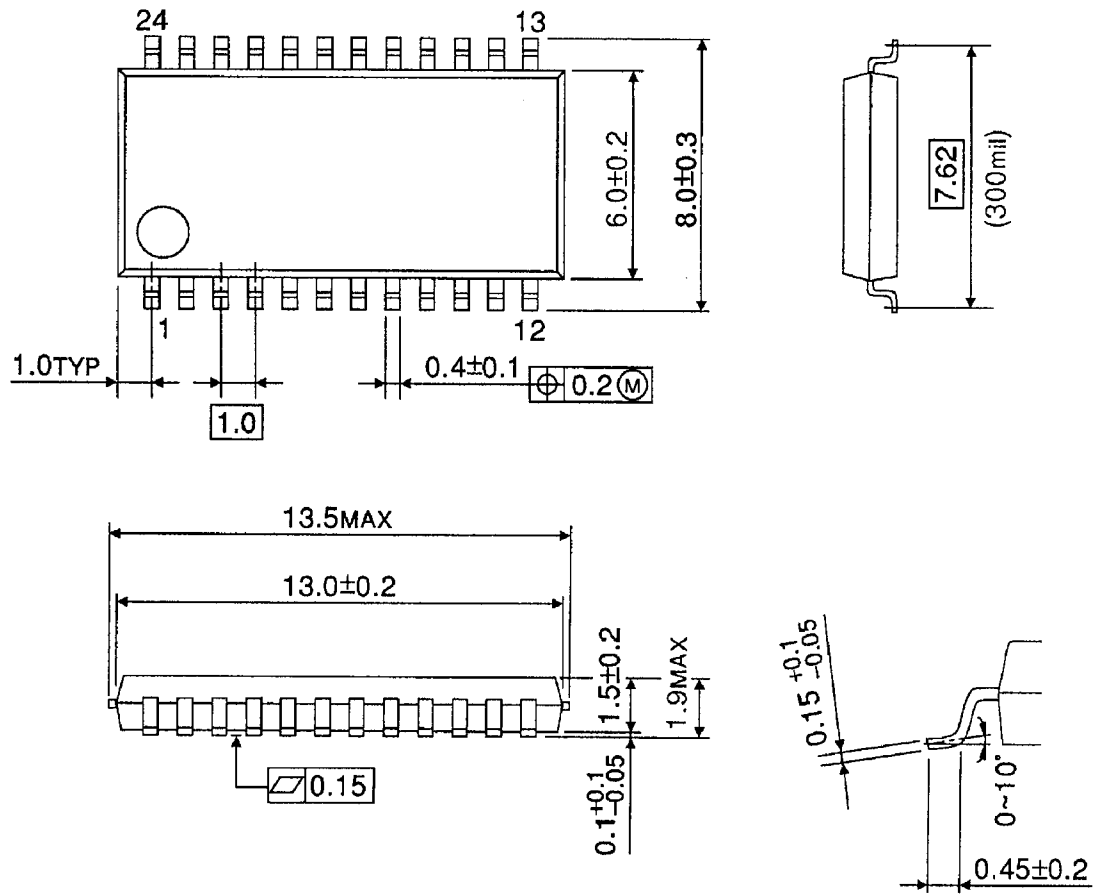
Note 3: Capacitor for noise suppression to be connected between the Power Supply (V<sub>CC</sub>, V<sub>M</sub>) and GND to stabilize the operation.

Note 4: Utmost care is necessary in the design of the output line, V<sub>M</sub> and GND line since IC may be destroyed due to short-circuit between outputs, air contamination fault, or fault by improper grounding.

## PACKAGE DIMENSIONS

SSOP24-P-300-1.00B

Unit: mm



Weight: 0.27 g (Typ.)

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000707EBA

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