## 150 MHz PIXEL VIDEO CONTROLLER FOR MONITORS

## FEATURE

- 150 MHZ Pixel Rate
- 2.7 ns Rise and Fall Time
- $I^{2} C$ Bus Controlled
- Support DC Coupling Appplication only
- Grey Scale Tracking Versus Brightness
- OSD Mixing
- Beam Current Attenuation (ABL)
- Pedestral Clamping on Output Stage
- Possibility of Light or Dark Grey OSD Background
- OSD Independent Contrast Control
- Adjustable Bandwidth
- Input Black Level Clamping with Built-in Clamping Pulse
- Stand-by Mode
- 5 V to 8 V Power Supply
- Perfectly matched with the TDA9535/9535L/ 9536/9555/9556 Amplifiers from ST


## DESCRIPTION

The TDA9210 is an $I^{2}$ C Bus controlled RGB preamplifier designed for Monitor applications, able to mix the RGB signals coming from any OSD device. The usual Contrast, Brightness, Drive and Cut-Off Controls are provided.
In addition, it includes the following features:

- OSD contrast,
- Bandwidth adjustment,
- Grey background,
- Internal back porch clamping pulse generator.


The RGB incoming signals are amplified and shaped to drive in DC coupling any commonly used video amplifiers without intermediate follower stages.
As for any ST Video pre-amplifier, the TDA9210 is able to drive a real load without any external interface.
One of the main advantages of ST devices is their ability to sink and source currents while most of the devices from our competitors have problems to sink large currents.
These driving capabilities combined with an original output stage structure suppress any static current on the output pins and therefore reduce dramatically the power dissipation of the device.
Extensive integration combined with high performance and advanced features make the TDA9210 one of the best choice for any CRT Monitor in the 14 " to 17" range.
Perfectly matched with the ST video amplifiers TDA9535/9536/9555/9556s, they offer a complete solution for high performance and cost-optimized Video Board Application.

## 1 - PIN CONNECTION

| IN1 | 1 | 20 | BLK |
| :---: | :---: | :---: | :---: |
| ABL | 2 | 19 | HSYN |
| IN2 | 3 | 18 | OUT1 |
| GNDL | 4 | 17 | $V_{\text {CCP }}$ |
| IN3 | 5 | 16 | OUT2 |
| GNDA | 6 | 15 | GNDP |
| $V_{\text {CCA }}$ | 7 | 14 | OUT3 |
| OSD1 | 8 | 13 | SDA |
| OSD2 | 9 | 12 | SCL |
| OSD3 | 10 | 11 | FBLK |

## 2 - PIN DESCRIPTION

| Pin Number | Symbol |  |
| :---: | :---: | :--- |
| 1 | IN1 | Red Video Input |
| 2 | ABL | ABL Input |
| 3 | IN2 | Green Video Input |
| 4 | GNDL | Logic Ground |
| 5 | IN3 | Blue Video Input |
| 6 | GNDA | Analog Ground |
| 7 | V CCA $^{\text {OSD1 }}$ | Analog V VC (5V) |
| 8 | OSD2 | Red OSD Input |
| 9 | OSD3 | Green OSD Input |
| 10 | FBLK | Blue OSD Input |
| 11 | SCL | Fast Blanking |
| 12 | SDA | SCL |
| 13 | OUT3 | SDA |
| 14 | GNDP | Blue Video Output |
| 15 | OUT2 | Power Ground |
| 16 | VCCP | Green Video Output |
| 17 | OUT1 | Power VCC (5 V to 8 V) |
| 18 | HSYNC/BPCP | HSYNC/BPCP |
| 19 | BLK | Blanking Input |
| 20 |  |  |

## 3 - BLOCK DIAGRAM



See Figure 8 for complete BPCP and OCL generation diagram

## 4 - FUNCTIONAL DESCRIPTION

## 4.1-RGB Input

The three RGB inputs have to be supplied through coupling capacitors ( 100 nF ).
The maximum input peak-to-peak video amplitude is 1 V .
The input stage includes a clamping function. The clamp uses the input serial capacitor as a "memory capacitor".

To avoid a discharge of the serial capacitor during the line (due to leakage current), the input voltage is referenced to the ground.
The clamp is gated by an internally generated "Back Porch Clamping Pulse" (BPCP). Register 8 allows to choose the way to generate this BPCP (see Figure 1).
When bit 0 is set to 0 , the BPCP is synchronized on the trailing or leading edge of HSYNC (Pin 19) (bit $1=0$ : trailing edge, bit $1=1$ : leading edge).

Additionally, the IC automatically works with either positive or negative HSYNC pulses.

