

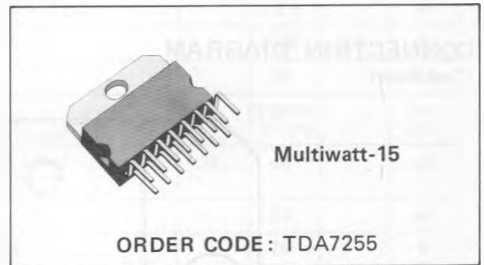
**22W FRONT REAR OR BRIDGE FULLY PROTECTED  
CAR RADIO AMPLIFIER**

- HIGH OUTPUT POWER
- POP FREE SWITCHING
- SHORT CIRCUIT PROTECTIONS:  $R_L$  SHORT - OUT TO GROUND - OUT TO  $V_S$
- MUTING  $\mu P$  COMPATIBLE
- VERY LOW CONSUMPTION STANDBY
- PROGRAMMABLE TURN ON DELAY
- LOW DISTORTION AND LOW NOISE
- DIFFERENTIAL INPUT

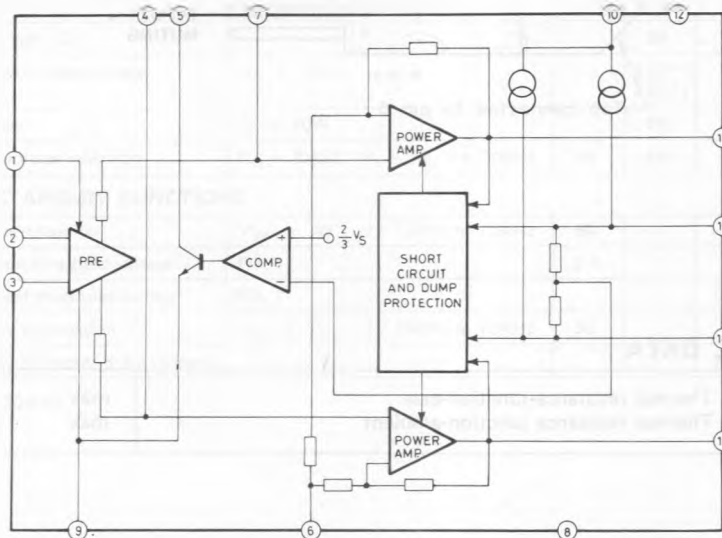
**Other Protections :**

- LOAD DUMP VOLTAGE SURGE
- LOUDSPEAKER DC CURRENT
- VERY INDUCTIVE LOAD
- OVERRATING TEMPERATURE
- OPEN GROUND

The TDA7255 a class B dual fully protected power amplifier designed for car radio applications. The device can be switched from Front-Rear to Bridge configuration by changing only the loudspeaker connection. An input fader for Front-Rear control is available. A high current capability allows to drive low impedance loads (up to  $1.6\Omega$ ).



**BLOCK DIAGRAM**



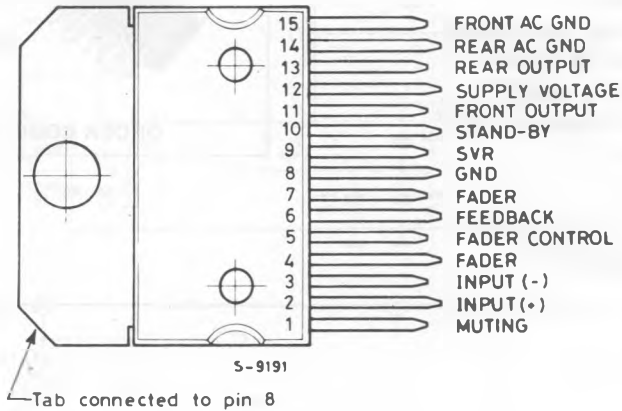
5-9109

## ABSOLUTE MAXIMUM RATINGS

$V_s$	Operating supply voltage	18	V
$V_s$	DC supply voltage	28	V
$V_s$	Peak supply voltage (for 50ms)	40	V
$I_o$	Output peak current (non repetitive $t = 0.1\text{ms}$ )	4.5	A
$I_o$	Output peak current (repetitive $f \geq 10\text{Hz}$ )	4	A
$P_{tot}$	Power dissipation at $T_{case} = 60^\circ\text{C}$	30	W
$T_{stg}, T_j$	Storage and junction temperature	-40 to 150	$^\circ\text{C}$

