## 3-1/2 Digit Auto-Ranging A/D Converter with Triplex LCD Drive and Display Hold Function

## FEATURES

- Auto-Range Operation for AC and DC Voltage and Resistance Measurements
- Two User Selected AC/DC Current Ranges 20 mA and 200 mA
- 22 Operating Ranges
- 9 DC/AC Voltage
- 4 AC/DC Current
- 9 Resistance and Low Power Ohms
- Display HOLD Function
- 3 1/2 Digit Resolution in Auto-Range Mode 1/2000
- Extended Resolution in Manual Mode .. 1/3000
- Memory Mode for Relative Measurements
$\pm 5 \%$ F.S.
- Internal AC to DC Conversion Op Amp
- Triplex LCD Drive for Decimal Points, Digits and Annunciators
- Continuity Detection and Piezoelectric Transducer Driver
- Compact Surface Mounted 64-pin Plastic Flat Package
- Low Drift Internal Reference $\qquad$ 75ppm $/{ }^{\circ} \mathrm{C}$
- 9V Battery Operation 10 mW
■ Low Battery Detection and LCD Annunciator
PIN CONFIGURATION



## GENERAL DESCRIPTION

The TC815 is a $31 / 2$ digit integrating analog-to-digital converter with triplex LCD display drive and automatic ranging. A display hold function is on-chip. Input voltage/ ohm attenuators ranging from 1 to $1 / 10,000$ are automatically selected. Five full-scale ranges are provided. The CMOS TC815 contains all the logic and analog switches needed to manufacture an auto-ranging instrument for ohms and voltage measurements. User selected 20 mA and 200 mA current ranges are available. Full-scale range and decimal point LCD annunciators are automatically set in auto-range operation. Auto-range operation is available during ohms (high and low power ohms) and voltage (AC and DC) measurements, eliminating expensive range switches in hand-held DMM designs. The auto-range feature may be bypassed allowing decimal point selection and input attenuator selection control through a single line input. Expensive rotary switches are not required.

During manual mode operation resolution is extended to 3000 counts full-scale. The extended range operation is indicated by a flashing 1 MSD. The extended resolution is also available during $200 \mathrm{k} \Omega$ and 2000 V full-scale autorange operation.

The memory mode subtracts a reading-up to $\pm 5 \%$ of full-scale-from subsequent measurements. Typical applications involve probe resistance compensation for resistance measurements, tolerance measurements, and tare weight measurements.

The TC815 includes an AC to DC converter for AC measurements. Only external diodes/resistors/capacitors are required.

A complete LCD annunciator set describes the TC815 meter function and measurement range during ohms, voltage and current operation. AC measurements are indicated as well as auto-range operation. A low battery detection circuit also sets the low battery display annunciator. The triplex LCD display drive levels may be set and temperature compensation applied via the $\mathrm{V}_{\text {DISP }}$ pin. With HOLD low the display is not updated. A HOLD mode LCD annunciator is activated.

## ORDERING INFORMATION

| Part No. | Package | Operating Temp. <br> Range |
| :--- | :--- | :---: |
| TC815CBU | 64-Pin PQFP <br>  <br>  <br> Formed Leads | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ |

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The "low ohms" measurement option allows in circuit resistance measurements by preventing semiconductor junctions from being forward biased.

A continuity buzzer output is activated with inputs less than $1 \%$ of full scale. An overrange input signal also enables the buzzer, except during resistance measurements, and flashes the MSD display. Featuring single 9V battery operation, 10 mW power consumption, a precision internal voltage reference ( $75 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max TC) and a compact surface mounted 64 -pin quad flat package, the TC815 is ideal for portable instruments.

## ABSOLUTE MAXIMUM RATINGS*

Supply Voltage ( $\mathrm{V}^{+}$to $\mathrm{V}^{-}$)........................................... 15 V
Analog Input Voltage ........................................... $\mathrm{V}^{+}$to $\mathrm{V}^{-}$
Reference Input Voltage ..................................... $\mathrm{V}^{+}$to $\mathrm{V}^{-}$
Voltage at Pin 45 ............................................GND $\pm 0.7 \mathrm{~V}$

Power Dissipation ( $\mathrm{T}_{\mathrm{A}} \leq 70^{\circ} \mathrm{C}$ )
Plastic Flat Package
1.14 mW

Operating Temperature
"C" Devices $\qquad$ $.0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ Storage Temperature ............................ $-65^{\circ} \mathrm{C}$ to $+160^{\circ} \mathrm{C}$ Lead Temperature (Soldering, 10 sec .) ................ $+300^{\circ} \mathrm{C}$ *Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions above those indicated in the operational sections of the specifications is not implied. Exposure to Absolute Maximum Rating Conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS: $\mathrm{V}_{\mathrm{S}}=9 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise specified, Figure 1 Test Circuit.

| Symbol | Parameter | Test Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Zero Input Reading Input Resistor | 200 mV Range w/o 10M $\Omega$ | -0000 | 0000 | +0000 | Digital <br> Reading |
|  |  | 200mV Range w/10M $\Omega$ Input | -0001 | - | +0001 |  |
|  |  | 20 mA and 200mA Range | -0000 | 0000 | +0000 |  |
| $\overline{R E}$ | Rollover Error | 200mV Range w/o $10 \mathrm{M} \Omega$ Input Resistor | - | - | $\pm 1$ <br> Count |  |
|  |  | 200mV Range w/10M $\Omega$ Input | - | - | $\pm 3$ |  |
|  |  | 20mA and 200mA Range | - | - | $\pm 1$ |  |
| NL | Linearity Error | Best Case Straight Line | - | - | $\pm 1$ | $\begin{gathered} \text { Count } \\ \hline \mathrm{pA} \\ \hline \end{gathered}$ |
| IIN | Input Leakage Current |  | - | - | 10 |  |
| $\mathrm{E}_{\mathrm{N}}$ | Input Noice | BW = 0.1 to 10 Hz | - | 20 | - | $\mu \mathrm{V}_{\text {p-p }}$ |
|  | AC Frequency Response | $\pm 1 \%$ Error | - | 40 to 500 | - | Hz |
|  |  | $\pm 5 \%$ Error | - | 40 to 2000 | - |  |
|  | Open Circuit Voltage | Excludes $200 \Omega$ Range for OHM Measurements | - | 570 | 660 | mV |
|  | Open Circuit Voltage | Excludes $200 \Omega$ Range for LO OHM Measurement | - | 285 | 350 | mV |
| $\mathrm{V}_{\text {COM }}$ | Analog Common Voltage | ( $\mathrm{V}^{+}-\mathrm{V}_{\text {COM }}$ ) | 2.5 | 2.6 | 3.3 | V |
| $\mathrm{V}_{\text {CTC }}$ | Common Voltage Temperature Coefficient |  | - | - | 50 | $\frac{\mathrm{ppm}}{{ }^{\circ} \mathrm{C}}$ |
|  | Display Multiplex Rate |  | - | 100 | - | Hz |
| $\overline{\mathrm{V} \text { IL }}$ | Low Logic Input | $\overline{20 \mathrm{~mA}}, \overline{\mathrm{AC}}, \mathrm{I}, \overline{\mathrm{Low} \Omega}, \overline{\mathrm{HOLD}}$ Range, $-\overline{\mathrm{MEM}}, \overline{\mathrm{OHM}}$ (Relative to DGND Pin 58) | - | - | 1 | V |
|  | Logic 1 Pull Up Current | $\overline{20 \mathrm{~mA}}, \overline{\mathrm{AC}}, \mathrm{I}$, LOW $\Omega$, HOLD Range, $-\overline{\mathrm{MEM}}, \overline{\mathrm{OHMs}}$ (Relative to DGND Pin 58) | - | 25 | - | $\mu \mathrm{A}$ |
|  | Buzzer Drive Frequency |  | - | 4 | - | kHz |
|  | Low Battery Flag Voltage | $\mathrm{V}_{\mathrm{CC}}$ to $\mathrm{V}_{\text {SSA }}$ | 6.3 | 6.60.8 | 7.0 | V |
|  | Operating Supply Current |  | - |  | 1.5 | mA |
| TC815-1 11/14/96 | 2 |  | © 2001 Microchip Technology Inc. |  |  | DS21474A |