- When bit 0 is set to $1, \mathrm{BPCP}$ is synchronized on the leading edge of the blanking pulse BLK (Pin 20). One can use a positive or negative blanking pulse by programming bit 0 in Register 9 (See Table 3 on page 13).
- BPCP width can be adjusted with bit 2 and 3 (see Register 8, Table 2 on page 12 ).
- If the application already provides the Back

Porch Clamping Pulse, bit 4 must be set to 1 (providing a direct connection between Pin 19 and internal BPCP).

## 4.2 - Fast Blanking Input

The fast blanking pin (FBLK) is TTL compatible. The blanking pulse can be:

- positive or negative
- line or Composite-type (but not Frame-type).


## 4.3-Contrast Adjustment (8 bits)

The contrast adjustment is made by controlling simultaneously the gain of the three internal amplifiers through the $\mathrm{I}^{2} \mathrm{C}$ bus interface. Register 1 allows the adjustment in a range of 48 dB .

Figure 1.


## 4.4-ABL Control

The TDA9210 includes an ABL (automatic beam limitation) input to attenuate the RGB Video signals depending on the beam intensity.
The operating range is 2 V (from 3 V to 1 V ). A typical 15 dB maximum attenuation is applied to the output signal whatever the contrast adjustment is. (See Figure 2 ).
When the ABL feature is not used, the ABL input (Pin 2) must be connected to a 5 V supply voltage.

Figure 2.


## 4.5-Brightness Adjustment (8 bits)

Brightness adjustment is controlled by the $\mathrm{I}^{2} \mathrm{C}$ Bus via Register 2. It consists of adding the same DC voltage to the three RGB signals, after contrast adjustment. When the blanking pulse equals 0 , the DC voltage is set to a value which can be adjusted between 0 and 2 V with 8 mV steps (see Figure 3).
The DC output level is forced to the "Infra Black" level $\left(V_{D C}\right)$ when the blanking pulse is equal to 1 .

## 4.6 - Drive Adjustment (3 x 8 bits)

In order to adjust the white balance, the TDA9210 offers the possibility of adjusting separately the overall gain of each channel thanks to the $\mathrm{I}^{2} \mathrm{C}$ bus (Registers 3, 4 and 5).
The very large drive adjustment range ( 48 dB ) allows different standards or custom color temperatures.
It can also be used to adjust the output voltages at the optimum amplitude to drive the CRT drivers, keeping the whole contrast control for the enduser only.
The drive adjustment is located after the Contrast, Brightness and OSD switch blocks, so it does not affect the white balance setting when the BRT is adjusted. It also operates on the OSD portion of the signal.

### 4.7 Cut-off Adjustment (Infra Black)

The cut-off voltage (Infra Black: $\mathrm{V}_{\mathrm{DC}}$ ) is the output level during the blanking period. This level is sampled after each line during an internal pulse (OCL) generated during the blanking pulse (see Figure 11).
A sample-and-hold block controls the $\mathrm{V}_{\mathrm{DC}}$ level.
$V_{D C}$ is adjustable separately for each channel from 0.2 to 2.5 V via the 8 -bit cut-off registers (registers 10, 11 and 12, see Table 1 on page 15).

## Caution:

Registers 10, 11 and 12: out of the 0 to 256 cut-off adjustment steps, the first 40 steps are not allowed.

## 4.8- OSD Inputs

The TDA9210 allows to mix the OSD signals into the RGB main picture. The four pins dedicated to this function are the following:

- Three TTL RGB inputs (Pins 8, 9, 10) connected to the three outputs of the corresponding OSD processor.
- One TTL fast blanking input (Pin 11) also connected to the FBLK output of the OSD processor.
When a high level is present on the FBLK, the IC acts as follows:
- The three main picture RGB input signals (IN1, IN2, IN3) are internally switched to the internal input clamp reference voltage.
- The three output signals are set to the voltage corresponding to the three OSD input logic states (0 or 1). (See Figure 3).
If the OSD input is at low level, the output and brightness voltages $\left(\mathrm{V}_{\mathrm{BRT}}\right)$ are equal.
If the OSD input is at high level, the output voltage is $\mathrm{V}_{\text {OSD }}$, where $\mathrm{V}_{\mathrm{OSD}}=\mathrm{V}_{\mathrm{BRT}}+O S D$ and $O S D$ is an $I^{2} \mathrm{C}$ bus-controlled voltage.
OSD varies between 0 V to 4.9 V by 320 mV steps via Register 7 ( 4 bits). The same variation is applied simultaneously to the three channels providing the OSD contrast.
The grey color can be obtained on output signals when:
- OSD1 = 1, OSD2 $=0$ and OSD3 $=1$,
- A special bit (bit 5 or 6 ) in Register 9 is set to 1 .

