SGS-THOMSON MICROELECTRONICS

BU931Z/ZP/ZPFI

NPN POWER DARLINGTON

- HIGH RUGGEDNESS
- INTEGRATED HIGH VOLTAGE ZENER

AUTOMOTIVE MARKET

 APPLICATION IN HIGH PERFORMANCE ELECTRONIC CAR IGNITION





DESCRIPTION

The BU931Z. BU931ZP and BU931ZPFI are silicon multiepitaxial biplanar NPN transistors in monolithic darlington configuration mounted respectively in TO-3 metal case, SOT-93 plastic package and ISO-WATT218 fully isolated package.

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value			Unit
VCBO	Collector-base Voltage (I _E = 0)	350			V
VCER	Collector-emitter Voltage ($R_{BE} = 100 \Omega$)	350			V
VCES	Collector-emitter Voltage (V _{BE} = 0)	350			V
VCEO	Collector-emitter Voltage (I _B = 0)	350			V
VEBO	Emitter-base Voltage (I _C = 0)	5			V
I _C	Collector Current	20			A
I _B	Base Current	5			A
		TO-3	SOT-93	ISOWATT218	
Ptot	Total Dissipation at T _c ≤ 25 °C	175	125	60	W
Tstg	Storage Temperature	- 40 to 200	- 40 to 150	- 40 to 150	°C
T,	Max. Operating Junction Temperature	200	150	150	°C

BU931Z/ZP/ZPFI

THERMAL DATA

			TO-3	SOT-93	ISOWATT218	
R _{th j-case}	Thermal Resistance Junction-case	Max	1	1	2.08•	°C/W

ELECTRICAL CHARACTERISTICS ($T_{case} = 25 \ ^{\circ}C$ unless otherwise specified)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
I _{CL}	Clamping Current	V _{CE} = 350 either or	I _B = 0 V _{BE} = 0			250 250	μΑ μΑ
ICE(off)	Collector-emitter off State Current (I _B = 0)	V _{CC} = 16 V V _{BE} = 300 m ³				0.5	mA
I _{EBO}	Emitter Cutoff Current (I _C = 0)	V _{EB} = 5 V				50	mA
Vc⊾	Clamping Voltage	either and same	$I_B = 0 \text{ or } V_{BE} = (I_C = 100 \text{ mA})$ $T_1 = 125 \text{ °C}$	350 350		500 500	V V
V _{CE(sat)} *	Collector-emitter Saturation Voltage	$I_{C} = 7 A$ $I_{C} = 8 A$ $I_{C} = 10 A$	l _B = 70 mA l _B = 100 mA l _B = 150 mA		1.25 1.45 1.65	1.6 1.8 2	V V V
		$I_{\rm C} = 8$ A	I _B = 70 mA I _B = 100 mA I _B = 150 mA		1.6 1.8 2		V V V
V _{BE(sat)} *	Base-emitter Saturation Voltage	$I_{\rm C} = 8 \text{ A}$ $I_{\rm C} = 10 \text{ A}$	I _B = 100 mA I _B = 250 mA			2.2 2.5	V V
V _{BE(on)} *	Base-emitter Voltage	$I_{C} = 5 A$ $T_{j} = -40 °C$ $T_{j} = 125 °C$ $I_{C} = 10 A$ $T_{j} = -40 °C$ $T_{j} = 125 °C$	V _{CE} = 2 V V _{CE} = 2 V	1.1	1.67 2	2.1 2.4	V V V V V V
VF*	Diode Forward Voltage	I _F = 10 A				2.5	V
E _{s/b}	Second Breakdown Energy Unclamped	L = 10 mH	I _C = 10 A		500		mJ
I _{s/b}	Second Breakdown Collector Current	$V_{GE} = 30$ t = 500 ms t = 250 ms t = 250 ms	for BU931Z for BU931ZP for BU931ZPFI	6 4 1.7			A A A
	USE TEST (see fig. 2)	V _{CC} = 24 V	L = 7 mH	8			A

* Pulsed : pulse duration = 300 μs, duty cycle = 1.5 %.



Collector-emitter Saturation Voltage.



Base-emitter Saturation Voltage.



DC Current Gain.



Collector-emitter Saturation Voltage.



Base-emitter Saturation Voltage.



Collector-emitter Saturation Voltage.



Figure 1 : Functional Test Circuit.







ISOWATT 218 PACKAGE CHARACTE-RISTICS AND APPLICATION

ISOWATT218 is fully isolated to 4000 V dc. Its thermal impedance, given in the data sheet, is optimized to give efficient thermal conduction together with excellent electrical isolation.

The structure of the case ensures optimum distances between the pins and heatsink. These distances are in agreement with VDE and UL creepage and clearance standards. The ISOWATT218 package eliminates the need for external isolation so reducing fixing hardware.

The package is supplied with leads longer than the standard TO-218 to allow easy mounting on pcbs. Accurate moulding techniques used in manufacture assures consistent heat spreader-to-heatsink capacitance.

ISOWATT218 thermal performance is better than that of the standard part, mounted with a 0.1 mm mica washer. The thermally conductive plastic has a higher breakdown rating and is less fragile than mica or plastic sheets. Power derating for ISO-WATT218 packages is determined by :



THERMAL IMPEDANCE OF ISOWATT 218 PACKAGE

Fig. 3 illustrates the elements contributing to the thermal resistance of transistor heatsink assembly, using ISOWATT218 package.

The total thermal resistance $R_{th(tot)}$ is the sum of each of these elements.

The transient thermal impedance, Z_{th} for different pulse durations can be estimated as follows :

1. for a short duration power pulse less than 1 ms ;

2. for an intermediate power pulse of 5 ms to 50 ms :

$$Z_{th} = R_{thJ-C}$$

3. for long power pulses of the order of 500 ms or greater :

It is often possible to discern these areas on transient thermal impedance curves.

Figure 3.

R thJ-C R thC-HS R thHS-amb