## CONNECTION DIAGRAM

(Top view)



## THERMAL DATA

$R_{thj-case}$	Thermal resistance junction-case	max	3	$^\circ\text{C/W}$
$R_{thj-amb}$	Thermal resistance junction-ambient	max	40	$^\circ\text{C/W}$

**ELECTRICAL CHARACTERISTICS** ( $V_s = 14.4V$ ,  $R_L = 4\Omega$ ,  $f = 1KHz$ ,  $T_{amb} = 25^\circ C$  unless otherwise specified)

Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$V_s$	Supply voltage	8		18	V
$I_d$	Total quiescent drain current		80		mA
$R_i$	Input resistance		70		$K\Omega$
$V_i$	Input saturation voltage	300			mV
$T_j$	Thermal shut down junction temperature		145		$^\circ C$

#### FRONT REAR APPLICATIONS (Fig. 2)

$P_o$	Output power	THD = 10% $R_L = 4\Omega$ $R_L = 2\Omega$ $R_L = 1.6\Omega$	5.5	6.5 11 12.5		W W W
d	Distortion	$P_o = 0.1W$ to 4W		0.05	0.5	%
$G_v$	Voltage gain			28		dB
$e_N$	Input noise voltage	$R_G = 10K\Omega$		2.5(**) 2 (*)		$\mu V$ $\mu V$
SVR	Supply voltage rejection	$R_G = 100K\Omega$ $V_r = 1V$ $f = 300Hz$	36	45		dB
CMR	Common mode rejection			55		dB
$\eta$	Efficiency	$P_o = 6.5W$ + 6.5W		70		%

#### BRIDGE APPLICATION (Fig. 1)

$V_{os}$	Output offset voltage				250	mV
$P_o$	Output power	THD = 10% $R_L = 4\Omega$ $R_L = 3.2\Omega$	18	22 25		W W
d	Distortion	$P_o = 0.1W$ to 2W		0.05		%
$G_v$	Voltage gain ( $C_L$ )			36		dB
$e_N$	Total input noise voltage	$R_G = 10K\Omega$		2.5(**) 2.0 (*)	10	$\mu V$ $\mu V$
$\eta$	Efficiency	$P_o = 20W$		66		%
SVR	Supply voltage rejection	$R_G = 10K\Omega$ , $V_r = 1V$ , $f = 300Hz$	45	58		dB

#### MUTING AND STAND-BY FUNCTIONS

Muting attenuation	$V_{ref} = 1V$ $f = 100Hz$ to 10KHz	60				dB
Muting-on threshold voltage	Pin. 1	2.4				V
Muting-off threshold voltage	Pin. 1				0.8	V
Stand-by attenuation	$V_{ref} = 1V$ $f = 100Hz$ to 10KHz	60				dB
Stand-by quiescent drain current					100	$\mu A$

(\*\*) B = 22Hz to 22KHz

(\*) B = curve A

Fig. 1 - Test and application circuit  
(Bridge amplifier)