If R9b5 is set to 1 , light grey is obtained on output. If R9b6 is set to 1 , dark grey is obtained on output. In the case where R9b5 and R9b6 are set to 0, the normal operation is provided on output signals.

## 4.9-Output Stage

The overall waveforms of the output signal are shown in Figure 3. The three output stages, which are large bandwidth output amplifiers, are able to deliver up to $4.4 \mathrm{~V}_{\mathrm{PP}}$ for $0.7 \mathrm{~V}_{\mathrm{PP}}$ on input.
When a high level is applied on the BLK input (Pin 20), the three outputs are forced to "Infra

Black" level $\left(\mathrm{V}_{\mathrm{DC}}\right)$ thanks to a sample and hold circuit (described below).
The black level (which is the output voltage outside the blanking pulse with minimum brightness and no Video input signals) is 400 mV higher than $\mathrm{V}_{\mathrm{DC}}$.
The brightness level $\left(\mathrm{V}_{\mathrm{BRT}}\right)$ is then obtained by programming register 2 (seeTable 1 on page 12).

Figure 3. Waveforms VOUT, BRT, CONT, OSD


Notes:

1. $V_{D C}=0.4$ to 2.5 V
2. $V_{B L A C K}=V_{D C}+0.4 V$
3. $\mathrm{V}_{\text {BRT }}=\mathrm{V}_{\text {BLACK }}+\mathrm{BRT}$ (with BRT $=0$ to 2 V )
4. $\mathrm{V}_{\text {CONT }}=\mathrm{V}_{\mathrm{BRT}}+\mathrm{CONT}=\mathrm{kx}$ Video $\mathrm{IN}\left(\mathrm{CONT}=4.4 \mathrm{~V}_{\mathrm{PP}}\right.$ max. for $\left.\mathrm{V}_{\mathrm{IN}}=0.7 \mathrm{~V}_{\mathrm{PP}}\right)$
5. $\mathrm{V}_{\mathrm{OSD}}=\mathrm{V}_{\mathrm{BRT}}+\mathrm{OSD}\left(\mathrm{OSD} \max .=4.9 \mathrm{~V}_{\mathrm{PP}}, O S D \min =0 \mathrm{~V}_{\mathrm{PP}}\right)$

### 4.10 - Bandwidth Adjustment

An advanced feature: Bandwidth adjustment, is implemented on the TDA9210.
For applications where rise/fall time $>5.5 \mathrm{~ns}$, this feature must not be used and the bandwidth has to be set to 0 (dec) (register 13, see Table 1 on page 15).
For applications where rise/fall time $<5.5$ ns, this feature offers several advantages:

- Depending on the external capacitive load and on the peak-to-peak output voltage, this adjustment avoids getting any slew-rate phenomenon.
- Electromagnetic radiation (EMI). Slowing down the signal of rise/fall time will decrease the EMI without significantly deteriorating the rise/fall time of the CRT driver.
- Video signal response. Using this adjustment will allow to optimize the high frequency transient phenomenons.
- Still picture mode. In this mode, high video swing is of greater interest than rise/fall time. The bandwidth adjustment can be used to avoid any slewrate phenomenon at the CRT driver output and to reduce EMI.


### 4.11 - CRT Cathode Coupling (Figure 4)

The TDA9210 is designed to be used in DC coupling mode only, enabling to build a powerful video system on a small PCB Board and giving a substantial cost saving compared with any other solution available on the market.
The preamplifier outputs control directly the cut-off levels.
The output DC level (VDC) is adjusted independently for each channel from 0.5 V to 2.5 V via the registers 10, 11 and 12.
Note that bit 2 must be set to 1 and bit 3 to 0 in Register 9 .

Figure 4. DC Coupling

### 4.12 - Stand-by Mode

The TDA9210 has a stand-by mode. As soon as the $\mathrm{V}_{\mathrm{Cc}}$ power (Pin 17) gets lower than 3V (typ.), the device is set in stand-by mode whatever the voltage on analog $\mathrm{V}_{\mathrm{CCA}}$ (Pin 7) is. The analog blocks are internally switched-off while the logic parts ( $I^{2} \mathrm{C}$ bus, power-on reset) are still supplied.
In stand-by mode, the power consumption is below 20 mW .

### 4.13-Serial Interface

The 2 -wire serial interface is an $I^{2} \mathrm{C}$ interface. The slave address of TDA9210 is DC hex.

| A6 | A5 | A4 | A3 | A2 | A1 | A0 | W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 1 | 0 | 1 | 1 | 1 | 0 | 0 |

The host MCU can write into the TDA9210 registers. Read mode is not available.
In order to write data into the TDA9210, after the "start" message, the MCU must send the following data (see Figure 5):

- the $I^{2} \mathrm{C}$ address slave byte with a low level for the R/W bit,
- the byte to the internal register address where the MCU wants to write data,
- the data.