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## PIN DESCRIPTIONS

| Pin No. (64-Pin Plastic) Quad Flat Package | Symbol | Description |
| :---: | :---: | :---: |
| 1 | NC |  |
| 2 | $\overline{\mathrm{OHM}}$ | Logic Input. "0" (Digital Ground) for resistance measurement. |
| 3 | $\overline{20 \mathrm{~mA}}$ | Logic Input. "0" (Digital Ground) for 20mA full-scale current measurement. |
| 4 | BUZ | Audio frequency, 4 kHz , output for continuity indication during resistance measurement. A noncontinuous 4 kHz signal is output to indicate an input overrange during voltage or current measurements, |
| 5 | XTAL1 | 32.768 kHz Crystal Connection |
| 6 | XTAL2 | 32.768 kHz Crystal Connection |
| 7 | $\mathrm{V}_{\text {DISP }}$ | Sets peak LCD drive signal: VP - VDD-VDISP. VDISP may also be used to compensate for temperature variation of LCD crystal threshold voltage. |
| 8 | BP1 | LCD Backplane \#1. |
| 9 | BP2 | LCD Backplane \#2. |
| 10 | BP3 | LCD Backplane \#3. |
| 11 | Lo $\Omega$ /A | LCD Annunciator segment drive for low ohms resistance measurement and current measurement. |
| 12 | $\Omega / \mathrm{V}$ | LCD Annunciator segment drive for resistance measurement and voltage measurement. |
| 13 | k/m/HOLD | LCD Annunciator segment drive for k ("kilo-ohms"), m ("milli-amps" and "milli-volts") and HOLD mode. |
| 14 | $\begin{gathered} \mathrm{BCPO} \\ \text { (Ones digit) } \end{gathered}$ | LCD segment drive for "b," "c" segments and decimal point of least significant digit (LSD). |
| 15 | ADG0 | LCD segment drive for "a," "g," "d" segments of LSD. |
| 16 | FEO | LCD segment drive for "f" and "e" segments of LSD. |
| 17 | NC |  |
| 18 | BCP1 | LCD segment drive for "b," "c" segments and decimal point of 2nd LSD. |
| 19 | AGD1 | LCD segment drive for "a," "g," "d" segments of 2nd LSD (Ten's digit). |
| 20 | FE1 | LCD segment drive for "f" and "e" segments of 2nd LSD. |
| 21 | BCP2 | LCD segment drive for "b," "c" segments and decimal point of 3rd LSD (Hundreds digit). |
| 22 | AGD2 | LCD segment drive for "a, " g , "d" segments of 3rd LSD. |
| 23 | FE2 | LCD segment drive for "b," "c" segments and decimal point of 3rd LSD. |
| 24 | BCP3 | LCD segment drive for "b," "c" segments and decimal point of MSD (Thousand's digit). |
| 25 | AC/-/AUTO | LCD annunciator drive signal for AC measurements, polarity, and auto-range operation. |
| 26 - | -MEM/BATT | LCD annunciator drive signal for low battery indication and memory (relative measurement) mode. |
| 27 | $V_{\text {SSD }}$ | Negative battery supply connection for internal digital circuits. Connect to negative terminal of battery. |
| 28 | $\mathrm{V}_{C C}$ | Positive battery supply connection |
| 29 A | ANALOG COM | Analog circuit ground reference point. Nominally 2.6 V below $\mathrm{V}_{\mathrm{cc}}$. |
| 30 | RM ${ }_{\text {REFH }}$ | Ratiometric (Resistance measurement) reference high voltage. |
| 31 | RM ${ }_{\text {Refl }}$ | Ratiometric (Resistance measurement) reference low voltage. |
| 32 | $\mathrm{C}_{\text {Refl }}$ | Reference capacitor negative terminal $\mathrm{C}_{\text {REF }} 0.1 \mu \mathrm{f}$. |
| 33 | Creft | Reference capacitor positive terminal CREF $0.1 \mu \mathrm{f}$. |
| 34 | ReFHI | Reference voltage for voltage and current measurement. Nominally 163.85 mV . |
| 35 | $\Omega$ R1 | Standard resistor connection for $200 \Omega$ full-scale. |
| 36 | $\Omega$ R2 | Standard resistor connection for $2000 \Omega$ full-scale. |
| 37 | $\Omega$ R3 | Standard resistor connection for $20 \mathrm{k} \Omega$ full-scale range. |
| 38 | $\Omega$ R4 | Standard resistor connection for 200k 2 full-scale range. |
| 39 | $\Omega$ R5 | Standard resistor connection for $2000 \mathrm{k} \Omega$ full-scale range. |
| 40 | VR3 | Voltage measurement $\div 100$ attenuator. |
| 41 | VR2 | Voltage measurement $\div 10$ attenuator. |

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## PIN DESCRIPTIONS (Cont.)