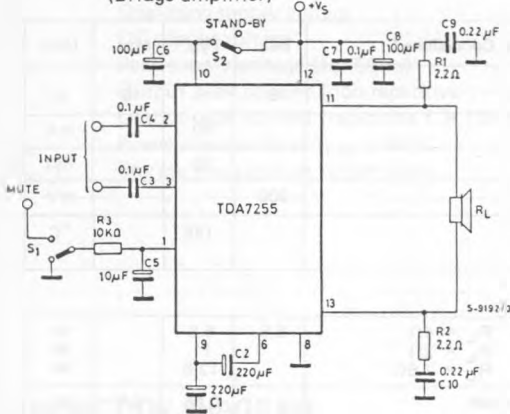
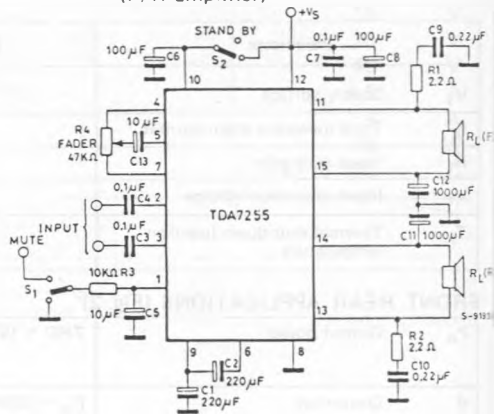
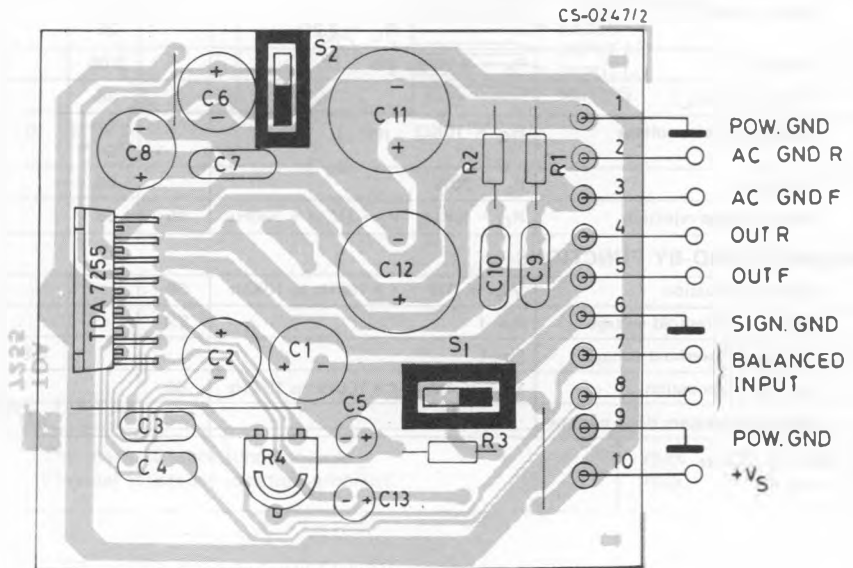


Fig. 2 - Test and application circuit  
(F/R amplifier)



- Two high impedance inputs available for balanced or unbalanced operation.
- The fader function is automatically inserted in front/rear configuration and allows the distribution of the power between the front and the rear. An external potentiometer must be connected between pins 4 and 7 with the control terminal connected to pin 5 through a decoupling capacitor. In bridge applications the pins 4-5-7 must be left open.
- Turn on delay. The output stages are muted during the turn on transient and start rising after the charge of the capacitor connected between pin 9 and ground. The capacitor also avoids pops during bridge F/R switching.

Fig. 3 - P.C. board and component layout of the circuits of Fig. 1 and 2 (1 : 1 scale)