All bytes are sent with MSB bit first. The transfer of written data is ended with a "stop" message.
When transmitting several data, the register addresses and data can be written with no need to repeat the start and slave addresses.

### 4.14 - Power-on Reset

A power-on reset function is implemented on the TDA9210 so that the $I^{2} \mathrm{C}$ registers have a deter-
mined status after power-on. The Power-on reset threshold for a rising supply on $\mathrm{V}_{\text {CCA }}$ (Pin 7) is 3.8 V (typ.) and 3.2 V when the $\mathrm{V}_{\mathrm{cc}}$ decreases.

Figure 5. $I^{2} \mathrm{C}$ Write Operation


### 4.15 - Specific Application Conditions

## Functioning with 5 V Power $\mathrm{V}_{\mathrm{cc}}$

To simplify the application, it is possible to supply the power $\mathrm{V}_{\mathrm{CC}}$ with 5 V (instead of 8 V nominal) at the expense of output swing voltage.

## Functioning without Blanking Pulse

If no blanking pulse is applied to the TDA9210, the internal BPCP can be connected to the sample and hold circuit (Register 8, bit $7=1$ and BLK pin grounded) so that the output DC level is still controlled by $1^{2} \mathrm{C}$.
To ensure the device correct behavior in the worst possible conditions, the Brightness Register must be set to 0 .

## 5 - ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter | Pin | Value | Units |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\text {CCA }}$ Max. | Supply Voltage on Analog $\mathrm{V}_{\mathrm{CC}}$ | 7 | 5.5 | V |
| $\mathrm{~V}_{\mathrm{CCP}}$ Max. | Supply Voltage on Power $\mathrm{V}_{\mathrm{CC}}$ | 20 | 8.8 | V |
| $\mathrm{~V}_{\text {in }}$ Max. | Voltage at any Input Pins (except Video inputs) and Input/Output Pins | - | 5.5 | V |
| $\mathrm{~V}_{1}$ Max. | Voltage at Video Inputs | $1,3,5$ | 1.4 | $\mathrm{~V}^{\circ}$ |
| $\mathrm{T}_{\text {stg }}$ | Storage Temperature | - | - | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {oper }}$ | Operating Junction Temperature | - | +150 | ${ }^{\circ} \mathrm{C}$ |

## 6 - THERMAL DATA

| Symbol | Parameter | Value | Units |
| :---: | :--- | :---: | :---: |
| $\mathrm{R}_{\text {th( }(\mathrm{j}-\mathrm{a})}$ | Max. Junction-ambient Thermal Resistance | 69 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{T}_{\mathrm{j}}$ | Typ. Junction Temperature at $\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}$ | 80 | ${ }^{\circ} \mathrm{C}$ |

## 7 - DC ELECTRICAL CHARACTERISTICS

$\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CCA}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{CCP}}=8 \mathrm{~V}$, unless otherwise specified.

| Symbol | Parameter | Test Condition s | Min. | Typ. | Max. | Units |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CCA}}$ | Analog Supply Voltage | Pin 7 | 4.5 | 5 | 5.5 | V |
| $\mathrm{~V}_{\mathrm{CCP}}$ | Power Supply Voltage | Pin 17 | 4.5 | 8 | 8.8 | V |
| $\mathrm{I}_{\mathrm{CCA}}$ | Analog Supply Current | $\mathrm{V}_{\mathrm{CCA}}=5 \mathrm{~V}$ |  | 70 |  | mA |
| $\mathrm{I}_{\mathrm{CCP}}$ | Power Supply Current | $\mathrm{V}_{\mathrm{CCP}}=8 \mathrm{~V}$ |  | 55 |  | mA |
| $\mathrm{~V}_{\mathrm{I}}$ | Video Input Voltage Amplitude |  |  | 0.7 | 1 | V |
| $\mathrm{Vo}_{0}$ | Output Voltage Range |  | 0.5 |  | $\mathrm{~V}_{\mathrm{CCP}}$ | V |
| $\mathrm{V}_{\text {IL }}$ | Low Level Input Voltage <br> High Level Input Voltage | OSD, FBLK, BLK, HSYNC | 2.5 V |  |  |  |
| $\mathrm{~V}_{\text {IH }}$ |  |  | 0.8 | V |  |  |
| $\mathrm{I}_{\mathrm{IN}}$ | Input Current | OSD, FBLK, BLK | -1 |  | 1 | $\mu \mathrm{~A}$ |
| $\mathrm{R}_{\text {HS }}$ | Input Resistor | HSYNC |  | 40 |  | $\mathrm{k} \Omega$ |

## 8 - AC ELECTRICAL CHARACTERISTICS

$\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CCA}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{CCP}}=8 \mathrm{~V}, \mathrm{~V}_{\mathrm{i}}=0.7 \mathrm{~V}_{\mathrm{PP},} \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}$
$R_{S}=100 \Omega$, serial between output pin and $C_{\text {LOAD }}$, unless otherwise specified.