| Pin No. (64-Pin Plastic) Quad Flat Package | Symbol | Description |
| :---: | :---: | :---: |
| 42 | VR5 | Voltage measurement $\div 10,000$ attenuator. |
| 43 | VR4 | Voltage measurement $\div 1000$ attenuator. |
| 44 | $V_{1}$ | Unknown voltage input $\div 1$ attenuator. |
| 45 | II | Unknown current input. |
| 46 | ACVL | Low output of AC to DC converter. |
| 47 | $\mathrm{Cl}_{1}$ | Integrator capacitor connection. Nominally 0.1 f. (Low dielectric absorption. Polypropylene dielectrics suggested.) |
| 48 | NC |  |
| 49 | CAZ | Auto-zero capacitor connection. Nominally $0.1 \mu \mathrm{f}$. |
| 50 | RX | Unknown resistance input. |
| 51 | CFI | Input filter connection. |
| 52 | ADI | Negative input of internal AC to DC operational amplifier. |
| 53 | ADO | Output of internal AC to DC operational amplifier. |
| 54 | R $\Omega$ BUF | Active buffer output for resistance measurement. Integration resistor connection. Integrator resistor nominally $220 \mathrm{k} \Omega$. |
| 55 | RVIBUF | Active buffer output for voltage and current measurement. Integration resistor connection. Integration resistor nominally $150 \mathrm{k} \Omega$. |
| 56 | ACVH | Positive output of AC to DC converter. |
| 57 | $V_{\text {SSA }}$ | Negative supply connection for analog circuits. Connect to negative terminal of 9V battery. |
| 58 | $\overline{\text { DGND }}$ | Internal logic digital ground. The logic "0" level. Nominally 4.7V below $\mathrm{V}_{\mathrm{CC}}$. |
| 59 | RANGE | Input to set manual operation and change ranges. |
| 60 | HOLD | Input to hold display. Connect to DIG GND. |
| 61 | MEM | Input to enter memory measurement mode for relative measurements. The two LSD's are stored and subtracted from future measurements. |
| 62 | DC/AC | Input that selects AC or DC option during voltage/current measurements. For resistance measurements, $\overline{\Omega / L O W \Omega}$ the ohms or low power (voltage) ohms option can be selected. |
| 63 | $T$ | Input to select current measurement. Set to logic "0" (Digital ground) for current measurement. |
| 64 | NC |  |

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Figure 1: Typical Application and Test Circuit

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## TC815



Figure 2: TC815 Analog Section

## RESISTANCE, VOLTAGE, CURRENT MEASUREMENT SELECTION

The TC815 is designed to measure voltage, current, and resistance. Auto-ranging is available for resistance and voltage measurements. The OHMS (Pin 2) and $\bar{\top}$ (Pin 63) input controls are normally pulled internally to $\mathrm{V}_{\mathrm{cc}}$.

By tying these pins to Digital Ground (Pin 58), the TC815 is configured internally to measure resistance, voltage, or current. The required signal combinations are shown in Table 1.

## Table 1. TC815 Measurement Selection Logic

| Function Select Pin |  |  |
| :---: | :---: | :--- |
| OHM (Pin 2) | $\mathbf{I}$ (Pin 63) | Selected Measurement |
| 0 | 0 | Voltage |
| 0 | 1 | Resistance |
| 1 | 0 | Current |
| 1 | 1 | Voltage |
| $0=$ Digital Ground |  |  |
| $1=$ Floating or Tied to $V_{C C}$ |  |  |

## Notes:

1. OHM \& Tare normally pulled internally high to $\mathrm{V}_{\mathrm{CC}}$ (Pin 28). This is considered a logic " 1 ."
2. Logic " 0 " is the potential at digital ground (Pin 58).

## RESISTANCE MEASUREMENTS OHMS \& LOW POWER OHMS

The TC815 can be configured to reliably measure incircuit resistances shunted by semiconductor junctions. The TC815 low power ohms measurement mode limits the probe open circuit voltage. This prevents semiconductor junctions in the measured system from turning on.

In the resistance measurement mode the $\Omega / \overline{\mathrm{LOW} \Omega}$ (Pin 62) input selects the low power ohms measurement mode. For low power ohms measurements $\Omega / \mathrm{LOW} \Omega$ (Pin 62) is momentarily brought low to digital ground potential. The TC815 sets up for a low power ohms measurement with a maximum open circuit probe voltage of 0.35 V above analog common. In the low power ohms mode an LCD display annunciator, $\overline{\mathrm{LOW} \Omega}$, will be activated. On power up the low power ohms mode is not active.

If the manual mode has been selected, toggling $\bar{\Omega} /$ $\overline{L O W \Omega}$ will reset the TC815 back to the auto-range mode. In manual mode, the decision to make a normal or low power ohms measurement should be made before selecting the desired range.

The low power ohms measurement is not available on the $100 \Omega$ full-scale range. Open circuit voltage on this range is below 2.8 V .

The standard resistance values are listed in Table 2.
Table 2. Ohms Range Ladder Network

| Full-Scale <br> Range | Standard <br> Resistance | Low Power <br> Ohms Mode |
| :--- | :--- | :---: |
| $200 \Omega$ | $163.85 \Omega(\mathrm{R} 1)$ | NO |
| $2000 \Omega$ | $1638.5 \mathrm{k} \Omega(\mathrm{R} 2)$ | YES |
| $20 \mathrm{k} \Omega$ | $16,385 \Omega(\mathrm{R} 3)$ | YES |
| $200 \mathrm{k} \Omega$ | $16385 \Omega(\mathrm{R} 4)$ | YES |
| $2,000 \mathrm{k} \Omega$ | $1,638,500 \Omega(\mathrm{R} 5)$ | YES |

N/A = Not available.
R8, a positive temperature coefficient resistor, and the 6.2 V zener, Z 1 in Figure 1 provide input voltage protection during ohms measurements.

## RATIOMETRIC RESISTANCE MEASUREMENTS

The TC815 measures resistance ratiometrically. Accuracy is set by the external standard resistors connected to Pin 35 through 39. Alow-power ohms mode may be selected on all but the $200 \Omega$ full-scale range. The low power ohms mode limits the voltage applied to the measured system. This allows accurate "in-circuit" measurements when a resistor is shunted by semiconductor junctions.

Full auto-ranging is provided. External precision standard resistors are automatically switched to provide the proper range.

Figure 3 shows a detailed block diagram of the TC815 configured for ratiometric resistance measurements. During the signal integrate phase the reference capacitor charges to a voltage inversely proportional to the measured resis-tance-RX. Figure 4 shows the conversion accuracy relies on the accuracy of the external standard resistors only.

Normally the required accuracy of the standard resistances will be dictated by the accuracy specifications of the users end product. Table 3 gives the equivalent ohms per count for various full-scale ranges to allow users to judge the required resistor for accuracy.

Table 3. Reference Resistors

| Full-Scale <br> Range | Reference <br> Resistor | $\Omega /$ Count |
| :--- | :--- | :---: |
| 200 k | 163.85 | 0.1 |
| 2 k | 1638.5 | 1 |
| 20 k | 16385 | 10 |
| 200 k | 163850 | 100 |
| 2 M | 1638500 | 1000 |

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Figure 3: Ratiometric Resistance Measurement Functional Diagram

## VOLTAGE MEASUREMENT

Resistive dividers are automatically changed to provide in range readings for 200 mV to 2000 V full-scale readings (Figure 2). The input resistance is set by external resistors R14/R13. The divider leg resistors are R9-R12. The divider leg resistors give a 200 mV signal $\mathrm{V}_{\mathrm{I}}$ (Pin 44) for full-scale voltages from 200 mV to 2000 V .