FRONT/REAR CHARACTERISTICS

Fig. 4 - Quiescent drain current vs. supply voltage

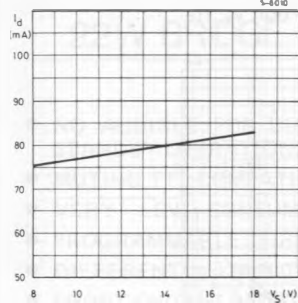


Fig. 5 - Quiescent output voltage vs. supply voltage

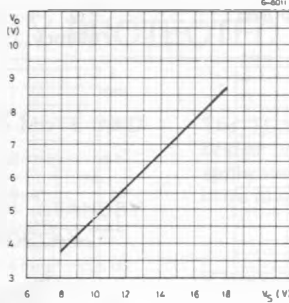


Fig. 6 - Output power vs. supply voltage

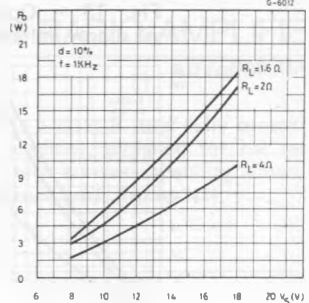


Fig. 7 - Distortion vs. frequency

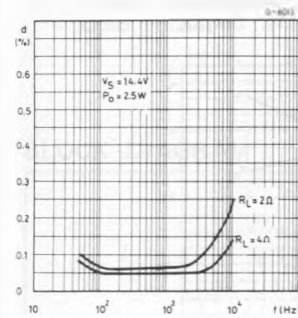


Fig. 8 - Supply voltage rejection vs. capacitor values (C2)

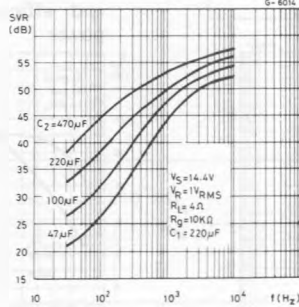


Fig. 9 - Supply voltage rejection vs. capacitor values (C1)

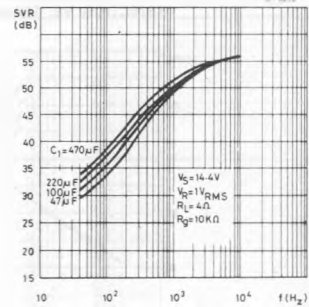


Fig. 10 - Output signal vs. fader control position

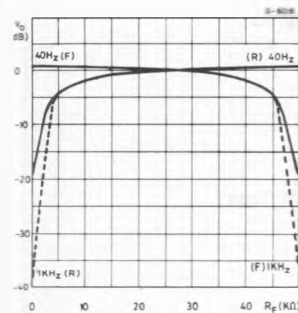


Fig. 11 - Power dissipation and efficiency vs. output power

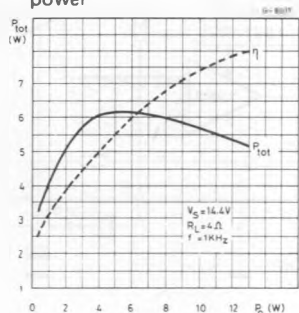
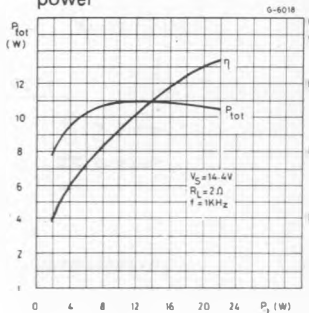


Fig. 12 - Power dissipation and efficiency vs. output power



BRIDGE CHARACTERISTICS

Fig. 13 -- Output power vs. supply voltage

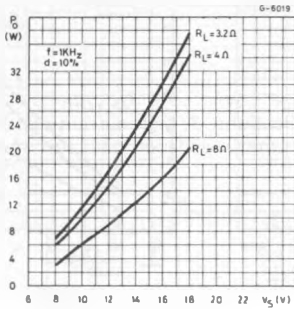


Fig. 14 - Distortion vs. frequency

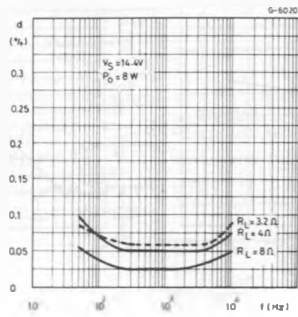


Fig. 15 - Supply voltage rejection vs. frequency

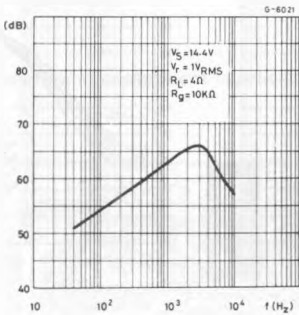


Fig. 16 - Power dissipation and efficiency vs. output power