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VIDEO INPUTS (PINS 1, 3, 5) |  |  |  |  |  |  |
| V | Video Input Voltage Amplitude | Max. Contrast and Drive |  | 0.7 | 1 | V |


| GAM | Maximum Gain | Max Contrast and Drive (CRT = DRV = 254 dec ) | 16 | dB |
| :---: | :---: | :---: | :---: | :---: |
| VOM | Maximum Video Output Voltage (Note 1) | Max Contrast and Drive (CRT = DRV = 254 dec ) | 4.4 | V |
| VON | Nominal Video Output Voltage | Contrast and Drive at POR (CRT = DRV = 180 dec ) | 2.2 | V |
| CAR | Contrast Attenuation Range | From max. Contrast (CRT=254 dec) to min. Contrast (CRT = 1 dec ) | 48 | dB |
| DAR | Drive Attenuation Range | From Max. Drive (DRV = 254 dec ) to min Drive (DRV = 1 dec ) | 48 | dB |
| GM | Gain Matching | Contrast and Drive at POR | $\pm 0.1$ | dB |
| $t_{R}, t_{F}$ | Rise Time, Fall Time (Note 1) | $\begin{aligned} & \mathrm{V}_{\text {OUT }}=2 \mathrm{~V}_{\mathrm{PP}}(\mathrm{BW}=15 \mathrm{dec}) \\ & \mathrm{V}_{\text {OUT }}=2 \mathrm{~V}_{\mathrm{PP}}(\mathrm{BW}=0 \mathrm{dec}) \end{aligned}$ | $\begin{aligned} & 2.7 \\ & 4.3 \end{aligned}$ | $\begin{aligned} & \text { ns } \\ & \text { ns } \end{aligned}$ |
| BW | Large Signal Bandwidth | $\mathrm{V}_{\text {OUT }}=2 \mathrm{~V}_{\text {PP }}$ | 130 | MHz |
| BW | Bandwidth Adjustment Range | $\begin{aligned} & \hline \mathrm{V}_{\mathrm{OUT}}=2 \mathrm{~V}_{\mathrm{PP}} \\ & \text { Minimum bandwidth }(\mathrm{BW}=0 \mathrm{dec}) \\ & \text { Maximum bandwidth }(\mathrm{BW}=15 \mathrm{dec}) \\ & \hline \end{aligned}$ | $\begin{gathered} 80 \\ 130 \end{gathered}$ | $\begin{aligned} & \mathrm{MHz} \\ & \mathrm{MHz} \end{aligned}$ |
| CT | Crosstalk between Video Outputs | $\begin{array}{ll} \hline V_{\text {OUT }}=2 \mathrm{~V}_{\text {PP }} & @ \mathrm{f}=10 \mathrm{MHz} \\ & @ \mathrm{f}=50 \mathrm{MHz} \end{array}$ | $\begin{aligned} & 60 \\ & 35 \end{aligned}$ | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \end{aligned}$ |

## VIDEO OUTPUT SIGNAL — BRIGHTNESS

| BRTmax | Maximum Brightness Level | Max. Brightness (BRT $=255 \mathrm{dec})$ <br> and Max. Drive (DRV $=254 \mathrm{dec})$ |  | 2 |  | V |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
| BRTmin | Minimum Brightness Level | Min. Brightness (BRT $=0 \mathrm{dec})$ <br> and Max. Drive (DRV $=254 \mathrm{dec})$ |  | 0 |  | V |
| VIP | Insertion Pulse |  |  | 0.4 | V |  |
| BRTM | Brightness Matching | Brightness and Drive at POR |  | $\pm 10$ | mV |  |

VIDEO OUTPUT SIGNAL — OSD

| OSDmax | Maximum OSD Output Level | Max. Drive (DRV $=254 \mathrm{dec})$ <br> Max. OSD (OSD $=15 \mathrm{dec})$ |  | 4.9 |  | V |
| :--- | :--- | :--- | :--- | :---: | :---: | :---: |
| OSDmin | Minimum OSD Output Level | Min. OSD (OSD $=0 \mathrm{dec})$ |  | 0 | V |  |