For applications which do not require a $10 \mathrm{~m} \Omega$ input impedance the divider network impedances may be lowered. This will reduce voltage offset errors induced by switch leakage currents.

## CURRENT MEASUREMENT

The TC815 measures current only under manual range operation. The two user selectable full-scale ranges are: 20 mA and 200 mA . Select the current measurement mode by holding the Tinput (Pin 63) low at digital ground potential.

The $\overline{\mathrm{OHM}}$ input (Pin 2) is left floating or tied to the positive supply.

Two ranges are possible. The $\overline{20 \mathrm{~mA}}$ full-scale range is selected by connecting the $\overline{20 \mathrm{~mA}}$ input (Pin 3) to digital ground. If left floating the 200 mA full-scale range is selected.

External current to voltage conversion resistors are used at the $l_{\text {input }}$ (Pin 45). For $\overline{20 m A}$ measurements a $10 \Omega$ resistor is used. The 200 mA range needs a $1 \Omega$ resistor. Fullscale is 200 mV .

PC board trace resistance between analog common and R16 (See Figure 1) must be minimized. In the 200 mA range, for example, a 0.05 trace resistance will cause a $5 \%$ current to voltage conversion error at $I_{1}(\operatorname{Pin} 45)$.

The extended resolution measurement option operates during current measurements.

To minimize rollover error the potential difference between ANALOG COMMON (Pin 29) and system common must be minimized.

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## Example: $200 \mathrm{k} \Omega$ Full-Scale Measurement

(a) $\quad V R=\frac{163.85 k \Omega}{163.85+220+R X} \times 0.64$
(b) $\quad V X=\frac{R X}{163.85 \mathrm{k} \Omega+220 \Omega+\mathrm{RX}} \times 0.64$
(c) "Ramp Up Voltage" = "Ramp Down Voltage"
$\therefore \frac{V X}{R_{l} C_{I}} \quad \times T_{I}=\frac{V R}{R_{l} C_{I}} \quad T_{D E}$
Where:
$\mathrm{R}_{I}=$ Integrating Resistor, $\mathrm{T}_{I}=$ Integrate Time
$\mathrm{C}_{I}=$ Integrating Capacitor, $\mathrm{T}_{\mathrm{DE}}=$ Deintegrate Time
(d) $R X=163.85\left(\frac{T_{D E}}{T_{I}}\right)$

Independent of $\mathrm{R}_{1}$, $\mathrm{C}_{\mathrm{l}}$ or Internal Voltage Reference

Figure 4. Resistance Measurement Accuracy Set by External Standard Resistor


```
External Crystal \(=32.768 \mathrm{kHz}\)
Internal Clock Period \(=T_{P}=2 / 32.768=61.04 \mu \mathrm{sec}\)
Total Conversion Time \(=T_{\text {CONV }}=8000\left(T_{P}\right)\)
\(=488.3 \mathrm{msec} \approx 2\) CONV/SEC
Integration Time \(=T_{I}=1638.5\left(T_{P}\right)=100.0 \mathrm{msec}\)
Maximum Reference Deintegrate Time \(=\)
\(T_{D E}=3000\left(T_{P}\right)=183.1 \mathrm{~ms}\) (Manual, Extended Resolution)
\(=2000\left(T_{P}\right)=122.1 \mathrm{msec}\) (Auto-Range)
Minimum Auto-Zero Time
\(=(8000-3000-1638.5)(T P)=205.1 \mathrm{msec}\) (Manual, Extended
Resolution)
\(=(8000-2000-1638.5) \quad\left(T_{P}\right)=266.2 \mathrm{msec}\) (Auto-Range)
```

Figure 5. Basic TC815 Conversion Timing

## MEASUREMENT OPTIONS AC TO DC MEASUREMENTS

In voltage and current measurements the TC815 can be configured for AC measurements. An on chip operational amplifier and external rectifier components perform the AC to DC conversion.

When power is first applied the TC815 enters the DC measurement mode. For AC measurements (current or voltage), $\overline{\mathrm{AC}} / \mathrm{DC}$ (Pin 62) is momentarily brought low to digital ground potential; the TC815 sets-up for AC measurements and the AC liquid crystal display annunciator activates. Toggling AC/DC low again will return the TC815 to DC operation.

If the manual operating mode has been selected toggling AC/DC will reset the TC815 back to the auto-range mode. In manual mode operation AC or DC operation should be selected first and then the desired range selected.

The minimum AC voltage full-scale voltage range is 2 V . The DC full-scale minimum voltage is 200 mV .

AC current measurements are available on the 20 mA and 100 mA full-scale current ranges.

## CONVERSION TIMING

The TC815 analog-to-digital converter uses the conventional dual slope integrating conversion technique with an added phase that automatically eliminates zero offset errors. The TC815 gives a zero reading with a zero volt input.

The TC815 is designed to operate with a 32.768 kHz crystal. The 32 kHz crystal is low cost and readily available; it serves as a time base oscillator crystal in many digital clocks. (See External Crystal Sources.)

The external clock is divided by two. The internal clock frequency is 16.348 kHz giving a clock period of $61.04 \mu \mathrm{sec}$. The total conversion - auto-zero phase, signal integrate

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and reference deintegrate - requires 8000 clock periods or 488.3 msec . There are approximately two complete conversions per second.

The integration time is fixed at 1638.5 clock periods or 100 msec . This gives rejection of $50 / 60 \mathrm{~Hz}$ AC line noise.

The maximum reference deintegrate time, representing a full-scale analog input, is 3000 clock periods or 183.1 msec during manual extended resolution operation. The 3000 counts are available in manual mode, extended resolution operation only. In auto-ranging mode, the maximum deintegrate time is 2000 clock periods. The 1000 clock periods are added to the auto-zero phase. An auto-ranging or manual conversion takes 8000 clock periods. After a zero crossing is detected in the reference deintegrate mode, the auto-zero phase is entered.

Figure 5 shows the basic TC815 timing relationships.

## MANUAL RANGE SELECTION

The TC815 voltage and resistance auto-ranging feature can be disabled by momentarily bringing RANGE (Pin 59) to digital ground potential (Pin 58). When the change from auto-to-manual ranging occurs the first manual range selected is the last range in the auto-ranging mode.

