

SGSF341/IF341 SGSF441/IF441/F541

FASTSWITCH HOLLOW-EMITTER NPN TRANSISTORS

- HIGH SWITCHING SPEED NPN POWER TRANSISTORS
- HOLLOW EMITTER FOR FAST SWITCHING
- HIGH VOLTAGE FOR OFF-LINE APPLICA-TIONS
- TOKHZ SWITCHING SPEED
- LOW COST DRIVE CIRCUITS
- LOW DYNAMIC SATURATION

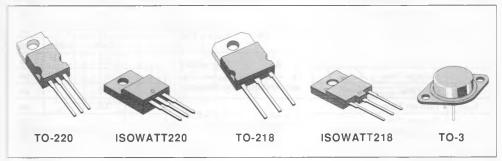
IPPLICATIONS

SMPS

DESCRIPTION

ow emitter FASTSWITCH NPN power transisrs have been specifically designed for 220V (and V with input doubler) off-line switching power supply applications. Hollow emitter transistors can operate up to 70kHz with simple drive circuits which helps to simplify design and improve reliability. These transistors are suitable for applications in bridge and two transistor forward medium power converters, 450W to 900W. When used in conjunction with a low voltage Power MOSFET in emitter switch configuration in flyback and forward converters, they can operate at up to 100kHz.

These hollow emitter FASTSWITCH transistors are available in TO-220, TO-218, ISOWATT220 and ISOWATT218 packages. The ISOWATT218 conforms to the creepage distance and isolation requirements of VDE, IEC, and UL specifications. Additionally these FASTSWITCH transistors are available in metal TO-3 packages.



ABSOLUTE MAXIMUM RATINGS

Symbol	D	SGS					11-14
	Parameter	F341	IF341	F441	IF441	F541	Unit
V _{CES}	Collector - Emitter Voltage (V _{BE} = 0)		850		V		
V _{CEO}	Collector - Emitter Voltage (I _B = 0)	400			V		
VEBO	Emitter - Base Voltage (I _C = 0)	7			V		
Ic	Collector Current	10			Α		
I _{CM}	Collector Peak Current (tp < 5ms)	15		Α			
IB	Base Current	6			Α		
I _{BM}	Base Peak Current (tp < 5ms)	10			Α		
Ptot	Total Dissipation at T _c ≤ 25°C	85	40	95	55	115	W
T _{stg}	Storage Temperature - 65 to	150	150	150	150	175	°C
T	Junction Temperature	150	150	150	150	175	°C

THERMAL DATA

			SGS					
			F341	IF341	F441	IF441	F541	
R _{thj-case}	Thermal Resistance Junction-case	Max	1.47	3.12	1.31	2.27	1.3	°C/W

ELECTRICAL CHARACTERISTICS (T_{case} = 25°C unless otherwise specified)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
ICES	Collector Cutoff Current (V _{BE} = 0)	V _{CE} = 850V				200	μА
I _{CEO}	Collector Cutoff Current (I _B = 0)	V _{CE} = 380V V _{CE} = 400V				200	μA mA
IEBO	Emitter Cutoff Current (I _C = 0)	V _{EB} = 7V				1	mA
V _{CEO (sus)*}	Collector Emitter Sustaining Voltage	I _C = 0.1A		400			٧
V _{CE} (sat)*	Collector Emitter Saturation Voltage	I _C == 6A I _C = 4A	I _B = 1.2A I _B = 0.6A			1.5 1.5	V
V _{BE (sat)} ,	Base Emitter Saturation Voltage	I _C = 6A I _C = 4A	I _B = 1.2A I _B = 0.6A			1.5 1.5	V

RESISTIVE LOAD

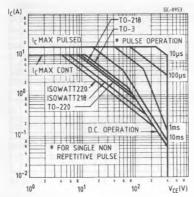
Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
ton	Turn-on Time		$V_{CC} = 250V$ $I_{B2} = -2I_{B1}$		0.5	1	μs
ts	Storage Time	I _C = 6A			1.6	2.5	μs
t _f	Fall Time	181 - 1.27			0.25	0.35	μs
ton	Turn-on Time	Ic = 6A	$V_{CC} = 250V$ $I_{B2} = -2I_{B1}$ ration Network		0.5		μS
t _s	Storage Time				1.1		μS
t _f	Fall Time	with Antisatur			0.2		μS
ton	Turn-on Time		V _{CC} = 250V V _{BE(off)} = -5V		0.5		μs
ts	Storage Time	$I_{C} = 6A$ $I_{B1} = 1.2A$			1.4		μs
tr	Fall Time	181 - 1.27			0.1		μs