VIDEO OUTPUT SIGNAL — DC LEVEL (DC COUPLING MODE)

| DCLmax DCLmin | Maximum Output DC Level Minimum Output DC Level | Max. Cut-off (Cut-off = 255 dec ) <br> Min. Cut-off (Cut-off = 40 dec ) | $\begin{aligned} & 2.5 \\ & 0.4 \end{aligned}$ | $\begin{aligned} & \hline \mathrm{V} \\ & \mathrm{~V} \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: |
| DCLstep | Output DC Level Step |  | 10 | mV |
| DCLTD | Output DC Level Drift | Tj variation $=100^{\circ} \mathrm{C}$ | 0.5 | \% |

Note 1: Assuming that $\mathrm{V}_{\mathrm{OM}}$ remains within the range of Vo (between 0.5 V and $\mathrm{V}_{\mathrm{CCP}}-0.5 \mathrm{~V}$ )
$\mathrm{t}_{\mathrm{R}}, \mathrm{t}_{\mathrm{F}}$ are calculated values, assuming an ideal input rise/fall time of $0 \mathrm{~ns}\left(\mathrm{t}_{\mathrm{R}}=\sqrt{\text { ?ROUT? }} \quad, \mathrm{t}_{\mathrm{F}}=\quad \sqrt{\text { ?RIN }}\right.$ ?

## AC ELECTRICAL CHARACTERISTICS

$\mathrm{T}_{\text {amb }}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CCA}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{CCP}}=8 \mathrm{~V}, \mathrm{~V}_{\mathrm{i}}=0.7 \mathrm{~V}_{\mathrm{PP}}, \mathrm{C}_{\text {LOAD }}=5 \mathrm{pF}$, unless otherwise specified

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Units |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: |

ABL (PIN 2)

| GABLmin | ABL Mini Attenuation | $\mathrm{V}_{\mathrm{ABL}} \geq 3.2 \mathrm{~V}$ |  | 0 |  | dB |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
| GABLmax | ABL Maxi Attenuation | $\mathrm{V}_{\mathrm{ABL}}=1 \mathrm{~V}$ |  | 15 | dB |  |
| $\mathrm{~V}_{\mathrm{ABL}}$ | ABL Threshold Voltage | For output attenuation |  | 3 |  | V |
| IABLhigh | High ABL Input Current | $\mathrm{V}_{\mathrm{ABL}}=3.2 \mathrm{~V}$ |  | 0 |  | $\mu \mathrm{~A}$ |
| IABLIow |  |  |  |  |  |  |
| Low ABL Input Current | $\mathrm{V}_{\mathrm{ABL}}=1 \mathrm{~V}$ | -2 | $\mu \mathrm{~A}$ |  |  |  |

## $9-I^{2} \mathrm{C}$ ELECTRICAL CHARACTERISTICS

$\mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CCA}}=5 \mathrm{~V}$, unless otherwise specified

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IL }}$ | Low Level Input Voltage | On Pins SDA, SCL |  |  | 1.5 | V |
| $\mathrm{V}_{\mathrm{IH}}$ | High Level Input Voltage |  | 3 |  |  | V |
| $\mathrm{I}_{\mathrm{IN}}$ | Input Current (Pins SDA, SCL) | $0.4 \mathrm{~V}<\mathrm{V}_{\text {IN }}<4.5 \mathrm{~V}$ | -10 |  | +10 | $\mu \mathrm{A}$ |
| $\mathrm{f}_{\text {SCLIMax. }}$ | SCL Maximum Clock Frequency |  | 0.25 |  | 200 | kHz |
| $\mathrm{V}_{\text {OL }}$ | Low Level Output Voltage | SDA Pin <br> when ACK Sink Current $=6 \mathrm{~mA}$ |  |  | 0.6 | V |

## $10-I^{2} \mathrm{C}$ INTERFACE TIMING REQUIREMENTS

(see Figure 11)

| Symbol | Parameter | Min. | Typ. | Max. | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {BUF }}$ | Time the bus must be free between two accesses | 1300 |  |  | ns |
| $\mathrm{t}_{\text {HDS }}$ | Hold Time for Start Condition | 600 |  |  | ns |
| $\mathrm{t}_{\text {SUP }}$ | Set-up Time for Stop Condition | 600 |  |  | ns |
| $\mathrm{t}_{\text {LOW }}$ | The Low Period of Clock | 1300 |  |  | ns |
| $\mathrm{t}_{\text {HIGH }}$ | The High Period of Clock | 600 |  |  | ns |
| $\mathrm{t}_{\text {HDAT }}$ | Hold Time Data | 300 |  |  | ns |
| $\mathrm{t}_{\text {SUDAT }}$ | Set-up Time Data | 250 |  |  | ns |
| $\mathrm{t}_{\mathrm{R}}, \mathrm{t}_{\mathrm{F}}$ | Rise and Fall Time of both SDA and SCL | 20 |  | 300 | ns |