The TC815 power-up circuit selects auto-range operation initially. Once the manual range option is entered, range changes are made by momentarily grounding the RANGE
control input. The TC815 remains in the manual range mode until the measurement function (voltage or resistance) or measurement option ( $\mathrm{AC} / \mathrm{DC}, \Omega / \mathrm{LO} \Omega$ ) changes. This causes the TC815 to return to auto-ranging operation.

The "Auto" LCD annunciator driver is active only in the auto-range mode.

Table 4 shows typical operation where the manual range selection option is used. Also shown is the extended resolution display format.

## EXTENDED RESOLUTION MANUAL OPERATION

The TC815 extends resolution by $50 \%$ when operated in the manual range select mode for current, voltage, and resistance measurements. Resolution increases to 3000 counts from 2000 counts. The extended resolution feature operates only on the $2000 \mathrm{k} \Omega$ and 2000 V ranges during autorange operation.

In the extended resolution operating mode readings above 1999 are displayed with a blinking " 1 " most significant digit. The blinking " 1 " should be interpreted as the digit 2. The three least significant digits display data normally.

An input overrange condition causes the most significant digit to blink and sets the three least significant digits to display " 000 ." The buzzer output is enabled for input voltage and current signals with readings greater than 2000 counts in both manual and auto-range operation.

Table 4. Manual Range Operation

|  |  |  |  | VOLTS | AC | OLTS | OH |  | LO OH |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INPU |  |  |  | 3.5 V |  | 2V | 18.2 |  | 2.35M |  |
|  |  |  | RANGE | DISPLAY | RANGE | DISPLAY | RANGE | DISPLAY | RANGE | DISPLAY |
| POW | R- |  | 200mV | "1"00.0mV | 2 V | "1".000V | $200 \Omega$ | "1"00.0 | $2 \mathrm{k} \Omega$ | "1".000k $\Omega$ |
| AUTO | -R |  | 2 V | "1".000V | 20V | 18.20 V | $2 \mathrm{k} \Omega$ | "1".000k | 10k $\Omega$ | "1"0.00k $\Omega$ |
| OPER | RAT |  | 20 V | "1"0.00V |  |  | 20k | $18.20 \mathrm{k} \Omega$ | $200 \mathrm{k} \Omega$ | "1"00.0k $\Omega$ |
|  |  |  | 200 V | 23.5 V |  |  |  |  | 2000k $\Omega$ | "1"350k $\Omega$ |
|  |  |  | RANGE C | ANGES |  |  |  |  |  |  |
|  | 0 | 乙 | RANGE | DISPLAY | RANGE | DISPLAY | RANGE | DISPLAY | RANGE | DISPLAY |
|  | P | 1 | 200 V | 23.5 V | 20V | 18.20 V | 20k $\Omega$ | 18.20k $\Omega$ | 2000k $\Omega$ | "1"350k $\Omega$ |
|  | E | 2 | 200mV | "1"00.0V | 2 V | "1".000V | $200 \Omega$ " | "00.0 $\Omega \mathrm{k} \Omega$ | $2 \mathrm{k} \Omega$ | "1".000k $\Omega$ |
|  | R | 3 | 2 V | 1.000 V | 20V | 18.20 V | $2 \mathrm{k} \Omega$ | "1".000k | 20k $\Omega$ | "1"0.00k $\Omega$ |
| U | A | 4 | 20 V | "1"3.50V | 200V | 18.2 V | $20 \mathrm{k} \Omega$ | $18.20 \mathrm{k} \Omega$ | $200 \mathrm{k} \Omega$ | "1"00.0k $\Omega$ |
|  | T | 5 | 200 V | 23.5 V | 600V | 19 V | 200k $\Omega$ | $18.2 \mathrm{k} \Omega$ | 2000k $\Omega$ | "1"350k $\Omega$ |
|  | 1 | 6 | 1000 V | 24 V | 2 V | "1".000V | 2000k $\Omega$ | 19k $\Omega$ | $2 \mathrm{k} \Omega$ | "1".000k $\Omega$ |
|  | O | 7 | 200mV | "1"00.0mV | 20V | 18.20 V | $200 \Omega$ | "1"00.0 $\Omega$ | 20k $\Omega$ | " 1 "0.00 k $\Omega$ |
|  | N | 8 | 2 V | "1".000V | 200V | 18.2 V | $2 \mathrm{k} \Omega$ | "1".000k $\Omega$ | $200 \mathrm{k} \Omega$ | "1"00.0k $\Omega$ |

## Notes:

1. A flashing MSD is shown as a " 1 ". A flashing MSD indicates the TC815 is over-ranging if all other digits are zero
2. The first manual range selected is the last range in the auto-ranging mode.
3. A flashing MSD with a non-zero display indicates the TC815 has entered the extended resolution operating mode. An additional 1000 counts of resolution is available. This extended operation is available only in manual operation for voltage, resistance and current measurements.
4. $\checkmark$ = momentary ground connection.

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* MODE ALSO OPERATES WHEN AUTO-RANGING OPERATION IS SELECTED AND $2 \mathrm{M} \Omega<\mathrm{RX}<2.999 \mathrm{M} \Omega$

Figure 6. Manual Range Selection; Resistance Measurements


Figure 7. Manual Range Selection; Current Measurements
For resistance measurements the buzzer signal does not indicate an overrange condition. The buzzer is used to indicate continuity. Continuity is defined as a resistance reading less than 19 counts.


Figure 8. Manual Range Selection; Voltage Measurements

## -MEM OPERATING MODE

Bringing MEM (Pin 61) momentarily low configures the TC815 "-MEM" operating mode. The -MEM LCD Annunciator becomes active. In this operating mode subsequent measurements are made relative to the last two digits ( $\leq 99$ ) displayed at the time MEM is low. This represents $5 \%$ of fullscale. The last two significant digits are stored and subtracted from all the following input conversions.
A few examples clarify operation:

## Example 1: In Auto-Ranging

$\operatorname{Ri}(\mathrm{N})=18.21 \mathrm{k} \Omega(20 \mathrm{k} \Omega$ Range $)=>$ Display $18.21 \mathrm{k} \Omega$
$\overline{\text { MEM }} \checkmark\ulcorner=>$ Store $0.21 \mathrm{k} \Omega$
$\mathrm{Ri}(\mathrm{N}+1)=19.87 \mathrm{k} \Omega$ (20 k $\Omega$ Range)

$$
=>\text { Display } 19.87-0.21=19.66 \mathrm{k} \Omega
$$

$\mathrm{Ri}(\mathrm{N}+2)=22.65 \mathrm{k} \Omega(200 \mathrm{k} \Omega$ Range $)$
$=>$ Display $22.7 \mathrm{k} \Omega$ \& MEM Disappears
Example 2: In Fixed Range $200.0 \Omega$ Full-Scale
$\operatorname{Ri}(N)=18.2 \Omega=>$ Display $18.2 \Omega$
$\overline{\text { MEM }} \neg \longleftarrow=>$ Store $8.2 \Omega$
$\operatorname{Ri}(N+1)=36.7 \Omega$

$$
=>\text { Display } 36.7-8.2=28.5 \Omega
$$

$\operatorname{Ri}(N+2)=5.8 \Omega$

$$
=>\text { Display } 5.8-8.2=-2.4 \Omega^{*}
$$

* Will display minus resistance if following input is less than offset stored at fixed range.