INDUCTIVE LOAD

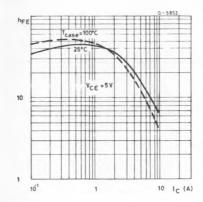
Symbol	Parameter	Test C	Test Conditions		Тур.	Max.	Unit
ts	Storage Time	I _C = 6A	h _{FE} = 5		1.4	2.8	μs
t _f	Fall Time	V _{CL} = 350V L = 300μH	$V_{BE(off)} = -5V$ $R_{B(off)} = 1.2\Omega$		0.1	0.2	μs
ts	Storage Time	I _C = 6A V _{CL} = 350V	h _{FE} = 5			4	μs
tı	Fall Time	L = 300μH T _c = 100°C	$V_{BE(off)} = -5V$ $R_{B(off)} = 1.2\Omega$			0.3	μs

Pulsed : Pulse duration = 300µs, duty cycle = 1.5%

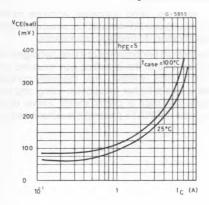
Sare Operating Areas



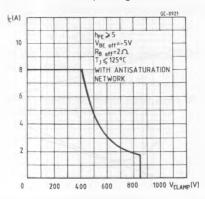
C Current Gain



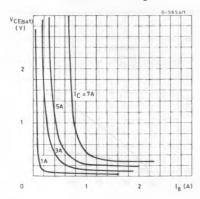
Collector-emitter Saturation Voltage



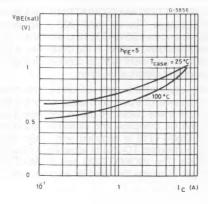
Reverse Biased Safe Operating Area



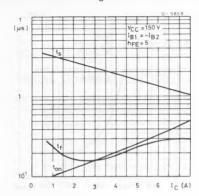
Collector-emitter Saturation Voltage



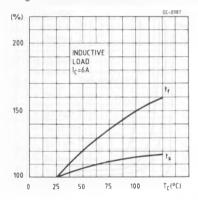
Base-emitter Saturation Voltage



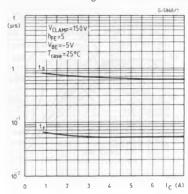
Resistive Load Switching Times



Switching Times Percentance Variation



Inductive Load Switching Times



ISOWATT PACKAGES CHARACTERISTICS AND APPLICATION

The ISOWATT220 and ISOWATT218 are fully isolated packages. The ISOWATT220 is isolated to 2000V dc and the ISOWATT218 to 4000V dc. Their thermal impedence, given in the datasheet, is optimised to give efficient thermal conduction together with excellent electrical isolation.

The structure of the case ensures optimum distances between the pins and heatsink. For the ISO-WATT218 these distances are in agreement with VDE and UL creepage and clearance standards. The ISOWATT218 is supplied with longer leads than the standard TO-218 to allow easy mounting on PCB's. The ISOWATT220 and ISOWATT218 packages eliminate the need for external isolation

so reducing fixing hardware. Accurate moulding techniques used in manufacture assures consistent heat spreader-to-heatsink capacitance.

The thermal performance of these packages is better than that of the standard part mounted with a 0.1mm mica washer. The thermally conductive plastic has a higher breakdown rating and is less fragile than mica or plastic sheets. Power derating for these ISOWATT packages is determined by :

$$P_D = \frac{T_j - T_c}{R_{th}}$$

THERMAL IMPEDANCE OF ISOWATT PACKAGES

Fig. 1 illustrates the elements contributing to the thermal resistance of a transistor heatsink assembly, using ISOWATT packages.

The total thermal resistance $R_{th(tot)}$ is the sum of each of these elements. The transient thermal impedance, Z_{th} for different pulse durations can be estimated as follows:

1 - For a short duration power pulse of less than

Zth < Rth.I-C

2 - For an intermediate power pulse of 5ms to 50ms seconds :

 $Z_{th} = R_{thJ-C}$

3 - For long power pulses of the order of 500ms seconds or greater :

 $Z_{th} = R_{thJ-C} + R_{thC-HS} + R_{thHS-amb}$

It is often possible to discern these areas on transient thermal impedance curves.

Figure 1.

R_{thJ-C} R_{thC-HS} R_{thHS-amb}