Figure 7. $\mathrm{I}^{2} \mathrm{C}$ Timing Diagram


## 11-I²C REGISTER DESCRIPTION

Table 1. Register Sub-addressed $-I^{2} \mathrm{C}$ Table 1

| Sub-address |  | Register Names |  | POR Value |  | Max. Value |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Hex | Dec |  |  | Hex | Dec | Hex | Dec |
| 01 | 01 | Contrast (CRT) - Note 2 | 8-bit DAC | B4 | 180 | FE | 254 |
| 02 | 02 | Brightness (BRT) | 8-bit DAC | B4 | 180 | FF | 255 |
| 03 | 03 | Drive 1 (DRV) - Note 2 | 8-bit DAC | B4 | 180 | FE | 254 |
| 04 | 04 | Drive 2 (DRV) - Note 2 | 8-bit DAC | B4 | 180 | FE | 254 |
| 05 | 05 | Drive 3 (DRV) - Note 2 | 8-bit DAC | B4 | 180 | FE | 254 |
| 06 | 06 | Not Used |  | - | - | - |  |
| 07 | 07 | OSD Contrast (OSD) | 4-bit DAC | 09 | 09 | OF | 15 |
| 08 | 08 | BPCP \& OCL | Refer to the ${ }^{2} \mathrm{C}$ table 2 | 04 | 04 |  |  |
| 09 | 09 | Miscellaneous | Refer to the ${ }^{2} \mathrm{C}$ table 3 | 1C | 28 |  |  |
| 0A | 10 | Cut Off Out 1 DC Level (Cut-off) - Note 3 | 8-bit DAC | B4 | 180 | FF | 255 |
| OB | 11 | Cut Off Out 2 DC Level (Cut-off) - Note 3 | 8-bit DAC | B4 | 180 | FF | 255 |
| OC | 12 | Cut Off Out 3 DC Level (Cut-off) - Note 3 | 8-bit DAC | B4 | 180 | FF | 255 |
| OD | 13 | Bandwidth Adjustment (BW) - Note 4 | 4-bit DAC | 07 | 07 | OF | 15 |

Note 2: For Contrast \& Drive adjustment, code 00 (dec) and 255 (dec) are not allowed.
Note 3: For Cut Off Output DC Level, codes below 40 (dec) are not allowed
Note 4: To be set to 0 (dec) for application with $\mathrm{t}_{\mathrm{R}} / \mathrm{t}_{\mathrm{F}}>5.5 \mathrm{~ns}$
Table 2. BPCP \& OCL Register (R8) - $I^{2} \mathrm{C}$ Table 2 (see also Figure 8)

| b7 | b6 | b5 | b4 | b3 | b2 | b1 | b0 | Function | POR Value |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :--- | :---: |
|  |  |  | 0 |  |  |  | 0 | Internal BPCP triggered by HSYNC | x |
|  |  |  | 0 |  |  |  | 1 | Internal BPCP triggered by BLK | x |
|  |  |  | 0 |  |  | 0 |  | Internal BPCP synchronized by the trailing edge |  |
|  |  |  | 0 |  |  | 1 |  | Internal BPCP synchronized by the leading edge |  |
|  |  |  | 0 | 0 | 0 |  |  | Internal BPCP Width $=0.33 \mu \mathrm{~s}$ | x |
|  |  |  | 0 | 0 | 1 |  |  | Internal BPCP Width $=0.66 \mu \mathrm{~s}$ |  |
|  |  |  | 0 | 1 | 0 |  |  | Internal BPCP Width $=1 \mu \mathrm{~s}$ | x |
|  |  |  | 0 | 1 | 1 |  |  | Internal BPCP Width $=1.33 \mu \mathrm{~s}$ | x |
|  |  |  | 1 |  |  |  |  | Internal BPCP $=$ BPCP input |  |
|  |  | 0 |  |  |  |  |  | Normal Operation | x |
|  |  | 1 |  |  |  |  |  | Reserved (Force BPCP to 1 in test) |  |
|  | 0 |  |  |  |  |  |  | Normal Operation |  |
|  | 1 |  |  |  |  |  |  | Reserved (Force OCL to 1 in test) |  |
| 0 |  |  |  |  |  |  |  | Internal OCL pulse triggered by BLK |  |
| 1 |  |  |  |  |  |  |  | Internal OCL pulse $=$ Internal BPCP |  |