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## Example 3: In Fixed Range

$$
\begin{aligned}
& \mathrm{Vi}(\mathrm{~N})=0.51 \mathrm{~V}=>\text { Display } 0.51 \mathrm{~V} \\
& \overline{\mathrm{MEM}} \mathrm{~V}=>\text { Store } 0.51 \mathrm{~V} \\
& \mathrm{Vi}(\mathrm{~N}+1)=3.68 \mathrm{~V} \\
& \quad=>\text { Display } 3.68-0.51=3.17 \mathrm{~V} \\
& \mathrm{Vi}(\mathrm{~N}+2)=0.23 \mathrm{~V} \\
& \quad=>\text { Display } 0.23-0.51=-0.28 \mathrm{~V} \\
& \mathrm{Vi}(\mathrm{~N}+3)=-5.21 \mathrm{~V} \\
& \quad=>\text { Display }-5.21-0.51=-5.72 \mathrm{~V}
\end{aligned}
$$

On power up the TC815 "-MEM" mode is not active. Once the "-MEM" is entered bringing MEM low again it returns the TC815 to normal operation.

The "-MEM" mode is also cancelled whenever the measurement type (resistance, voltage, current $\overline{\mathrm{AC}} / \mathrm{DC}, \Omega /$ $\overline{\mathrm{LO}} \Omega$ ) or range is changed. The LCD -MEM annunciator will be off in normal operation.

In the auto-range operation if the following input signal cannot be converted on the same range as the stored value, the "-MEM" mode is cancelled. The LCD annunciator is turned off.
the "-MEM" operating mode can be very useful in resistance measurements when lead length resistance would cause measurement errors.

## AUTOMATIC RANGE SELECTION OPERATION

When power is first applied the TC815 enters the autorange operating state. The auto-range mode may be entered from manual mode by changing the measurement function (resistance or voltage) or by changing the measurement option ( $\overline{\mathrm{AD}} / \mathrm{DC}, \Omega / \overline{\mathrm{LO} \Omega}$ ).

The automatic voltage range selection begins on the most sensitive scale first: 200 mV for DC or 2.000 V for AC measurements. The voltage range selection flow chart is given in Figure 9.

Internal input protection diodes to $\mathrm{V}_{\mathrm{DD}}$ (Pin 28) and $\mathrm{V}_{\text {SSA }}$ (Pin 57) clamp the input voltage. The external $10 \mathrm{M} \Omega$ input resistance (See Figure 1, R14 and R13) limits current safely in an overrange condition.

The voltage range selection is designed to maximize resolution. For input signals less than $9 \%$ of full-scale (count reading <180) the next most sensitive range is selected.

An overrange voltage input condition is flagged whenever the internal count exceeds 2000 by activating the buzzer output (Pin 4). This 4 kHz signal can directly drive a piezo electric acoustic transducer. An out of range input signal causes the 4 kHz signal to be on 122 msec , off for 122 ms , on for 122 msec and off for 610 msec (See Figure 15).

During voltage auto-range operation the extended resolution feature operates on the 2000 V range only (See extended resolution operating mode discussion).

The resistance automatic range selection procedure is shown in Figure 10. The $200 \Omega$ range is the first range selected unless the TC815 low ohms resistance measurement option is selected. In low ohms operation the first fullscale range tried is $2 \mathrm{k} \Omega$.

The resistance range selected maximizes sensitivity. If the conversion results in a reading less than 180 the next most sensitive full-scale range is tried.

If the conversion is less than 19 in auto-range operation a continuous 4 kHz signal is output at BUZ (Pin 4). An


Figure 9. Auto-Range Operation; Voltage Measurement

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Figure 10. Auto-Range Operation; Resistance Measurement


Figure 11. Low Battery Detector
overrange input does not activate the buzzer.
Out of range input conditions are displayed by a blinking most significant digit with the three least significant digits set to "000."

The extended resolution feature operates only on the $2000 \mathrm{k} \Omega$ and 2000 V full-scale range during auto-range operation. A blinking " 1 " most significant digit is interpreted as the digit 2. The three least significant digits display data normally.

## LOW BATTERY DETECTION CIRCUIT

The TC815 contains a low battery detector. When the 9 V battery supply has been depleted to a 7 V nominal value the LCD display low battery annunciator is activated.

The low battery detector is shown in Figure 11. The low battery annunciator is guaranteed to remain OFF with the battery supply greater than 0.7 V . The annunciator is guaranteed to be ON before the supply battery has reached 6.3V.

## TRIPLEX LIQUID CRYSTAL DRIVE

The TC815 directly drives a triplexed liquid crystal display (LCD) using $1 / 3$ bias drive. All data, decimal point, polarity and function annunciator drive signals are developed by the TC815. A direct connection to a triplex LCD display is possible without external drive electronics. Standard and custom LCD displays are readily available from LCD manufacturers.

The LCDs must be driven with an AC signal having zero DC component for long display life. The liquid crystal polarization is a function of the RMS voltage appearing across the backplane and segment driver. The peak drive signal applied to the LCD is: $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\mathrm{DISP}}$.


Figure 12. 1/3 Bias LCD Drive

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If $\mathrm{V}_{\text {DISP }}$, for example, is set at a potential 3 V below $\mathrm{V}_{\mathrm{CC}}$ the peak drive signal is:

$$
V p=V_{C C}-V_{D I S P}=3 V
$$

An "OFF" LCD segment has an RMS voltage of $\mathrm{Vp} / 3$ across it or 1 volt. An "ON" segment has a 0.63 V p signal across it or 1.92 V for $\mathrm{V}_{\mathrm{CC}}-\mathrm{V}_{\text {DISP }}=3 \mathrm{~V}$.

Since the $V_{\text {DISP }}$ pin is available, the user may adjust the "ON" and "OFF" LCD levels for various manufacturer's displays by changing Vp . The liquid crystal threshold voltage moves down with temperature.
"OFF" segments may become visible at high LCD operating temperatures. A voltage with a -5 to $-20 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ temperature coefficient can be applied to $\mathrm{V}_{\text {DISP }}$ to accommodate the liquid crystal temperature operating characteristics if necessary.