Table 3. Miscellaneous Register (R9) - $I^{2} \mathrm{C}$ Table 3

| b7 | b6 | b5 | b4 | b3 | b2 | b1 | b0 | Function | POR Value |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :--- | :---: |
|  |  |  |  |  |  |  | 0 | Positive Blanking Polarity | x |
|  |  |  |  |  |  |  | 1 | Negative Blanking Polarity | x |
|  |  |  |  |  |  | 0 |  | Soft Blanking = OFF |  |
|  |  |  |  |  |  | 1 |  | Soft Blanking = ON |  |
|  |  |  | x | 0 | 1 |  |  | bit 3 and bit 2 must be set to 0 and 1 respectively (Note 5) |  |
|  | 0 | 0 |  |  |  |  |  | Light Grey on OSD Outputs = OFF | x |
|  | 0 | 1 |  |  |  |  |  | Light Grey on OSD Outputs = ON | x |
|  | 0 | 0 |  |  |  |  |  | Dark Grey on OSD Outputs = OFF | x |
|  | 1 | 0 |  |  |  |  |  | Dark Grey on OSD Outputs = ON |  |
| - |  |  |  |  |  |  |  | Reserved (SOG = OFF) |  |
| - |  |  |  |  |  |  |  | Reserved (SOG = ON) |  |

Note 5: After Power ON, bit3 and bit 2 of register 9 must mandatorily be set to 0 and 1 respectively.
Table 4. Bandwidth Adjustment (R13) $-I^{2} \mathrm{C}$ Table 4

| b7 | b6 | b5 | b4 | b3 | b2 | b1 | b0 | Function | POR Value |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :--- | :---: |
|  |  |  |  | 1 | 1 | 1 | 1 | 130 MHz | x |
|  |  |  |  | 0 | 1 | 1 | 1 | 100 MHz | x |
|  |  |  |  | 0 | 0 | 0 | 0 | 80 MHz (Note 6) |  |
|  |  | 0 | 0 |  |  |  |  | Normal Operation |  |
|  |  | 0 | 1 |  |  |  |  | BW DAC output connected to BLK input (for test) |  |
|  |  | 1 | 0 |  |  |  |  | BW DAC complementary output connected to BLK input (for <br> test) |  |

Note 6: For application with $t_{R} / t_{F}>5.5 \mathrm{~ns}$, this register has to be set to 0 (dec).
Figure 8. BPCP and OCL Generation


## 12-INTERNAL SCHEMATICS

Figure 9.


Figure 10.


Figure 11.


Figure 12.


Figure 13.


Figure 14.


Figure 15.


Figure 16.


Figure 17.


Figure 18. TDA9210-TDA9535/9536 Demonstration Board: Silk Screen and Trace


Figure 19. Amplifier and Preamplifier Outputs. Trace Routing (detail)


Figure 20. TDA9210 - TDA9535/9536 Demonstration Board Schematic


## 13 - PACKAGE MECHANICAL DATA

20 Pins - Plastic Dip


Tie Bar Center

| Dimensions | Millimeters |  |  | Inches |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. | Min. | Typ. | Max. |
| A | 3.25 | 3.30 | 3.35 | 0.128 | 0.130 | 0.132 |
| a1 |  | 0.508 |  |  | 0.020 |  |
| B | 1.39 |  | 1.65 | 0.055 |  | 0.065 |
| b | 0.381 | 0.457 | 0.533 | 0.015 | 0.018 | 0.021 |
| b1 | 0.20 | 0.254 | 0.30 | 0.008 | 0.010 | 0.012 |
| C | 5.20 | 5.33 | 5.46 | 0.205 | 0.210 | 0.215 |
| D | 24.9 | 25.15 | 25.4 | 0.980 | 0.990 | 1.000 |
| E | 7.8 | 8.5 | 9.1 | 0.307 | 0.335 | 0.358 |
| e | 2.29 | 2.54 | 2.79 | 0.090 | 0.100 | 0.110 |
| e3 | 22.60 | 22.86 | 23.11 | 0.890 | 0.900 | 0.910 |
| e4 | 7.36 | 7.62 | 7.87 | 0.290 | 0.300 | 0.310 |
| F | 6.22 | 6.35 | 6.50 | 0.245 | 0.250 | 0.255 |
| L | 3.42 | 3.68 | 3.93 | 0.135 | 0.145 | 0.155 |
| R1 | 3.17 | 3.30 | 3.42 | 0.125 | 0.130 | 0.135 |
| R2 |  | $7 d$ |  |  | 70 | 0.006 |
| K | 0.152 |  |  | 0.030 |  |  |
| K1 |  | 0.762 |  |  | 0.060 | 0.030 |

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