The TC815 internally generates two intermediate LCD drive potentials $\left(\mathrm{V}_{\mathrm{H}} \& \mathrm{~V}_{\mathrm{L}}\right)$ from a resistive divider (Figure 12) between $\mathrm{V}_{\mathrm{CC}}$ (Pin 28) and $\mathrm{V}_{\text {DISP }}$ (Pin 7). The ladder impedance is approximately $150 \mathrm{k} \Omega$. This drive method is commonly known as $1 / 3$ bias. With $V_{\text {DISP }}$ connected to digital ground $V_{p} \approx 5.0 \mathrm{~V}$.

The intermediate levels are needed so that drive signals giving RMS "ON" and "OFF" levels can be generated. Figure 13 shows a typical drive signal and the resulting wave forms for "ON" and "OFF." RMS voltage levels across a selected LCD element.


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Figure 14. Typical LCD Display Configuration TC815 Triplex

| PAD | BP1 | BP2 | BP3 | PAD | COM1 | COM2 | COM3 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | BP1 | $/$ | $/$ | 19 | $/$ | $/$ | $/$ |
| 2 | $/$ | BP2 | $/$ | 20 | $/$ | $/$ | $/$ |
| 3 | $/$ | $/$ | BP3 | 21 | $/$ | $/$ | $/$ |
| 4 | $/$ | LO $\Omega$ | A | 22 | $/$ | $/$ | $/$ |
| 5 | $/$ | $\Omega$ | V | 23 | $/$ | $/$ | $/$ |
| 6 | HOLD | k | m | 24 | $/$ | $/$ | $/$ |
| 7 | b 1 | c 1 | $/$ | 25 | $/$ | $/$ | $/$ |
| 8 | a 1 | g 1 | d 1 | 26 | $/$ | $/$ | $/$ |
| 9 | f 1 | e 1 | $/$ | 27 | $/$ | $/$ | $/$ |
| 10 | b 2 | c 2 | P 2 | 28 | $/$ | $/$ | $/$ |
| 11 | a 2 | g 2 | d 2 | 29 | $/$ | $/$ | $/$ |
| 12 | f 2 | e 2 | $/$ | 30 | $/$ | $/$ | $/$ |
| 13 | b 3 | c 3 | P 3 | 31 | $/$ | $/$ | $/$ |
| 14 | a 3 | g 3 | d 3 | 32 | $/$ | $/$ | $/$ |
| 15 | f 3 | e 3 | $/$ | 33 | $/$ | $/$ | $/$ |
| 16 | b 4 | c 4 | P 4 | 34 | $/$ | $/$ | $/$ |
| 17 | AC | $\square$ | Auto | 35 | $/$ | $/$ | $/$ |
| 18 | $\square$ | -MEM | $/$ | 36 | $/$ | $/$ | $/$ |

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## TC815



Figure 15. TC815 Timing Waveform for Buzzer Output

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## EXTERNAL CRYSTAL

The TC815 is designed to operate with a $32,768 \mathrm{~Hz}$ crystal. This frequency is internally divided by two to give a $61.04 \mu \mathrm{sec}$ clock period. One conversion takes 8000 clock periods or 488.3 msec ( $\approx 2$ conversions $/$ second). Integration time is 1638.5 clock periods or 100 msec .

The 32 kHz quartz crystal is readily available and inexpensive. The 32 kHz crystal is commonly used in digital clocks and counters.

Several crystal sources exist. A partial listing is:

- Statek Corporation

512 N. Main
Orange, CA 92668
(714) 639-7810

TWX: 910-593-1355
TELEX: 67-8394

- Daiwa Sinku Corporation

1389, shinzaike - AZA-Kono
Hirakacho, Kakogawa Hyogo, Japan
Tel: 0794-26-3211

- International Piezo LTD

24-26, Sze Shan Street
Yau Ton, Hong Kong
TLX: 35454 XTAL HX
Tel: 3-3501151
Contact manufacturer for full specifications.

## "BUZZER" DRIVE SIGNAL

The TC815 BUZ output (Pin 4) will drive a piezo electric audio transducer. The signal is activated to indicate an input overrange condition for current and voltage measurements or continuity during resistance measurements.

During a resistance measurement a reading less than 19 on any full-scale range causes a continuous 4 kHz signal to be output. This is used as a continuity indication.

A voltage or current input measurement overrange is indicated by a noncontinuous 4 kHz signal at the BUZ output. The LCD display MSD also flashes and the three least significant digits are set to display zero. The buzzer drive signal for overrange is shown in Figure 15. The buzzer output is active for any reading over 2000 counts in both manual and auto-range operation. The buzzer is activated during an extended resolution measurement.

The BUZ signal swings from $\mathrm{V}_{\mathrm{CC}}$ (Pin 28) to Digital Ground (Pin 58). The signal is at $\mathrm{V}_{\mathrm{CC}}$ when not active.

The buz output is also activated for 15 msec whenever a range change is made in auto-range or manual operation. Changing the type of measurement (voltage, current, or resistance) or measurement option (AC/DC, $\Omega / \mathrm{LO} \Omega$ ) will
also activate the buzzer output for 15 msec . A range change during a current measurement will not activate the buzzer output.

Vendors for piezo electric audio transducers are:

- Gulton Industries

Piezo Products Division
212 Durham Avenue
Metuchen, New Jersey 08840
(201) 548-2800

Typical P/Ns: 102-95NS, 101-FB-00

- Taiyo Yuden (USA) Inc.

Arlington Center
714 West Algonquin Road
Arlington Heights, lllinois 60005
Typical P/Ns: CB27BB, CB20BB, CB355BB

## Display Decimal Point Selection

The TC815 provides a decimal point LCD drive signal. The decimal point position is a function of the selected fullscale range, as shown in Table 5.

Table 5. Decimal Point Selection

| Full-Scale Range | $\mathbf{1}$ | ${ }^{*}$ | $\mathbf{9}$ | ${ }^{*}$ | $\mathbf{9}$ |
| :--- | :---: | :---: | :---: | :---: | :---: |${ }^{*}{ }^{*}$| $\mathbf{9}$ |
| :---: |
| $2000 \mathrm{~V}, 2000 \mathrm{k} \Omega$ |
| $200 \mathrm{~V}, 200.0 \mathrm{k} \Omega$ |

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## AC-to-DC Converter Operational Amplifier

The TC815 contains an on-chip operational amplifier that may be connected as a rectifier for AC-to-DC voltage and current measurements. Typical operational amplifier characteristics are:

- Slew Rate: $1 \mathrm{~V} / \mu \mathrm{sec}$
- Unity-Gain Bandwidth: 0.4 MHz
- Open-Loop Gain: 44dB
- Output Voltage Swing (Load = $10 \mathrm{k} \Omega) \pm 1.5 \mathrm{~V}$
(Referenced to Analog Common)
When the AC measurement option is selected, the input buffer receives an input signal through switch S 14 rather than switch S11 (see Figure 1). With external circuits, the AC operating mode can be used to perform other types of functions within the constraints of the internal operational amplifier. External circuits that perform true RMS conversion or a peak hold function are typical examples.


## Component Selection

## Integration Resistor Selection

The TC815 automatically selects one of two external integration resistors. RVBUF (pin 55) is selected for voltage and current measurement. R $\Omega$ BUF (pin 54) is selected for resistance measurements.

## RVBUF Selection (Pin 55)

In auto-range operation, the TC815 operates with a 200 mV maximum full-scale potential at $\mathrm{V}_{\mathrm{I}}$ (pin 44). Resistive dividers at VR2 (pin 41), VR3 (pin 40), VR4 (pin 43), and VR5 (pin 42) are automatically switched to maintain the 200VV full-scale potential.

In manual mode, the extended operating mode is activated giving a 300 mV full-scale potential at $\mathrm{V}_{1}$ (pin 44).

The integrator output swing should be maximized, but saturations must be avoided. The integrator will swing within 0.45 V of $\mathrm{V}_{\mathrm{CC}}$ (pin 28) and 0.5 V of $\mathrm{V}_{\text {SS }}$ (pin 57) without saturating. $\mathrm{A} \pm 2 \mathrm{~V}$ swing is suggested. The value of RVBUF is easily calculated, assuming a worst-case extended resolution input signal:

$$
\begin{aligned}
& \mathrm{V}_{\text {INT }}=\text { Integrator swing }= \pm 2 \mathrm{~V} \\
& \mathrm{t}_{1} \quad=\text { Integration time }=100 \mathrm{msec} \\
& \mathrm{C}_{\mathrm{I}} \quad=\text { Integration capacitor }=0.1 \mu \mathrm{~F} \\
& \mathrm{~V}_{\text {MAX }}=
\end{aligned}
$$

$$
\text { RVBUF }=\frac{\mathrm{V}_{\operatorname{MAX}}\left(\mathrm{T}_{\mathrm{l}}\right)}{\mathrm{V}_{\text {INT }}\left(\mathrm{C}_{\mathrm{l}}\right)} \approx 150 \mathrm{k} \Omega
$$

## R $\Omega$ BUF Selection (Pin 54)

In ratiometric resistance measurements, the signal at $R_{X}$ (pin 50) is always positive with respect to analog common. The integrator swings negative.

The worst-case integrator swing is for the $200 \Omega$ range with the manual, extended resolution option.

The input voltage, $\mathrm{V}_{\mathrm{x}}$ (pin 50 ) is easily calculated (Figure 16):
$\mathrm{V}_{\text {ANCOM }}=$ Potential at Analog Common $\approx 2.7 \mathrm{~V}$
$\mathrm{R}_{8}=220 \Omega$
$R_{I}=163.85 \Omega$
$\mathrm{R}_{\mathrm{X}}=300 \Omega$
$\mathrm{R}_{\mathrm{S}} \quad=$ Internal switch 33 resistance $\approx 600 \Omega$
$R \Omega B U F=\frac{\left(V_{C C}-V_{\text {ANCOM }}\right) R_{X}}{\left(R_{X}+R_{S}+R_{1}+R_{8}\right)}=0.63 V$

For a 3.1 V integrator swing, the value of R $\Omega B \mathrm{BUF}$ is easily calculated:
$\mathrm{V}_{\text {INT }} \quad=$ Integrator swing $=3.1 \mathrm{~V}$
$\mathrm{t}_{1} \quad=$ Integration time $=100 \mathrm{msec}$
$\mathrm{C}_{\mathrm{I}} \quad=$ Integration capacitor $=0.1 \mu \mathrm{~F}$
$R_{X} \operatorname{Max}=300 \Omega$
$V_{X}$ Max $=700 \mathrm{mV}$
R $\Omega B U F=\frac{\left(V_{X} \text { MAX }\right)\left(T_{1}\right)}{\mathrm{C}_{\mathrm{l}}\left(\mathrm{V}_{\text {INT }}\right)} \approx 220 \mathrm{k} \Omega$


Figure 16. R $\Omega$ BUF Calculation ( $200 \Omega$ Manual Operation)

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With a low battery voltage of 6.6 V , analog common will be approximately 3.6 V above the negative supply terminal. With the integrator swinging down from analog common toward the negative supply, a 3.1 V swing will set the integrator output to 0.5 V above the negative supply.

## Capacitors - $\mathrm{C}_{\mathrm{INT}}, \mathrm{C}_{\mathrm{AZ}}$ and $\mathrm{C}_{\text {REF }}$

The integration capacitor, $\mathrm{C}_{\mathrm{INT}}$, must have low dielectric absorption. A $0.1 \mu \mathrm{~F}$ polypropylene capacitor is suggested. The auto-zero capacitor, $\mathrm{C}_{\mathrm{AZ}}$, and reference capacitor, $\mathrm{C}_{\text {REF }}$, should be selected for low leakage and dielectric absorption. Polystyrene capacitors are good choices.

## Reference Voltage Adjustment

The TC815 contains a low temperature drift internal voltage reference. The analog common potential (pin 29) is established by this reference. Maximum drift is a low 75 $\mathrm{ppm} /{ }^{\circ} \mathrm{C}$. Analog common is designed to be approximately 2.6V below $\mathrm{V}_{\mathrm{CC}}$ (pin 28). A resistive divider (R18/R19, Functional Diagram) sets the TC815 reference input voltage (REFHI, pin 34) to approximately 163.85 mV .

With an input voltage near full scale on the 200 mV range, R19 is adjusted for the proper reading.

## Display Hold Feature

The LCD will not be updated when HOLD (pin 60) is connected to GND (pin 58). Conversions are made, but the display is not updated. A HOLD mode LCD annunciator is activated when HOLD is low.

The LCD HOLD annunciator is activated through the triplex LCD driver signal at pin 13.

## Flat Package Socket

Sockets suitable for prototype work are available. A USA source is:

- Nepenthe Distribution

2471 East Bayshore, Suite 520
Palo Alto, CA 94303
(415) 856-9332

TWX: 910-373-2060
"CBQ" Socket, Part No. IC51-064-042

## Resistive Ladder Networks

Resistor attenuator networks for voltage and resistance measurements are available from:

- Caddock Electronics

1717 Chicago Avenue
Riverside, CA 92507
Tel: (714) 788-1700
TWX: 910-332-6108

| Attenuator <br> Accuracy | Attenuator <br> Type | Caddock <br> Part Number |
| :--- | :--- | ---: |
| $0.1 \%$ | Voltage | $1776-\mathrm{C} 441$ |
| $0.25 \%$ | Voltage | $1776-\mathrm{C} 44$ |
| $0.25 \%$ | Resistance | T1794-204-1 |

## 3-1/2 Digit Auto-Ranging A/D Converter with Triplex LCD Drive and Display Hold Function

## TC815

## PACKAGE DIMENSIONS



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